ISSN 2782-2427

# CONTROL SCIENCES 5/2021



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#### **CONTROL SCIENCES** Scientific Technical Journal

6 issues per year ISSN 2782-2427 **Open access** 

Published since 2021

**Original Russian Edition** Problemy Upravleniya Published since 2003

FOUNDER AND PUBLISHER V.A. Trapeznikov **Institute of Control Sciences** of Russian Academy of Sciences

Editor-in-Chief D.A. Novikov, Corr. Member of RAS

**Deputy Editor-in-Chief** F.F. Pashchenko

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URL: http://controlsciences.org

Published: October 28, 2021

Registration certificate of Эл № ФС 77-80482 of 17 February 2021 issued by the Federal Service for Supervision of Communications, Information Technology, and Mass Media

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## ANALYSIS OF INFORMATION INCONSISTENCY IN BELIEF FUNCTION THEORY. PART I: EXTERNAL CONFLICT<sup>1</sup>

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**Abstract**. The analysis results of information inconsistency within belief function theory (the Dempster–Shafer theory of evidence) are reviewed. This theory has been intensively developing over the past 10–15 years. Part I of the survey considers the measure of external conflict between bodies of evidence. The concepts of conflict and non-conflict bodies of evidence and the basic requirements applied to measures of external conflict are discussed. Different axioms of the measure of external conflict are analyzed. The general forms of measures of external conflict that satisfy the system of axioms are given. Different methods for constructing measures of external conflict (metric, algebraic, and structural approaches; evaluation by combining rules) are presented. The robust estimation of external conflict, the relationship between its measure and the metric on the set of bodies of evidence, and the consistency of combining rules and measures of external conflict are discussed. Many illustrative examples are provided.

Keywords: theory of belief functions, combining rules, inconsistency of bodies of evidence, measure of external conflict.

#### INTRODUCTION

In many problems of data analysis and decisionmaking, information coming from various sources (experts, analysts, monitoring systems, etc.) can be conveniently described using the so-called bodies of evidence, the main concept in belief function theory (the Dempster-Shafer theory of evidence) [12]. A body of evidence is an aggregate of a set of certain subsets (the so-called focal elements) from a universal set and a mass function defined on these subsets. At present, belief function theory finds numerous applications in data analysis, processing of expert information, decision-making, etc. This is due to the understandable and interpretable foundations of the theory itself, associated with the classical probabilistic approach, and a wide range of tools to model and evaluate important factors such as the uncertainty and inaccuracy of information, the reliability of information sources, the inconsistency of information from different sources, the amount of ignorance, etc. In addition,

trust function theory has a well-developed apparatus for combining bodies of evidence (combining rules suggested by Dempster, Yager, Inagaki, and others) considering the factors mentioned.

This paper presents the main results related to analyzing the inconsistency (conflict) of information sources within belief function theory that have been obtained over the past 15 years. The inconsistency of information obtained from different sources is their important a priori characteristic, which should be considered when deciding on the possibility of aggregating these sources or choosing sources from a set for aggregation. For example, one analyst predicts that in a month, the company's stock will have a value within the interval [40, 50]; according to another's opinion, a value within the interval [70, 85] (in conventional units). In this case, we have a large (external) conflict between their evidence.

Although the very concept of conflict (inconsistency) of information sources appeared in the pioneering works on the theory of evidence, the study of the concept of conflict using belief function theory has recently attracted the attention of many researchers, becoming a separate scientific direction. Within this direction, the axiomatics of measures of external and inter-

<sup>&</sup>lt;sup>1</sup>This work was supported by the Russian Foundation for Basic Research, project no. 20-11-50077.



nal conflict, different methods for estimating the conflict, and the properties of measures of conflict were studied; several practical problems related to estimating the conflict were considered. In particular, estimating the amount of information conflict finds application in conflict management [3], the analysis of expert information and recommendations [4–6], the analysis of voting results [7], the design of trading strategies [8], image detection [9], etc. In addition, the concept of a metric between bodies of evidence, related to conflict, is used to approximate bodies of evidence [10-13] (to replace a "complex" body of evidence with a large number of focal elements by a body of evidence with a smaller number of focal elements), to classify data [14], to cluster data [15], etc. Due to the variety of such problems, this paper does not consider the applied aspects of inconsistency analysis within belief function theory.

The remainder of this paper is organized as follows. Section 1 recalls the fundamentals of belief function theory. In Section 2, we discuss combining rules for bodies of evidence. Section 3 deals with the concepts of conflict and non-conflict bodies of evidence and requirements applied to measures of external conflict. In Section 4, we analyze different axioms for a measure of conflict and present its general form satisfying a certain system of axioms. Section 5 considers different methods for estimating external conflict: metric (Section 5.1), structural (Section 5.2), algebraic (Section 5.3), and combining rules-based (Section 5.4). In Section 6, we describe a robust conflict estimation procedure. In the Conclusions, we summarize some findings of this study.

In addition to the conflict between the bodies of evidence, researchers consider the conflict of information provided by the same evidence (internal conflict). Part II of the survey will be devoted to this concept.

#### 1. FUNDAMENTALS OF THE DEMPSTER—SHAFER THEORY OF EVIDENCE

Let  $X = \{x_1, ..., x_n\}$  be a finite set, and  $2^X$  be the set of all subsets from X. In the Dempster–Shafer theory of evidence, a *basic belief assignment* (BBA, also termed a mass function) is a set function m:  $2^X \rightarrow [0, 1]$  that satisfies the condition  $\sum_{A \in 2^X} m(A) = 1$ .

As a rule, the normalized BBAs are considered:  $m(\emptyset) = 0$ . Unnormalized BBAs are studied within the so-called *Transferable Belief Model* introduced in the paper [16]. In this case, a value  $m(\emptyset) > 0$  is interpreted as a measure of belief that the true value  $x \notin X$ . A subset  $A \subseteq X$  is called a focal element of a BBA m if m(A) > 0. A pair  $F = (\mathcal{A}, m)$  composed of the set of all focal elements  $\mathcal{A} = \{A\}$  and a corresponding BBA m(A),  $A \in \mathcal{A}$ , is called a body of evidence. We denote by  $\mathcal{F}(X)$  the set of all bodies of evidence on X and by  $\mathcal{P}(X)$  the set of all probability measures on X.

If a body of evidence  $F = (\mathcal{A}, m)$  is known, the Dempster-Shafer theory operates some set functions. Among them, the most important ones are the belief function  $Bel(A) = \sum_{B \subseteq A} m(B)$  and its dual called the plausibility function  $Pl(A) = 1 - Bel(A^c)$ , where  $A^c$ indicates the complement of the set A. A mass function can be uniquely restored by the belief function Möbius transform: using the m(A) = $\sum_{B \subseteq A} (-1)^{|A \setminus B|} Bel(B)$ . The inequality  $Bel(A) \leq Pl(A)$ holds  $\forall A \subseteq X$ , and the length of the interval [Bel(A), Pl(A)] determines the degree of uncertainty of the event A; see [4]. The belief and plausibility functions will be denoted by  $Bel_F$  and  $Pl_F$ , respectively, whenever their dependence on the body of evidence  $F = (\mathcal{A}, m)$  should be emphasized. The function  $Pl(x) = \sum_{A \in \mathcal{A}: x \in A} m(A)$ ,  $x \in X$ , is called the contour function of a body of evidence.

A body of evidence  $F = (\mathcal{A}, m)$  and the corresponding set functions can be defined in vector form: **m** is a  $2^{|x|}$ -dimensional vector composed of the values of the mass function m(A),  $A \in 2^x$ , for some ordering of all subsets from  $2^x$ ; **Pl** is a  $2^{|x|}$ -dimensional vector composed of the values Pl(A),  $A \in 2^x$ ; and so on.

There are two main interpretations of BBAs and, therefore, two approaches to constructing belief function theory. According to the first interpretation, dating back to the paper [1], m(A) is the probability of a random event (random set) A; for details, see [17]. In the statistical sense, a BBA is treated as the relative frequency of the fact that the true alternative belongs to the set A: m(A) = q(A)/N, where q(A) is the number of observed sets  $A \subseteq X$ , and N is the total number of observations. In an alternative approach developed in the book [2], belief and plausibility functions are first introduced as set functions that satisfy the weakened axiomatics of a probability measure. These functions are then used to estimate the degree of uncertainty of information sources (evidence). Therefore, the Dempster-Shafer theory is also called the theory of evidence or belief function theory.

As shown in the paper [18],

$$Bel(A) = \inf_{P \in \mathcal{P}_{Bel}} P(A), \ Pl(A) = \sup_{P \in \mathcal{P}_{Bel}} P(A), \ A \subseteq X$$

where  $\mathcal{P}_{Bel}$  denotes the set of probability measures  $P \in \mathcal{P}(X)$  agreed with Bel, i.e.,  $Bel(A) \leq P(A)$  $\forall A \subseteq X$ . The converse does not hold: the lower bound of a set of probability measures is not necessarily a belief function. The set  $\mathcal{P}_{Bel}$  is also called the credal set of the function Bel.

**Example 1.** Suppose that ten experts predict the prospects for developing three technologies  $X = \{a, b, c\}$ : three experts spoke in favor of technologies a or b, two in favor of technology b, four in favor of technologies b or c, and one in favor of technology c. Without a known distribution between the alternatives (the uncertainty of information), the mass functions are given by

$$m(\{a, b\}) = \frac{3}{10}, \ m(\{b, c\}) = \frac{4}{10},$$
$$m(\{b\}) = \frac{2}{10}, \ m(\{c\}) = \frac{1}{10}.$$

Thus, we have the body of evidence  $F = (\mathcal{A}, m)$  with the set of focal elements  $\mathcal{A} = \{\{a, b\}, \{b, c\}, \{b\}, \{c\}\}\}$ . This body of evidence can be used to find the values Bel(A) and  $Pl(A) \forall A \subseteq X$ . In particular,  $Bel(\{a\}) = 0$ ,  $Pl(\{a\}) = 0.3$ ,  $Bel(\{b\}) = 0.2$ ,  $Pl(\{b\}) = 0.9$ ,  $Bel(\{c\}) = 0.1$ , and  $Pl(\{c\}) = 0.5$ . These values can be considered lower and upper bounds for the probability of good prospects for developing particular technologies:  $0 \le P(\{a\}) \le 0.3$ ,  $0.2 \le P(\{b\}) \le 0.9$ , and  $0.1 \le P(\{c\}) \le 0.5$ . Note that the greatest uncertainty exists in the forecast of technology b.

The so-called *pignistic probability*  $Bet_F \in \mathcal{P}(X)$  can be assigned to each body of evidence  $F = (\mathcal{A}, m)$  [19]. This characteristic is the probability of an event  $x_i \in X$ , i = 1, ..., n, provided that the random variables within the focal elements obey the uniform distribution:

$$Bet_F(x_i) = \sum_{A \in \mathcal{A}, x_i \in A} \frac{m(A)}{|A|}, \ i = 1, ..., n.$$

These values coincide with the Shapley values

$$Bet_{F}(x_{i}) = \sum_{A \in \mathcal{A}, x_{i} \in A} \frac{\left(n - |A|\right)! \left(|A| - 1\right)!}{n!} \left(Bel(A) - Bel(A \setminus \{x_{i}\})\right)$$

of the corresponding belief function Bel; see [20]. The pignistic probability of an arbitrary set  $B \subseteq X$  is given by  $Bet_F(B) = \sum_A \frac{|A \cap B|}{|A|} m(A)$ . Note that  $Bet_F \in \mathcal{P}_{Bel}$ . Belief function theory is developed not only for finite sets X. For example, the real axis (the so-called continuous belief structures) [21] or fuzzy focal elements [22] are studied.

Various special cases of belief functions and the corresponding bodies of evidence are considered in applications. They reflect simple and widespread structures of propositions on the belonging of a true alternative to a certain set. In particular, a belief function (and the corresponding body of evidence) is said to be:

- *categorical* if it has only one focal element; the corresponding body of evidence is the simplest one and will be denoted by  $F_A = (A, 1)$ ;

- vacuous if the entire set X is the only focal element of this function,  $F_X = (X, 1)$ ; such a body of evidence carries no information about the belonging of the true alternative to any subset of the set X (in this case, Bel(A) = 0 and  $Pl(A) = 1 \quad \forall A \neq \emptyset, X$ );

- consonant if its focal elements are nested, i.e.,  $\forall A, B \in \mathcal{A} : A \subseteq B$  or  $B \subseteq A$ ; the corresponding body of evidence is "refining";

- simple if the BBA has no more than two focal elements and, in the case of two focal elements, X is one of them; this body of evidence consists of the simplest meaningful proposition " $x \in A$ " with a certain degree of belief m(A) and a meaningless proposition with a degree of belief 1-m(A);

- *dogmatic* if  $X \notin \mathcal{A}$  (the body of evidence does not contain a meaningless proposition: m(X) = 0).

Any body of evidence  $F = (\mathcal{A}, m)$  can be represented as  $F = \sum_{A \in \mathcal{A}} m(A) F_A$ , i.e., as a convex combination of categorical bodies of evidence. In particular, the body of evidence from Example 1 can be written in the form  $F = 0.2F_{\{b\}} + 0.1F_{\{c\}} + 0.3F_{\{a,b\}} + 0.4F_{\{b,c\}}$ . A simple body of evidence can be represented as  $F_A^{\omega} = (1-\omega)F_A + \omega F_X$ , where  $\omega \in [0, 1]$ . In particular,  $F_A^0 = F_A$  and  $F_A^1 = F_X$ . Conversely, a finite sum  $F = \sum_i \alpha_i F_{A_i}$ ,  $\alpha_i \ge 0 \forall i$ ,  $\sum_i \alpha_i = 1$ , defines a certain body of evidence.

The amount of ignorance in the information contained in a body of evidence F can be estimated using the so-called imprecision indices [23]: the more focal elements of high power and with a greater mass a body of evidence has, the greater value these indices will take. An example of such an index is the generalized Hartley measure [24, 25] H(F) = $\sum_{A \in \mathcal{A}} m(A) \log_2 |A|$ . Below, we will use the normalized generalized Hartley measure  $H_0(F) =$ 



 $H(F)/\log_2 |X|, \quad H_0: \mathcal{F}(X) \to [0, 1].$  Note that  $H_0(F) = 1 \Leftrightarrow F = F_X, \quad \text{and} \quad H_0(F) = 0 \Leftrightarrow F = P,$ where  $P \in \mathcal{P}(X).$ 

Despite the popularity and high demand, belief function theory was criticized [26–28], and several studies were initiated to clarify the theory's limits of applicability, interpretability of results, etc. As demonstrated in the paper [18], many critical remarks are due to mixing two points of view on belief functions: the first considers a belief function as a generalization of a probabilistic measure, and the second as a way of representing evidence.

#### 2. COMBINING RULES FOR BODIES OF EVIDENCE AND THE CONCEPT OF CONFLICT

A convenient tool in the theory of evidence is the ability to combine bodies of evidence, i.e., to aggregate information obtained from different sources. A combining rule is understood as a certain operation  $\otimes: \mathcal{F}(X) \times \mathcal{F}(X) \rightarrow \mathcal{F}(X)$ . There are several general schemes for constructing such rules. The most widespread one is the so-called generalized conjunctive combining rule  $\otimes_{\cap}$  with the following scheme [29]:  $F = F_1 \otimes_{\cap} F_2$ ,  $F = (\mathcal{A}, m_{\cap})$ ,  $F_1 = (\mathcal{A}_1, m_1)$  and  $F_2 = = (\mathcal{A}_2, m_2)$ , where

$$m_{\cap}(A) = \sum_{B \cap C = A} \tilde{m}(B, C),$$

and the values of the set function  $\tilde{m}: 2^{x} \times 2^{x} \to [0, 1]$ satisfy the matching conditions with the bodies of evidence  $F_1$  and  $F_2$ :

$$\sum_{\substack{C \in 2^{X} \\ B \in 2^{X}}} \tilde{m}(B, C) = m_{1}(B),$$

$$\sum_{\substack{B \in 2^{X} \\ B, C \in 2^{X}}} \tilde{m}(B, C) = m_{2}(C),$$
(1)

This system may have a set of solutions, yielding a set of new bodies of evidence  $F = F_1 \otimes_{\bigcirc} F_2$ . The latter set will be denoted by  $\mathcal{R}_{\bigcirc}(F_1, F_2)$ .

**Example 2.** Let  $X = \{x_1, x_2\}$ ,  $F_1 = 0.8F_{\{x_1\}} + 0.2F_X$ , and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$ . To find the set  $\mathcal{R}_{\cap}(F_1, F_2)$ , we solve the system

$$\sum_{C} \tilde{m}(\{x_1\}, C) = 0.8, \sum_{C} \tilde{m}(X, C) = 0.2,$$
  
$$\sum_{B} \tilde{m}(B, \{x_1\}) = 0.7, \sum_{B} \tilde{m}(B, \{x_2\}) = 0.3,$$
  
$$\tilde{m}(B, C) \in [0, 1] \quad \forall B, C \in 2^X.$$

Since  $\tilde{m}(\cdot, \emptyset) = \tilde{m}(\emptyset, \cdot) = \tilde{m}(\cdot, X) = \tilde{m}(\{x_2\}, \cdot) = 0$ , we have a system of four equations and inequalities in four variables. The solution is the numbers  $\tilde{m}(\{x_1\}, \{x_1\}) = 0.5 + t$ ,  $\tilde{m}(\{x_1\}, \{x_2\}) = 0.3 - t$ ,  $\tilde{m}(X, \{x_1\}) = 0.2 - t$ ,

and  $\tilde{m}(X, \{x_2\}) = t$ ,  $t \in [0, 0.2]$ . Then  $\mathcal{R}_{\frown}(F_1, F_2) = \{F_1 \otimes_{\frown} F_2 = (0.3 - t)F_{\oslash} + 0.7F_{\{x_1\}} + tF_{\{x_2\}} : t \in [0, 0.2]\}$ .

In particular, if the information sources are independent, then  $\tilde{m}(B, C) = m_1(B)m_2(C) \quad \forall B, C \in 2^X$ , and we obtain the so-called unnormalized Dempster rule  $\otimes_{ND}$ :  $m_{ND}(A) = \sum_{B \cap C = A} m_1(B) m_2(C) \quad \forall A \in 2^X$ . In this case, it may happen that  $K = K(F_1, F_2) = m_{ND}(\emptyset) =$  $\sum_{B \cap C = \emptyset} m_1(B) m_2(C) > 0.$  The value  $K \in [0, 1]$  is called the canonical measure of conflict. It characterizes the degree of conflict of information sources described by the bodies of evidence  $F_1$  and  $F_2$ : the greater this value is, the more inconsistent information the sources will provide. Uniformly distributing the amount of conflict over all focal elements of the new body of evidence, we get the classical Dempster rule  $\otimes_D : m_D(A) = \frac{m_{ND}(A)}{1-K} \quad \forall A \in 2^X \setminus \emptyset$  [1]. If the information sources have complete conflict (  $B \cap C = \emptyset$ for all  $B \in \mathcal{A}$  and  $C \in \mathcal{A}$ ), the Dempster combining rule becomes inapplicable: K = 1. Relating the amount of conflict to the mass of the entire set  $m_{ND}(X)$  (multiplying the mass of the meaningless proposition  $x \in X$  by the amount of conflict), we get Yager rule  $\bigotimes_{Y} : m_{Y}(A) = m_{ND}(A)$ the  $\forall A \in$ 

 $2^X \setminus \{\emptyset, X\}, m_Y(X) = m_{ND}(X) + K$  [30].

The combining operation  $\otimes_{ND}$  is associative and hence can be used to combine any finite number of bodies of evidence. Information on several other combining rules was provided in the report [31].

In some sense, the disjunctive consensus rule  $\otimes_{\cup}$  [32, 33] is dual to the Dempster rule:

$$m_{\cup}(A) = \sum_{B \cup C = A} m_1(B) m_2(C), \ A \in 2^X$$

(The body of evidence obtained using this rule is denoted by  $F = F_1 \otimes_{\cup} F_2 = (\mathcal{A}, m_{\cup})$ .) The conjunctive and disjunctive combining rules satisfy an analog of de Morgan's law [32]:

$$\overline{F_1 \otimes_{ND} F} = \overline{F}_1 \otimes_{\cup} \overline{F}_2,$$

where the complement  $\overline{F} = (\overline{\mathcal{A}}, \overline{m})$  of a body of evidence  $F = (\mathcal{A}, m)$  is defined by  $\overline{\mathcal{A}} = \{A^c : A \in \mathcal{A}\}$ and  $\overline{m}(A) = m(A^c) \quad \forall A \in \overline{\mathcal{A}}$ .

**Example 3.** Suppose that two independent expert groups predict the prospects of developing three technologies  $X = \{a, b, c\}$ . Information from the first group is described by the body of evidence  $F_1 = 0.2F_{\{b\}} + 0.1F_{\{c\}} + 0.3F_{\{a,b\}} + 0.4F_{\{b,c\}}$  (Example 1). Information from the second group is described by the body of evidence

 $F_2 = 0.7F_{\{a,c\}} + 0.3F_{\{a,b,c\}}$ . The results of combining these bodies of evidence using different conjunctive rules and the disjunctive consensus rule are presented in Table 1. (Only the masses of focal elements are indicated there.)

For these bodies of evidence, the canonical measure of conflict is  $K = m_{DN}(\emptyset) = 0.14$ . The last row of Table 1 contains the values of the imprecision index  $H_0$  for each combination result. ♦

Remark 1. A conjunctive rule is optimistic in the following sense. Suppose there are two independent information sources. According to the first rule, the true alternative belongs to a set A (a categorical body of evidence  $F_A$ ); according to the second one, to a set B (a categorical body of evidence  $F_B$ ). After the conjunctive combination of these bodies, we establish that the true alternative belongs to the set  $A \cap B$ . Applying the disjunctive consensus rule, we obtain that the true alternative belongs to the set  $A \cup B$ ; for details, see the paper [34]. In this sense, the rule under consideration is pessimistic.

The so-called matrix of evidence  $R_F^{\otimes} = \left\{ r_F^{\otimes}(A,B) \right\}_{A,B \in 2^X}$  with the elements  $r_F^{\otimes}(A,B) =$  $(F \otimes F_B)(A)$ , where  $F_B$  is a categorical body of evidence, can be assigned to each combining rule and a fixed body of evidence  $F = (\mathcal{A}, m)$ ; see [35, 36]. In particular, the matrix representations  $R_F^{\wedge}$  and  $R_F^{\cup}$  for the unnormalized Dempster rule  $\otimes_{ND}$  and the disjunctive consensus rule  $\otimes_{\cup}$ , respectively, are widespread: due to the remark above, they set completely information about the body of evidence F itself (since  $F \otimes_{ND} F_X = F$  and  $F \otimes_{\cup} F_{\varnothing} = F$ ) and about the combination results with other categorical bodies of evidence. For different categorical bodies of evidence, these results are written in the matrix  $R_{E}^{\otimes}$  column-by-

The results of combining bodies of evidence

column. The matrices  $R_F^{\cap}$  and  $R_F^{\cup}$  are called the specialization and generalization matrices of the body of evidence F, respectively, because they contain information about all bodies of evidence of the form  $F_B^{\cap} = (\mathcal{A}_B^{\cap}, m)$  and  $F_B^{\cup} = (\mathcal{A}_B^{\cup}, m)$ , respectively, where  $\mathcal{A}_{B}^{\cap} = \{A \cap B : A \in \mathcal{A}\}$  and  $\mathcal{A}_{B}^{\cup} = \{A \cup B : A \in \mathcal{A}\}$  $A \in \mathcal{A}$ }. Obviously,  $R_{F_1 \otimes_{ND} F_2}^{\cap} = R_{F_1}^{\cap} \cdot R_{F_2}^{\cap}$ . The papers [35] and [37] considered matrix operators corresponding to certain parametric families of conjunctive and disjunctive rules (the so-called  $\alpha$ -junctions). For  $\alpha = 1$ , they turn into the operators  $R_F^{\cap}$  and  $R_F^{\cup}$ .

Example 4. Consider the bodies of evidence  $F_1 = 0.8F_{\{x_1\}} + 0.2F_x$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set  $X = \{x_1, x_2\}$  (Example 2). For these bodies, we obtain:

$$R_{F_1}^{\cap} = \begin{pmatrix} 1 & 0 & 0.8 & 0 \\ 0 & 1 & 0 & 0.8 \\ 0 & 0 & 0.2 & 0 \\ 0 & 0 & 0 & 0.2 \end{pmatrix}, R_{F_2}^{\cap} = \begin{pmatrix} 1 & 0.3 & 0.7 & 0 \\ 0 & 0.7 & 0 & 0.7 \\ 0 & 0 & 0.3 & 0.3 \\ 0 & 0 & 0 & 0 \end{pmatrix},$$
$$R_{F_1}^{\cup} = \begin{pmatrix} 0 & 0 & 0 & 0 \\ 0.8 & 0.8 & 0 & 0 \\ 0.8 & 0.8 & 0 & 0 \\ 0.2 & 0.2 & 1 & 1 \end{pmatrix}, R_{F_2}^{\cup} = \begin{pmatrix} 0 & 0 & 0 & 0 \\ 0.7 & 0.7 & 0 & 0 \\ 0.3 & 0 & 0.3 & 0 \\ 0 & 0.3 & 0.7 & 1 \end{pmatrix}.$$

(The rows and columns correspond to the subsets  $\emptyset$ ,  $\{x_1\}, \{x_2\}, \text{ and } X .\} \blacklozenge$ 

#### 3. THE CONCEPT OF CONFLICT AND NON-CONFLICT **BODIES OF EVIDENCE**

In the general case, the measure of external conunderstood а flict is as certain functional

Table 1

 $Con_{ext}: \mathcal{F}(X) \times \mathcal{F}(X) \rightarrow [0, 1].$ Each body of evidence is determined by a set of focal elements and a mass function on these sets. Hence, the conflict between two bodies of evidence depends on the mutual arrangement of their sets of focal elements and the values of their mass functions.

First of all, the measure of conflict should reflect the extreme situations of conflict: be maximum in case of complete conflict of two bodies of evidence and minimum in case of no conflict.

Α	$F_1$	$F_2$	$F_1 \otimes_{_{N\!D}} F_2$	$F_1 \otimes_D F_2$	$F_1 \otimes_Y F_2$	$F_1 \otimes_{\cup} F_1$
Ø	_	_	0.14	_		_
$\{a\}$	-	_	0.21	0.244	0.21	_
$\{b\}$	0.2	-	0.06	0.07	0.06	-
$\{c\}$	0.1	_	0.38	0.442	0.38	_
$\{a, b\}$	0.3	_	0.09	0.104	0.09	_
$\{a, c\}$	_	0.7	_	_	_	0.07
$\{b, c\}$	0.4	-	0.12	0.14	0.12	_
${a, b, c}$	_	0.3	_	_	0.14	0.93
$H_{0}$	0.442	0.742	0.132	0.154	0.272	0.974
	•	•	•	•	•	•

A natural condition for complete conflict is that all pairs of focal elements do not intersect for two bodies of evidence. More precisely, bodies of evidence  $F_1 = (\mathcal{A}_1, m_1)$  and  $F_2 = (\mathcal{A}_2, m_2)$  are said to be in complete conflict if  $A \cap B = \emptyset \quad \forall A \in \mathcal{A}_1, \forall B \in \mathcal{A}_2$ .

However, different degrees of non-conflict are possible. For example, the paper [38] considered the following degrees of non-conflict for bodies of evidence  $F_1 = (\mathcal{A}, m_1)$  and  $F_2 = (\mathcal{A}_2, m_2)$ :

- strong non-conflict:  $\bigcap_{A \in \mathcal{A} \cup \mathcal{A}} A \neq \emptyset$ ,

- (simple) non-conflict:  $A \cap B \neq \emptyset$   $\forall A \in \mathcal{A}_1,$  $\forall B \in \mathcal{A}_2,$ 

- weak non-conflict:  $\mathcal{P}_{Bel_1} \cap \mathcal{P}_{Bel_2} \neq \emptyset$ , where  $Bel_i$  is a belief function corresponding to a body of evidence  $F_i$ , i = 1, 2.

As is known [38], strong non-conflict implies (simple) non-conflict, and non-conflict implies weak non-conflict. For categorical bodies of evidence  $F_A$ and  $F_B$ , all these concepts of non-conflict reduce to the non-empty intersection of their focal elements:  $A \cap B \neq \emptyset$ . However, e.g., if  $F_1 = F_2 = \alpha F_A +$  $(1-\alpha)F_B$ , where  $A \cap B = \emptyset$  and  $\alpha \in (0,1)$ , then such identical bodies of evidence are in weak non-conflict only: they will be neither (simply) non-conflict nor strongly non-conflict.

Due to the conditions of complete conflict and nonconflict, the measure of external conflict  $Con_{ext}$  should satisfy the following conditions [4, 38, 39]:

E1:  $Con_{ext}(F_1, F_2) = Con_{ext}(F_2, F_1) \ \forall F_1, F_2 \in \mathcal{F}(X)$ (symmetry).

**E2**:  $Con_{ext}(F_1, F_2) = 0$  if  $F_1$  and  $F_2$  are weakly non-conflict.

In particular, any body of evidence F will be weakly non-conflict with itself and the meaningless body of evidence  $F_x$ . Therefore, under condition E2, we have the following corollaries:

a)  $Con_{ext}(F,F) = 0 \quad \forall F \in \mathcal{F}(X)$  (nilpotency).

b)  $Con_{ext}(F_X, F) = 0 \quad \forall F \in \mathcal{F}(X)$  (non-conflict with ignorance).

**E3**:  $Con_{ext}(F_1, F_2) = 1$  if  $A \cap B = \emptyset$   $\forall A \in \mathcal{A}_1$ ,  $\forall B \in \mathcal{A}_2$  (complete conflict).

The next condition conceptually relates to the specialization of evidence [32].

A body of evidence  $F' = (\mathcal{A}', m')$  is called a specialization of a body of evidence  $F'' = (\mathcal{A}'', m'')$  (and denoted by  $F' \sqsubseteq F''$ ) if there exists a partition  $\mathcal{A}' = \mathcal{A}'_1 \cup \ldots \cup \mathcal{A}'_k$ ,  $\mathcal{A}'_i \cap \mathcal{A}'_j = \emptyset$   $\forall i \neq j$ ,  $k = |\mathcal{A}''|$  such that  $\bigcup_{A \in \mathcal{A}'} A \subseteq B_i$  and  $\sum_{A \in \mathcal{A}'} m'(A) = m''(B_i)$ ,  $\forall B_i \in \mathcal{A}'', i = 1, ..., k$ .

In other words, a body of evidence F' refines (specializes) a body of evidence F''. The latter body is called a generalization of a body of evidence F'. Note that  $F \sqsubseteq F_X \quad \forall F \in \mathcal{F}(X)$ .

If  $F' \sqsubseteq F''$ , then the imprecision of F' does not exceed that of F'' in terms of any index (e.g.,  $H(F') \le H(F'')$  for the generalized Hartley measure).

E4:  $Con_{ext}(F', F) \ge Con_{ext}(F'', F) \quad \forall F, F', F'' \in \mathcal{F}(X), F' \sqsubseteq F''$  (antimonotonicity with respect to specialization).

Note that the canonical measure of conflict K satisfies all conditions mentioned, except E2 (particularly is not nilpotent). At the same time, it satisfies

**E2'**:  $Con_{ext}(F_1, F_2) = 0$  if  $F_1$  and  $F_2$  are (simply) non-conflict.

#### 4. AN AXIOMATICS FOR A MEASURE OF EXTERNAL CONFLICT

The paper [40] considered an axiomatics for a measure of conflict based on strengthening conditions E1 - E4:

A1:  $Con_{ext}(F_1, F_2) = Con_{ext}(F_2, F_1) \forall F_1, F_2 \in \mathcal{F}(X).$ 

A2:  $Con_{ext}(F_1, F_2) = 0 \Leftrightarrow F_1$  and  $F_2$  are weakly non-conflict.

**A3**:  $Con_{ext}(F_1, F_2) = 1 \Leftrightarrow A \cap B = \emptyset \quad \forall A \in \mathcal{A}_1, \forall B \in \mathcal{A}_2.$ 

A4:  $Con_{ext}(F', F) \ge Con_{ext}(F'', F) \quad \forall F, F', F'' \in \mathcal{F}(X), F' \sqsubseteq F''.$ 

Also, the following additional axiom was introduced therein:

A5: If  $Con_{ext}(F_1, F_2) = a \in [0, 1]$ , then  $\exists F_i^{(k)} \in \mathcal{F}(X)$ , k = 1, 2, i = 1, 2:  $F_i = (1-a)F_i^{(1)} + aF_i^{(2)}$ , i = 1, 2,  $Con_{ext}(F_1^{(1)}, F_2^{(1)}) = 0$ ,  $Con_{ext}(F_1^{(2)}, F_2^{(2)}) = 1$ . This axiom assumes that information sources can be linearly divided into completely conflict and weakly non-conflict parts proportionally to the amount of conflict estimated by a given measure. This is a rather strong requirement. In particular, the canonical measure of conflict does not satisfy this axiom. (This fact can be easily verified by an appropriate example.)

The following theorem on the measure of conflict is true.

**Theorem 1** [40]. If a functional  $Con_{ext}$ :  $\mathcal{F}(X) \times \mathcal{F}(X) \rightarrow [0, 1]$  satisfies axioms A1 – A5, it has the form

$$Con_{\text{ext}}(F_1, F_2) =$$

$$\inf \{ m_{\bigcirc}(\emptyset) : F = (\mathcal{A}, m_{\bigcirc}) \in \mathcal{R}_{\bigcirc}(F_1, F_2) \}.$$
(2)

**Example 5.** Consider the bodies of evidence  $F_1 = 0.8F_{\{x_1\}} + 0.2F_X$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set  $X = \{x_1, x_2\}$ . Applying the generalized conjunctive combining rule  $\bigotimes_{\cap}$ , we obtain  $m_{\cap}(\emptyset) = 0.3 - t$ ,  $t \in [0, 0.2]$  (Example 2). Hence,  $Con_{ext}(F_1, F_2) = \min\{0.3 - t : t \in [0, 0, 2]\} = 0.1$ . Note that for these bodies of evidence, the canonical measure of conflict is K = 0.24.

As shown in the paper [40], the measure (2) can be calculated by the formula

$$Con_{ext}(F_1, F_2) = \inf \{Con_{ext}(P_1, P_2): P_1 \in \mathcal{P}_{Bel_1}, P_2 \in \mathcal{P}_{Bel_2}\},\$$
  
where  $\mathcal{P}_{Bel_i}$  denotes the creedal sets corresponding to  
the bodies of evidence  $F_i$ ,  $i = 1, 2$ . For probability  
measures,

$$Con_{\text{ext}}(P_1, P_2) = 1 - \sum_{i=1}^{n} \min\{P_1(x_i), P_2(x_i)\}.$$
 (3)

**Example 6.** For the previous example, we have:  $\mathcal{P}_{Bel_1} = \left\{ \alpha F_{\{x_1\}} + (1-\alpha) F_{\{x_2\}} : 0.8 \le \alpha \le 1 \right\}, \quad \mathcal{P}_{Bel_2} = \{F_2\} = \{0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}\}.$  Then  $Con_{ext}(F_1, F_2) = \inf\{Con_{ext}(P_1, P_2) : P_1 \in \mathcal{P}_{Bel_1}, P_2 \in \mathcal{P}_{Bel_2}\} = \inf_{0.8 \le \alpha \le 1} (1 - \min\{\alpha; 0.7\} - \min\{1-\alpha; 0.3\}) = \inf_{0.8 \le \alpha \le 1} (1 - 0.7 - (1-\alpha)) = 0.1.$ 

**Remark 2**. For probability measures  $P_1, P_2 \in \mathcal{P}(X)$ , the canonical measure of conflict K is given by

$$K(P_1, P_2) = \sum_{i=1}^{n} P_1(x_i) \sum_{j=1, j \neq i}^{n} P_2(x_j) =$$
$$\sum_{i=1}^{n} P_1(x_i) (1 - P_2(x_i)) = 1 - \sum_{i=1}^{n} P_1(x_i) P_2(x_i)$$

(Compare with formula (3).) It follows that the canonical measure of conflict is generally not nilpotent since  $K(P, P) = 1 - \sum_{i=1}^{n} P^2(x_i), P \in \mathcal{P}(X)$ . In particular, for the uniform distribution *P* on *X*, we obtain  $K(P, P) = 1 - \frac{1}{|X|}$ .

The paper [4] considered an axiomatics for the measure of external conflict *Con* on an arbitrary

measure of external conflict  $Con_{ext}$  on an arbitrary finite set of bodies of evidence  $M = \{F_1, ..., F_l\}$ ,  $F_i \in \mathcal{F}(X), i = 1, ..., l$ . Let  $2^M$  be the set of all subsets from the set M. By definition,  $Con_{ext}(B) = 0$  if  $|B| = 1, B \in 2^M$ , and  $Con_{ext}(\emptyset) = 0$ . In addition to direct generalizations of axioms A1 - A4 to the set case, the monotonicity axiom was introduced:

A6:  $Con_{ext}(B) \leq Con_{ext}(C)$  if  $B \subseteq C$  and  $B, C \in 2^{M}$ .

For example, due to associativity, the canonical measure of conflict can be extended to an arbitrary finite set M of bodies of evidence. As is easily demonstrated, the canonical measure of conflict will satisfy the monotonicity axiom A6 on this set.

#### 5. METHODS TO ESTIMATE EXTERNAL CONFLICT

There are several approaches to estimating the conflict of bodies of evidence. They can be conventionally divided into metric, structural, algebraic, and combining rule-based ones.

#### 5.1. Metric methods to estimate conflict

The metric approach is one of the most popular methods to estimate conflict [41, 42]. Metrics can be introduced on the bodies of evidence themselves and their set functions or matrix representations. The definition of metrics on the set  $\mathcal{F}(X)$  of bodies of evidence considers, to one degree or another, structural features of bodies of evidence: the power of the focal elements, the degree of intersection for the focal elements of two bodies of evidence, the mutual arrangement of the focal elements, and others. For measuring conflict, a metric *d* must be normalized so that  $d: \mathcal{F}(X) \times \mathcal{F}(X) \rightarrow [0, 1]$ .

Metric between bodies of evidence based on the generalized Euclidean distance. This metric is given by

$$\frac{d_{\mathcal{Q}}(F_1, F_2) = \left\| \mathbf{m}_1 - \mathbf{m}_2 \right\|_{\mathcal{Q}}}{\sqrt{\frac{1}{2} (\mathbf{m}_1 - \mathbf{m}_2)^{\mathsf{T}} \mathcal{Q}(\mathbf{m}_1 - \mathbf{m}_2)}},$$
(4)

where  $\mathbf{m}_i$  denotes a  $2^{|X|}$ -dimensional column vector composed of the mass function values  $m_i(A)$ ,  $A \in 2^X$ , i=1, 2, for some ordering of all subsets from  $2^X$ ;  $Q = (q_{A,B})_{A,B \in 2^X}$  is the matrix of similarity measures between the focal elements with the entries  $q_{A,B} = q_{B,A} \in [0,1]$  such that  $q_{A,A} = 1$  for  $A \in 2^X \setminus \emptyset$ , and  $q_{A,B} = 0$  if  $A \cap B = \emptyset$ . Then  $d_{\varrho} \in [0,1]$ . For example, a popular metric of this type was described in [41]. As a similarity measure, it involves *the Jaccard index* [43]:  $J = (jac_{A,B})_{A,B \in 2^X}$ , where  $jac_{A,B} = \frac{|A \cap B|}{|A \cup B|}$  for  $A, B \neq \emptyset$  and  $d_{\emptyset,\emptyset} = 0$ . The positive definiteness of the matrix (the fact that  $d_J$  is a metric on the set  $\mathcal{F}(X) \times \mathcal{F}(X)$ ) was proved in the paper [44]. In particular, for categorical bodies of evidence, we have  $d_J(F_A, F_B) = \sqrt{1 - \frac{|A \cap B|}{|A \cup B|}}$ . In [45], the metric (4) with the Jaccard similarity measure was generalized to the class of continuous belief structures.

**Example 7.** Consider the bodies of evidence  $F_1 = 0.8F_{\{x_1\}} + 0.2F_X$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set  $X = \{x_1, x_2\}$  (Example 2). We have:  $J = \begin{pmatrix} 1 & 0 & 1/2 \\ 0 & 1 & 1/2 \\ 1/2 & 1/2 & 1 \end{pmatrix}$ ,  $\mathbf{m}_1 = (0.8; 0; 0.2)^{\mathrm{T}}$ ,  $\mathbf{m}_2 = (0.7; 0.3; 0)^{\mathrm{T}}$ , and  $d_J(F_1, F_2) = \frac{1}{2\sqrt{5}} \approx 0.224$ .

Among other similarity measures used in formula (4), we mention the Sørensen and Simpson coefficients:  $sor_{A,B} = \frac{2|A \cap B|}{|A|+|B|}$  and  $sim_{A,B} = \frac{|A \cap B|}{\min\{|A|,|B|\}}$  [46]. However, the corresponding matrix is not always positive definite, and the distance function is not always a (complete) metric. The paper [47] proposed a metric of the form (4) with a metric space X and the similarity measure  $q_{A,B} = \frac{1}{1+c \cdot d_H(A,B)}$ , where  $d_H(A, B)$  is the Hausdorff metric between sets A and B of the space

A and *B* of the space X and *c* > 0. This similarity measure<sup>1</sup> does not satisfy the condition  $q_{A,B} = 0$  for  $A \cap B = \emptyset$ . This metric was used in [47] to aggregate information from different sensors under uncertain measurement errors and uncertain noise characteristics.

**Example 8.** Consider the bodies of evidence  $F_1 = 0.5F_{[40,50]} + 0.3F_{[45,55]} + 0.2F_x$  and  $F_2 = 0.3F_{[30,60]} + 0.7F_{[40,50]}$  on the set X = [20,70], which predict the company's stock value. We find the distance between them using the metric (4) with  $q_{A,B} = \frac{1}{1+0.2d_H(A,B)}$ , where  $d_H$  de-

notes the Hausdorff metric. According to the body of evidence  $F_1$ , the stock value will belong to the interval [40, 50] with the degree of belief 0.5, the interval [45, 55] with the degree of belief 0.3, and the interval X = [20, 70] with the degree of belief 0.2 (in conditional units). The body of evidence  $F_2$  is interpreted by analogy. Consider the vector representations of these bodies of evidence on the sets  $\mathcal{A} \cup \mathcal{A}_2 = \{[40, 50], [45, 55], [30, 60], [20, 70]\}$  only (with the specified ordering). Then  $\mathbf{m}_1 = (0.5; 0.3; 0; 0.2)^T$  and  $\mathbf{m}_2 = (0.7; 0; 0.3; 0)^T$ . For intervals  $[a_1, b_1]$  and  $[a_2, b_2]$ , the Hausdorff measure is given by  $d_H([a_1, b_1], [a_2, b_2]) =$ 

 $\max \{ |a_1 - a_2|, |b_1 - b_2| \}.$  Therefore, the matrix  $Q = (q_{[a_i, b_i], [a_j, b_j]})_{i=1}^4$  has the form

$$Q = \begin{pmatrix} 1 & 1/2 & 1/3 & 1/5 \\ 1/2 & 1 & 1/4 & 1/6 \\ 1/3 & 1/4 & 1 & 1/3 \\ 1/5 & 1/6 & 1/3 & 1 \end{pmatrix},$$

and 
$$d_Q(F_1, F_2) = \sqrt{\frac{1}{2}(\mathbf{m}_1 - \mathbf{m}_2)^T Q(\mathbf{m}_1 - \mathbf{m}_2)} \approx 0.282.$$
  
The metric

 $d(F_1, F_2) = \frac{1}{2|X|-1} \sum_{A=2^X} |A| |m_1(A) - m_2(A)|,$ 

suggested in [11], is another one calculated directly between bodies of evidence with information uncertainty. This metric<sup>2</sup> was used therein to approximate complex bodies of evidence by those with a simpler structure of focal elements.

Metric between set functions describing bodies of evidence bijectively. Consider distance functions (metrics, semimetrics, and pseudometrics) between set functions describing bodies of evidence bijectively. Among them, we separate the Minkowski metrics  $l_p$ ,  $1 \le p \le \infty$ , between belief functions or plausibility functions:

$$d_{p}(F_{1}, F_{2}) = c \left( \sum_{A \in 2^{X}} |Pl_{1}(A) - Pl_{2}(A)|^{p} \right)^{\frac{1}{p}} = c \left( \sum_{A \in 2^{X}} |Bel_{1}(A) - Bel_{2}(A)|^{p} \right)^{\frac{1}{p}}.$$

(Since  $|Pl_1(A) - Pl_2(A)| = |Bel_1(A^c) - Bel_2(A^c)| \quad \forall A \in 2^x$ , these metrics coincide; see the papers [11, 48].) For  $p = \infty$ , the sum is replaced by the maximum over all subsets  $A \in 2^x$ . Below, the normalizing factor c > 0 is chosen so that  $d \in [0, 1]$ . Such metrics are often used to approximate bodies of evidence.

Metrics between set functions describing bodies of evidence not bijectively. One example is the Minkowski metric  $l_p$ ,  $1 \le p \le \infty$ , between pignistic probabilities [49, 50]

$$d_{Bet,p}(F_1, F_2) = c \left( \sum_{A \in 2^X} \left| Bet_{F_1}(A) - Bet_{F_2}(A) \right|^p \right)^{\frac{1}{p}},$$

between the probabilities of basic set elements

$$d_{Bet,p,x}(F_1, F_2) = c \left( \sum_{x \in X} \left| Bet_{F_1}(x) - Bet_{F_2}(x) \right|^p \right)^{\frac{1}{p}},$$

<sup>&</sup>lt;sup>1</sup> The positive definiteness of the similarity matrix was established in [47] for a particular case.

 $<sup>^{2}</sup>$  The paper [11] considered the unnormalized metric.

or between the so-called contour functions of bodies of evidence [51]

$$d_{Pl,p,x}(F_1, F_2) = c \left( \sum_{x \in X} |Pl_1(x) - Pl_2(x)|^p \right)^{\frac{1}{p}}$$

However, these functions are only pseudometrics on  $\mathcal{F}(X) \times \mathcal{F}(X)$ : they satisfy the reflexivity axiom d(F, F) = 0 instead of the coincidence axiom  $d(F_1, F_2) = 0 \Leftrightarrow F_1 = F_2$ .

In particular, the pseudometric  $d_{Bet,1,x}$  was used in [14] to train a classifier with classes described by belief functions.

Normalized semimetrics based on the scalar product of vector representations of evidence bodies. (A semimetric satisfies all axioms of a metric, except for the triangle inequality.) An example of such a semimetric is the function introduced in [42]:

$$d(F_1, F_2) = 1 - \frac{\mathbf{Pl}_1^T \cdot \mathbf{Pl}_2}{\|\mathbf{Pl}_1\| \|\mathbf{Pl}_2\|},$$

where  $\mathbf{Pl}_i$  is a  $2^{|X|}$ -dimensional column vector composed of the plausibility function values  $Pl_i(A)$ ,  $A \in 2^X$ , i = 1, 2;  $\|\cdot\|$  denotes the Euclidean norm.

**Example 9.** Consider the bodies of evidence  $F_1 = 0.8F_{\{x_1\}} + 0.2F_x$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set  $X = \{x_1, x_2\}$  (Example 2). We have:  $\mathbf{Pl}_1 = (1; 0.2; 1)^T$ ,  $\mathbf{Pl}_2 = (0.7; 0.3; 0)^T$ , and  $d(F_1, F_2) = 1 - \frac{\mathbf{Pl}_1^T \cdot \mathbf{Pl}_2}{\|\mathbf{Pl}_1\| \|\mathbf{Pl}_2\|} \approx 0.301$ .

Metrics based on comparing uncertainty intervals  $\{[Bel_1(A), Pl_1(A)]\}_{A \in 2^X}$  and  $\{[Bel_2(A), Pl_2(A)]\}_{A \in 2^X}$  that describe bodies of evidence  $F_1$  and  $F_2$  bijectively. In this case, a metric on the set of bodies of evidence can be defined by extending the metric on uncertainty intervals:

$$d_{I,p}(F_{1}, F_{2}) = c \times \left(\sum_{A \in 2^{X}} d_{I}^{p} \left( [Bel_{1}(A), Pl_{1}(A)], [Bel_{2}(A), Pl_{2}(A)] \right) \right)^{\frac{1}{p}}, (5)$$

$$1 \le p \le \infty,$$

where  $d_i$  denotes the metric on the intervals. For example, the Hausdorff metric  $d_H$  can be used as  $d_i$ ; see Example 8. The paper [12] considered a metric of the form (5) with  $d_i = d_{W,p}([a_1, b_1], [a_2, b_2])$  (the Wasserstein *p*-metric of two uniform distributions on intervals  $[a_1, b_1]$  and  $[a_2, b_2]$ ). This metric was applied to approximate bodies of evidence. In particular,  $d_{W,2}([a_1, b_1], [a_2, b_2]) = \sqrt{(mean_1 - mean_2)^2 + \frac{1}{3}(rad_1 - rad_2)^2}$ , where *mean<sub>i</sub>* and *rad<sub>i</sub>* are the middle and half-length of an interval  $[a_i, b_i]$ , i = 1, 2. As shown in [12], for p = 2, the normalizing factor is  $c = 2^{-\frac{|X|+1}{2}}$ . This metric estimates the deviation between the information uncertainties of bodies of evidence.

**Example 10.** Consider the bodies of evidence  $F_1 = 0.8F_{\{x_1\}} + 0.2F_X$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set  $X = \{x_1, x_2\}$  (Example 2). We find the values of the metrics  $d_{W,2}(F_1, F_2)$  and  $d_{H,2}(F_1, F_2)$ . For |X| = 2 and p = 2, the normalizing factor is  $c = \frac{1}{\sqrt{2}}$ . The calculation results are presented in Table 2.

Then 
$$d_{W,2}(F_1, F_2) = \frac{1}{10}\sqrt{\frac{13}{3}} \approx 0.208$$
 and  $d_{H,2}(F_1, F_2) = 0.3$ .

**The Wasserstein metric**. According to the paper [52], a metric *d* defined on  $2^x$  can be extended to the set  $\mathcal{F}(X)$  by solving the Kantorovich problem

$$d_W(F_1,F_2) = \min_{\tilde{m}} \sum_{A \in 2^X} \sum_{A \in 2^X} \tilde{m}(A,B) d(A,B).$$

Here, the minimum is taken over all set functions satisfying the matching conditions (1). The result can be treated as a Wasserstein 1-metric between bodies of evidence. This metric has a good interpretation as a solution of the Kantorovich optimization problem.

**Example 11.** Consider the bodies of evidence  $F_1 = 0.8F_{\{x_1\}} + 0.2F_X$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set  $X = \{x_1, x_2\}$  (Example 2). We have:  $\tilde{m}(\{x_1\}, \{x_1\}) = 0.5 + t$ ,  $\tilde{m}(\{x_1\}, \{x_2\}) = 0.3 - t$ ,  $\tilde{m}(X, \{x_1\}) = 0.2 - t$ ,  $\tilde{m}(X, \{x_2\}) = t$ ,  $t \in [0, 0.2]$ , and  $\tilde{m}(A, B) = 0$  for all other pairs

Table 2

The values of the metrics  $d_{\mu,2}$  and  $d_{H,2}$  between the uncertainty intervals [Bel<sub>i</sub>(A), Pl<sub>i</sub>(A)], i = 1, 2

4		$F_1$				$F_2$			$d^2$	$d^2$
A	$Bel_1$	$Pl_1$	mean <sub>1</sub>	$rad_1$	Bel <sub>2</sub>	$Pl_2$	mean <sub>2</sub>	$rad_2$	$a_{W,2}$	$a_{H,2}$
${x_1}$	0.8	1	0.9	0.1	0.7	0.7	0.7	0	13/300	0.09
$\{x_2\}$	0	0.2	0.1	0.1	0.3	0.3	0.3	0	13/300	0.09
X	1	1	1	0	1	1	1	0	0	_



 $(A, B) \in 2^{X} \times 2^{X}$ . On the set  $2^{X}$ , let  $d(A, B) = d_{J}(F_{A}, F_{B}) = \sqrt{1 - \frac{|A \cap B|}{|A \cup B|}}$ . Then the Wasserstein metric between the bodies of evidence  $F_{1}$  and  $F_{2}$  is given by

$$d_{W}(F_{1}, F_{2}) = \min_{0 \le t \le 0.2} \left\{ (0.3 - t) \cdot 1 + (0.2 - t) \frac{1}{\sqrt{2}} + t \frac{1}{\sqrt{2}} \right\} = \min_{0 \le t \le 0.2} \left\{ \frac{3 + \sqrt{2}}{10} - t \right\} = \frac{1 + \sqrt{2}}{10} \approx 0.241. \blacklozenge$$

Metrics between the specialization (generalization) matrices of bodies of evidence. The paper [53] introduced and examined the following metrics on the set  $\mathcal{F}(X)$ :

$$d^{\frown}(F_1, F_2) = c_{\frown} \left\| R_{F_1}^{\frown} - R_{F_2}^{\frown} \right\|,$$
  
$$d^{\cup}(F_1, F_2) = c_{\cup} \left\| R_{F_1}^{\cup} - R_{F_2}^{\cup} \right\|,$$

where  $R_{F_i}^{\cap}$  and  $R_{F_i}^{\cup}$  are the specialization and generalization matrices, respectively, of a body of evidence  $F_i$ , i = 1, 2 (Section 3), and  $\| \|$  denotes the matrix norm. Clearly,  $c_{\cap} = \left( \max_{A,B \in 2^X} \| R_{F_A}^{\cap} - R_{F_B}^{\cap} \| \right)^{-1}$  and  $c_{\cup} = \left( \max_{A,B \in 2^X} \| R_{F_A}^{\cup} - R_{F_B}^{\cup} \| \right)^{-1}$ . The generalizations of these metrics for the operator representations of  $\alpha$ -junctions were considered in the paper [54]. It was demonstrated therein that  $d^{\cap} = d^{\cup}$ . Metrics based on specialization

matrices have higher robustness to small changes in bodies of evidence. **Example 12.** For the bodies of evidence  $F_1 = 0.8 \times F_{\{x_1\}} + 0.2F_X$  and  $F_2 = 0.7F_{\{x_1\}} + 0.3F_{\{x_2\}}$  on the set

 $X = \{x_1, x_2\}$ , the specialization matrices  $R_{F_1}^{\cap}$  and  $R_{F_2}^{\cap}$  have been found in Example 4. For example, for the matrix norm

$$\|A\|_{1} = \max_{j} \sum_{i} |a_{ij}|, \text{ we obtain } c_{\cap} = \frac{1}{2} \text{ and}$$
$$d_{1}^{\cap}(F_{1}, F_{2}) = \frac{1}{2} \|R_{F_{1}}^{\cap} - R_{F_{2}}^{\cap}\|_{1} =$$
$$\frac{1}{2} \| \begin{pmatrix} 0 & -0.3 & 0.1 & 0\\ 0 & 0.3 & 0 & 0.1\\ 0 & 0 & -0.1 & -0.3\\ 0 & 0 & 0 & 0.2 \end{pmatrix} \|_{1} = 0.3. \blacklozenge$$

As noted in several publications, using only one metric or functional to characterize external conflict is not enough. In particular, the canonical measure of conflict will take large values for a pair of identical (weakly non-conflict) bodies of evidence "close" to the equiprobable distribution (Remark 2), for which the metric component will be equal to zero. On the other hand, the conflict between bodies of evidence is often not reduced to calculating the metric component only. Indeed, categorical consonant bodies of evidence  $F_A$  and  $F_B$ , where  $A \subseteq B$ , are strongly non-conflict, and  $K(F_A, F_B) = 0$ . However, the metric (4) with the Jacquard index for these bodies of evidence, given by  $d_J(F_A, F_B) = \sqrt{1 - \frac{|A|}{|B|}}$ , will take large values if  $|A| \ll |B|$ .

Therefore, some researchers proposed using a set of conflict measures and metrics on the set of evidence bodies  $\mathcal{F}(X)$ . For example, the paper [50] considered the pair  $(K(F_1, F_2), d_{Bet,\infty}(F_1, F_2))$ , where  $d_{Bet,\infty}(F_1, F_2) = \max_{A \subseteq X} |Bet_{F_1}(A) - Bet_{F_2}(A)|$ . Large values of each component in this pair guarantee a large conflict between the bodies of evidence. In particular, a large value of the metric  $d_{Bet,\infty}(F_1, F_2)$  indicates that the bodies of evidence are far from probability distributions.

In the general case, the canonical measure of conflict can be represented as [55]

where

$$E_{Q}(\mathbf{m}_{i}) = \frac{1}{2} - \|\mathbf{m}_{i}\|_{Q}^{2} = \frac{1}{2} \sum_{B} m_{i}(B) \left(1 - \sum_{C} q_{B,C} m_{i}(C)\right),$$
  
 $i = 1, 2, \text{ and } r_{B,C} = \begin{cases} 0, B \cap C = \emptyset \lor B = C, \\ 1 - q_{B,C}, \text{ otherwise.} \end{cases}$ 

The last term in formula (6) characterizes the interaction of weakly intersecting focal elements. The functional  $E_Q: \mathcal{F}(X) \rightarrow [0, \frac{1}{2}]$  is close by properties to the entropy functional and describes the amount of internal conflict in a body of evidence. In particular, for completely conflict bodies of evidence,  $K(F_1, F_2) = E_I(\mathbf{m}_1) + E_I(\mathbf{m}_2) + \|\mathbf{m}_1 - \mathbf{m}_2\|_I^2$ , where *I* denotes an identity matrix of compatible dimensions. The use of entropies and divergences for estimating the uncertainty of bodies of evidence was considered in [56].

#### 5.2. Structural methods to estimate conflict

The paper [39] presented a structural approach to estimating conflict considering the degree to which the focal elements of one body of evidence are included in the focal elements of another body of evidence. The set inclusion index  $Inc(A, B) = \begin{cases} 1, A \subseteq B, \\ 0, A \notin B, \end{cases}$   $A, B \in 2^{x}$ ,



was adopted to calculate the degree of inclusion of one set of focal elements into another set:

$$\delta(\mathcal{A}_1, \mathcal{A}_2) = \max\{d(\mathcal{A}_1, \mathcal{A}_2), d(\mathcal{A}_2, \mathcal{A}_1)\},\$$
  
$$d(\mathcal{A}_1, \mathcal{A}_2) = \frac{1}{|\mathcal{A}||\mathcal{A}_2|} \sum_{A_1 \in \mathcal{A}} \sum_{A_2 \in \mathcal{A}} Inc(A_1, A_2).$$

The measure of conflict was defined by the formula

$$Con_{ext}(F_1, F_2) = (1 - \delta(\mathcal{A}_1, \mathcal{A}_2)) d_J(F_1, F_2),$$

where  $d_j$  is the distance (4) with the Jaccard index. This measure satisfies conditions E1, E2', and E3, as well as corollaries a) and b).

**Remark 3.** The set inclusion index Inc(A,B) can be treated as a value of a belief measure  $\eta_A$  corresponding to a categorical body of evidence  $F_A$  on a set *B*. Then an arbitrary belief function *Bel* corresponding to a body of evidence  $F = (\mathcal{A}, m)$  can be represented as  $Bel = \sum_{A \in \mathcal{A}} m(A)\eta_A$ .

#### 5.3. Algebraic methods to estimate conflict

The paper [57] introduced an algebraic approach to measuring conflict based on set similarity measures:  $Q = (q_{A,B})$ , where  $q_{A,B} = q_{A,B} \in [0, 1]$ ,  $q_{A,A} = 1$  for  $A \in 2^X \setminus \emptyset$ , and  $q_{A,B} = 0$  if  $A \cap B = \emptyset$ .

A functional  $Con_Q: \mathcal{F}(X) \times \mathcal{F}(X) \rightarrow [0,1]$  is said to be bilinear if  $Con_Q(\alpha F_1 + \beta F_2, F) =$  $\alpha Con_Q(F_1, F) + \beta Con_Q(F_2, F)$  for all  $\alpha, \beta \in [0, 1]$ ,  $\alpha + \beta = 1$ , and  $F, F_1, F_2 \in \mathcal{F}(X)$ . Let us weaken condition E4 (antimonotonicity with respect to specialization) as follows:

E4': If  $F = (\mathcal{A}_0, m)$ ,  $F' = (\mathcal{A} \cup \{A'\}, m')$ ,  $F'' = (\mathcal{A} \cup \{A'\}, m')$ ,  $m'(A) = m''(A) \quad \forall A \in \mathcal{A}, m'(A') = m''(A'')$ , and  $q_{A',B} \leq q_{A'',B} \quad \forall B \in \mathcal{A}_0$ , then  $Con_o(F', F) \geq Con_o(F'', F)$ .

**Theorem 2** [57]. *The functional*  $Con_{\varrho}$  *is a bilinear measure of conflict on the set*  $\mathcal{F}(X) \times \mathcal{F}(X)$  *that satisfies conditions* E1, E3, *and* E4' *if and only if* 

$$Con_{Q}(F_{1}, F_{2}) = \sum_{A \in \mathcal{A}, B \in \mathcal{A}} \gamma(A, B)m_{1}(A)m_{2}(B) = K(F_{1}, F_{2}) + \sum_{A \cap B \neq \emptyset} \gamma(A, B)m_{1}(A)m_{2}(B),$$

$$(7)$$

where the coefficients  $\gamma(A, B) = Con_Q(F_A, F_B) \in [0, 1]$ have the following properties:  $\gamma(A, B) = \gamma(B, A)$ ,  $\gamma(A', B) \ge \gamma(A'', B)$  if  $q_{A',B} \le q_{A'',B}$ , and  $\gamma(A, B) = 1$  if  $A \cap B = \emptyset$ . Due to (7), the canonical measure of conflict  $K(F_1, F_2)$  is least among all bilinear measures of conflict:  $Con_Q(F_1, F_2) \ge K(F_1, F_2)$ .

**Remark 4.** The coefficients  $\gamma(A, B) = Con_Q(F_A, F_B) = \psi(q_{A,B})$ ,  $A, B \neq \emptyset$ , satisfy conditions a)-c) of Theorem 2 if  $\psi$  is a nonincreasing function such that  $\psi(1) = 0$ ,  $\psi(0) = 1$ , and  $q_{A,B} = |A \cap B| / \min\{|A|, |B|\}$ . In particular, if  $q_{A,B} = \begin{cases} 1, A \cap B \neq \emptyset, \\ 0, A \cap B = \emptyset \end{cases}$  is a primitive measure of intersection, then  $Con_Q(F_1, F_2) = K(F_1, F_2)$ .

#### 5.4. Combining rule-based methods to estimate conflict

For two bodies of evidence  $F_1 = (\mathcal{A}, m_1)$ and  $F_2 = (\mathcal{A}_2, m_2)$ , the canonical measure of conflict  $K(F_1, F_2)$  is equal to the mass of an empty set obtained by combining,  $F_1 \otimes_{ND} F_2 = (\mathcal{A} \land \mathcal{A}_2, m_{ND})$ ,  $\mathcal{A} \land \mathcal{A}_2 = = \{A \cap B : A \in \mathcal{A}_1, B \in \mathcal{A}_2\}$ , using the unnormalized Dempster rule:  $K(F_1, F_2) = m_{DN}(\emptyset) =$  $\sum_{B \cap C = \emptyset} m_1(B)m_2(C)$ .

The measure of conflict from Theorem 1 is equal to the infimum of the masses of all empty sets yielded by combining using the generalized conjunctive rule  $\otimes_{\bigcirc}$ .

These examples show that the measure of conflict must be agreed with combining rules for bodies of evidence. In particular, this issue was discussed in the paper [55].

A direct generalization of the Dempster combining rule is its weight analog: if  $F_1 = (\mathcal{A}_1, m_1)$  and  $F_2 = (\mathcal{A}_2, m_2)$ , then  $F_Q = (\mathcal{A}, m_Q) = F_1 \otimes_Q F_2$ , where  $\mathcal{A} = \mathcal{A} \wedge \mathcal{A}_2$ ,

$$m_{\mathcal{Q}}(C) = \frac{1}{K_{\mathcal{Q}}} \sum_{A \cap B = C} q_{A,B} m_1(A) m_2(B),$$

and  $Q = (q_{A,B})$  is a measure of set similarity:  $q_{A,B} = q_{A,B} \in [0,1], q_{A,A} = 1$  for  $A \in 2^X \setminus \emptyset$ , and  $q_{A,B} = 0$  if  $A \cap B = \emptyset$ . The normalization coefficient is  $K_Q = \sum_{C} \sum_{A \cap B = C} q_{A,B} m_1(A) m_2(B) = \sum_{A,B} q_{A,B} m_1(A) m_2(B)$ . This rule was studied, e.g., in the paper [58].

Then  $Con_{Q}(F_{1}, F_{2}) = 1 - K_{Q}$  (7) with  $\gamma(A, B) = 1 - q_{A,B}$ ,  $A, B \in \mathcal{A}$ , is a measure of conflict agreed with the combining rule  $\otimes_{Q}$ .



**Example 13.** Consider the bodies of evidence  $F_1 = 0.2F_{\{b\}} + 0.1F_{\{c\}} + 0.3F_{\{a,b\}} + 0.4F_{\{b,c\}}$  and  $F_2 = 0.7F_{\{a,c\}} + 0.3F_{\{a,b,c\}}$  on the set  $X = \{a,b,c\}$  (Example 3). For  $\gamma(A, B) = 1 - jac_{A,B} = 1 - \frac{|A \cap B|}{|A \cup B|}$ , we have:  $Con_J(F_1, F_2) = 59/120$  and  $K(F_1, F_2) = 0.14$ . Clearly, the amount of conflict increases significantly due to consideration of weakly intersecting pairs of focal elements.

The conflict measure (7) is also convenient if the focal elements are subsets of the real axis  $\mathbb{R}$ .

**Example 14.** Consider two bodies of evidence  $F_1 = 0.5F_{[40,50]} + 0.3F_{[45,55]} + 0.2F_x$  and  $F_2 = 0.3F_{[30,60]} + 0.7F_{[40,50]}$  on the set X = [20,70], which predict the company's stock value (Example 8). For  $\gamma(A, B) = 1 - jac_{A,B} = 1 - \frac{|A \cap B|}{|A \cup B|}$ , where |A| is the Lebesgue measure of a set A on the real axis  $\mathbb{R}$ , we have  $Con_J(F_1, F_2) = 0.163$  and  $K(F_1, F_2) = 0$ .

The measure of conflict (7) can be applied in the case of fuzzy focal elements.

**Example 15.** Consider Example 14 with  $F_1 = 0.5F_{(35,40,50,55)} + 0.3F_{(40,45,55,60)} + 0.2F_x$  and  $F_2 = 0.3F_{(20,30,60,70)} + 0.7F_{(35,40,50,55)}$ , and let  $\tilde{A} = (a_1, a_2, a_3, a_4)$ ,  $a_1 \le a_2 \le a_3 \le a_4$  be a trapezoidal fuzzy number on the set X = [20, 70] with the membership function  $\mu_{\tilde{A}}(x) = \max\left\{0, \min\left\{\frac{x-a_1}{a_2-a_1}, 1, \frac{a_4-x}{a_4-a_3}\right\}\right\}, a_1 < a_2 \le a_3 < a_4$ . In this example, the kernels of all the fuzzy focal elements coincide with the corresponding arise focal elements from

this example, the kerners of an the fuzzy rocal elements coincide with the corresponding crisp focal elements from the previous example: Ker $\tilde{A} = \text{Ker}(a_1, a_2, a_3, a_4) =$  $[a_2, a_3]$ . Since  $\gamma(\tilde{A}, \tilde{B}) = 1 - \frac{|\tilde{A} \cap \tilde{B}|}{|\tilde{A} \cup \tilde{B}|} = 1 - \int_X \min\{\mu_{\tilde{A}}(x), \mu_{\tilde{B}}(x)\} dx / \int_X \max\{\mu_{\tilde{A}}(x), \mu_{\tilde{B}}(x)\} dx$ , we obtain  $Con_J(F_1, F_2) = 0.365$ .

#### 6. ROBUST METHODS TO ESTIMATE EXTERNAL CONFLICT

Some methods to calculate measures of conflict (in particular, the canonical measure of conflict) are unstable to "small" changes in bodies of evidence. At the same time, the bodies of evidence can be formed subjectively and depend on the characteristics of information sources. For example, assume that one expert makes a "cautious" forecast about the company's stock value in the interval (30, 40) c.u., whereas another expert makes an "optimistic" forecast (38, 45) c.u. In this case, we have two categorical bodies of evidence:  $F_A$  and  $F_B$ , where A = (30, 40) and B = (38, 43); the canonical measure of the conflict is  $K(F_A, F_B) = 0$ . In fact, when refining his forecast, the "cautious" expert meant the body of evidence  $F_1 = 0.8F_{(30,38)} + 0.2F_{(38,40)}$ ; the "optimistic" expert could also refine his forecast as the body of evidence  $F_2 = 0.2F_{(35,38)} + 0.7F_{(38,42)} + 0.1F_{(42,44)}$ . As a result,  $K(F_1, F_2) = 0.7$ .

An approach to increase the robustness of conflict estimation involves specialization-generalization procedures. A specialization procedure divides focal elements into smaller sets, simultaneously distributing the mass function over them. Without loss of generality, assume the following. For a body of evidence  $F_1 = (\{A_i\}, m_1)$ , a specialization is another body of evidence  $F_2 = (\{B_{ij}\}, m_2) = S(F_1)$  such that  $\bigcup_j B_{ij} = A_i$ and  $\sum_j m_2(B_{ij}) = m_1(A_i) \quad \forall i$ . (This fact will be denoted by  $F_2 \sqsubseteq F_1$ .) For a body of evidence  $F_1 = (\{A_i\}, m_1)$ , a generalization is another body of evidence  $F_3 = (\{C_{ij}\}, m_3) = G(F_1)$  such that  $\bigcap_j C_{ij} = A_i$ and  $\sum_j m_3(C_{ij}) = m_1(A_i) \quad \forall i$ . (This fact will be denoted by  $F_i \sqsubset F_3$ .)

Consider the specializations and generalizations that are sufficiently close to an original body of evidence  $F = (\mathcal{A}, m)$ . The degree of closeness will be measured by an imprecision index  $f : \mathcal{F}(X) \to [0, 1]$ [23]. (One example is the normalized generalized Hartley measure  $H_0(F) = \frac{1}{\ln|X|} \sum_{A \in \mathcal{A}} m(A) \ln|A|$ .) Any specialization does not increase the imprecision index:  $f(F_2) \leq f(F_1)$  for  $F_2 \sqsubseteq F_1$ . Similarly, any generalization does not reduce the imprecision index.

We denote by  $S_{\varepsilon}(F) = \{F' \sqsubseteq F : f(F) - f(F') < \varepsilon\}$ and  $G_{\varepsilon}(F) = \{F \sqsubseteq F' : f(F') - f(F) < \varepsilon\}$  the sets of all specializations and generalizations, respectively, of a body of evidence F that belong to its  $\varepsilon$ -neighborhood with respect to the imprecision index f. Also, let  $SG_{\varepsilon}(F) = S_{\varepsilon}(F) \cup G_{\varepsilon}(F)$ . A measure of conflict can be defined as  $K_{\varepsilon}(F_1, F_2) = \underset{\tilde{F}_i \in SG_{\varepsilon}(F_i), i=1, 2}{\text{MEAN}} (K(\tilde{F}_1, \tilde{F}_2))$ , where MEAN indicates some averaging operator. For example, let  $SG_{\varepsilon}(F_1) = \{F_{1,j}\}$  and  $SG_{\varepsilon}(F_2) = \{F_{2,k}\}$ , where  $F_{i,s} = F_{i,s}(\theta)$ , i = 1, 2, and  $\theta \in [0, 1]^N$  is the vector of parameters (mass function values). Then  $K(F_{1,j}, F_{2,k}) = \varphi_{j,k}(\theta)$ , where  $\theta \in D_{j,k} = \{\theta: |f(F_{1,j}(\theta)) - f(F_1)| < \varepsilon, |f(F_{2,k}(\theta)) - f(F_2)| < \varepsilon\}$ . The MEAN operator can be averaging of the mean integral values:

$$K_{\varepsilon}(F_1, F_2) = \frac{1}{\left|\{F_{1,j}\}\right| \cdot \left|\{F_{2,k}\}\right|} \sum_{j,k} I_{j,k} ,$$

where  $I_{j,k} = \frac{1}{V(D_{j,k})} \int_{D_{j,k}} \varphi_{j,k}(\mathbf{\theta}) d\mathbf{\theta}$ , and *V* denotes the Lebesgue measure on the parameter space. The set  $SG_{\varepsilon}(F)$  of specializations and generalizations for a body of evidence *F* can be formed by other methods.

**Example 16** [4]. Let  $X = \{x_1, x_2, x_3\}$ ,  $F_1 = F_{\{x_1, x_2\}}$ , and  $F_2 = F_{\{x_3\}}$ . In this case,  $K_0(F_1, F_2) = 1$ . Consider specializations and generalizations for the bodies of evidence  $F_1$  and  $F_2$ . We estimate the imprecision using the normalized generalized Hartley measure  $H_0$ . Choosing  $c = \frac{\ln 2}{\ln 3}$  gives  $H_0(F_1) = c$  and  $H_0(F_2) = 0$ . For specializations, we have  $S(F_1) = \{F_{1,0}, F_{1,1}\}$ , where  $F_{1,0} = F_1$ ,  $F_{1,1} = \theta F_{\{x_1\}} + (1-\theta)F_{\{x_2\}}$ ,  $H_0(F_{1,1}) = 0$ ,  $S(F_2) = \{F_{2,0}\}$ , and  $F_{2,0} = F_2$ . The generalizations of the bodies of evidence  $F_1$  and  $F_2$  are:

$$G(F_1) = \{F_{1,0}, F_{1,2}\},\$$

$$G(F_2) = \{F_{2,0}, F_{2,1}, \dots, F_{2,4}\},\$$
where  $F_{1,2} = \Theta F_{\{x_1, x_2\}} + (1 - \Theta) F_{\{x_1, x_2, x_3\}}$  and  $H_0(F_{1,2}) = c\Theta + 1 - \Theta$ ;

$$\begin{split} F_{2,1} &= \theta F_{\{x_1, x_3\}} + (1-\theta) F_{\{x_3\}} \text{ and } H_0(F_{2,1}) = c\theta ,\\ F_{2,2} &= \theta F_{\{x_2, x_3\}} + (1-\theta) F_{\{x_3\}} \text{ and } H_0(F_{2,2}) = c\theta ,\\ F_{2,3} &= \theta F_{\{x_1, x_3\}} + (1-\theta) F_{\{x_2, x_3\}} \text{ and } H_0(F_{2,3}) = c ,\\ F_{2,4} &= \theta F_{\{x_1, x_2, x_3\}} + (1-\theta) F_{\{x_3\}} \text{ and } H_0(F_{2,4}) = \theta . \end{split}$$

As a result,  $SG(F_1) = \{F_{1,0}, F_{1,1}, F_{1,2}\}$  and  $SG(F_2) = \{F_{2,0}, F_{2,1}, ..., F_{2,4}\}$ .

Letting  $\varepsilon < 1-c$  yields  $SG_{\varepsilon}(F_1) = \{F_{1,0}, F_{1,2}\}$  and  $SG_{\varepsilon}(F_2) = \{F_{2,0}, F_{2,1}, F_{2,2}, F_{2,4}\}$ . Now we have:  $K(F_{1,0}, F_{2,0}) = 1$ ,  $K(F_{1,0}, F_{2,1}) = K(F_{1,0}, F_{2,2}) = 1-\theta$  for  $0 < c\theta < \varepsilon$ ,  $K(F_{1,0}, F_{2,4}) = 1-\theta$  for  $0 < \theta < \varepsilon$ ,  $K(F_{1,2}, F_{2,0}) = \theta$  for  $0 < (1-\theta)(1-c) < \varepsilon$ ,  $K(F_{1,2}, F_{2,1}) = K(F_{1,2}, F_{2,2}) = \theta_1(1-\theta_2)$  for  $0 < (1-\theta_1)(1-c) < \varepsilon$  and  $0 < c\theta_2 < \varepsilon$ , and  $K(F_{1,2}, F_{2,4}) = \theta_1(1-\theta_2)$  for  $0 < (1-\theta_1) \times (1-c) < \varepsilon$  and  $0 < \theta_2 < \varepsilon$ .

Now,  $K_{\varepsilon}(F_1, F_2) = \frac{1}{2\cdot 4} \sum_{j,k} I_{j,k}$ , where  $I_{j,k} = \frac{1}{V(D_{j,k})} \times \int_{D_{j,k}} \varphi_{j,k}(\boldsymbol{\theta}) d\boldsymbol{\theta}$ , that is,  $I_{0,0} = 1$ ,  $I_{0,1} = I_{0,2} = \frac{c}{\varepsilon} \int_0^{\frac{c}{\varepsilon}} (1-\theta) d\theta = 1 - \frac{1}{2\varepsilon}\varepsilon$ ,  $I_{0,4} = \frac{1}{\varepsilon} \int_0^{\varepsilon} (1-\theta) d\theta = 1 - \frac{1}{2}\varepsilon$ ,  $I_{2,0} = \frac{1-c}{\varepsilon} \int_{1-\frac{\varepsilon}{\varepsilon}}^{1} \theta d\theta = 1 - \frac{1}{2\varepsilon}\varepsilon$ 

$$\begin{split} 1 &-\frac{1}{2(1-c)}\varepsilon, \qquad I_{2,1} = I_{2,2} = \frac{1-c}{\varepsilon} \cdot \frac{c}{\varepsilon} \int_{1-\frac{\varepsilon}{1-c}}^{1} \int_{0}^{\varepsilon} \theta_{1} (1-\theta_{2}) d\theta_{1} d\theta_{2} = \\ & \left(1-\frac{1}{2(1-c)}\varepsilon\right) \left(1-\frac{1}{2c}\varepsilon\right), \qquad I_{2,4} = \frac{1-c}{\varepsilon} \frac{1}{\varepsilon} \int_{1-\frac{\varepsilon}{1-c}}^{1} \int_{0}^{\varepsilon} \theta_{1} (1-\theta_{2}) d\theta_{1} d\theta_{2} = \\ & \left(1-\frac{1}{2(1-c)}\varepsilon\right) \left(1-\frac{1}{2}\varepsilon\right). \\ & \text{Thus, } K_{\varepsilon}(F_{1}, F_{2}) = 1-\varepsilon \frac{(2-c)(c+1)}{8c(1-c)} + \varepsilon^{2} \frac{(2+c)}{4c(1-c)} \text{ for } \varepsilon < 1-c. \blacklozenge$$

#### CONCLUSIONS

This paper has reviewed current research on the inconsistency (conflict) of information from several sources within belief function theory. In particular, the following aspects can be highlighted:

• At present, the scientific community has formed certain requirements to measures of external conflict. We mention the most important ones:

- A measure of conflict should reflect different degrees of conflict in bodies of evidence: from complete non-conflict to a particular degree of conflict (weak, simple, or strong).

 A measure of conflict should be antimonotonic with respect to specialization.

• These requirements underlie the axiomatics of a measure of external conflict. A measure of external conflict satisfying a given system of axioms has been found in a general form.

• There are several methods to estimate external conflict (metric, structural, algebraic, and combining rule-based).

• Together with a measure of conflict, belief function theory considers the concept of a distance between bodies of evidence. However, a measure of conflict is not reduced to calculating a distance.

• A measure of conflict can be calculated robustly using the generalization and specialization procedure.

At the same time, the analysis of information inconsistency within belief function theory has several open problems. Among them, we separate the following:

• constructing measures of external conflict that better reflect various degrees of conflict for bodies of evidence bodies and their structural features;

• investigating the agreement between combining rules and measures of conflict when choosing bodies of evidence for combination;

• studying interrelations between the concepts of the logical chain "inconsistency"—"agreement"— "mutual influence" of information sources;

• finding a general form of a measure of conflict that satisfies other systems of axioms;

• examining measures of conflict for bodies of evidence defined in a metric space.



Applied problems of conflict analysis of information sources, conflict management, etc. are topical as well.

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*This paper was recommended for publication by P.Yu. Chebotarev, a member of the Editorial Board.* 

> Received July 19, 2021, and revised August 27, 2021. Accepted August 31, 2021.

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#### Cite this article

Lepskiy, A.E. Analysis of Information Inconsistency in Belief Function Theory. Part I: External Conflict. *Control Sciences* 5, 2–16 (2021). http://doi.org/10.25728/cs.2021.5.1

Original Russian Text © Lepskiy, A.E., 2021, published in Problemy Upravleniya, 2021, no. 5, pp. 3–19.

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DOI: http://doi.org/10.25728/cs.2021.5.2

## THE STRUCTURE OF CREATIVE ACTIVITY

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**Abstract.** The specifics of creative activity are considered. There are three phases of such activity: discovering a new knowledge domain (subject matter) and accumulating basic knowledge, mastering the knowledge domain, and mass productive use. The life cycle of creative activity is analyzed. As shown by the analysis, creativity is concentrated in the stage of goal-setting only. A qualitative model for mastering knowledge (experience) and a graph-theoretic structural model of a knowledge domain are proposed. New models can be developed, and well-known models can be used to describe and study each phase of creative activity, including those introduced by the authors earlier: in the first phase, optimal distribution models for the researcher's efforts between the tested hypotheses and optimal scheduling models for tested hypotheses; in the second phase, mathematical models of experience; in the third phase, structural and algorithmic models and optimization models.

Keywords: creative activity, experience, creativity, knowledge domain, making and testing hypotheses.

#### INTRODUCTION

Activity is a dynamic interaction of a human with the reality in which he represents an actor (*subject*) purposefully influencing a *subject matter* (*object*) [Ошибка! Источник ссылки не найден.]. Activity is a form of human actions aimed at cognizing and transforming the surrounding world, humans themselves, and the conditions of their existence.

*Elementary* activity is understood as an activity whose goals, technologies, and result have no internal structure.<sup>1</sup>

In the monograph [2], an activity that is not elementary was called complex. In other words, *complex activity* (CA) is an activity with a nontrivial internal structure, multiple and (or) changing goals, actor, technology, and the subject matter's role in the goal context. The monograph [2] proposed a classification of activities and identified, in particular, regular activity and creative activity.

*Regular CA* is an activity performed using a known technology to obtain a priori specified results. The structure and technology of regular CA are deterministic.

*Creative CA* is an activity with a partially defined (incompletely known) technology at its beginning. Therefore, the technology of creative CA is developed when implementing this activity. The unknown technology is due to uncertain demand and (or) a priori uncertain specifications of the activity result.

Historically, there have been two paradigms for the definition and study of *creativity*. Within the first (activity) paradigm, creativity is treated as an activity. The most striking examples are *research activity* [3] and *artistic activity* [4].

<sup>&</sup>lt;sup>1</sup> In the case of elementary activity, there is no need to consider the actor and subject matter together with the activity itself: they play the role of an intuitively clear context. During such activity, only the subject matter evolves in accordance with the technology used by the actor.

<sup>•</sup> Creativity is a human activity that produces new material and nonmaterial values of social significance [5].

<sup>•</sup> Creativity is any practical or theoretical human activity in which new results (*knowledge*, decisions,



methods of action, or material products) arise (at least for the actor) [6].

• Creativity is an activity resulting in new material or nonmaterial values [7].

Within the second paradigm (*the psychology of creativity*), creativity is interpreted as "an interaction leading to development" [8, 9]. According to Ya.A. Ponomarev [8] and his followers, the subject must admit the influence of the object (environment) on himself; therefore, unlike activity, this interaction implies a cross-action of the object. The mechanism of this cross-action is associated with different categories: intuition, insight, cognitive unconsciousness, defocusing of attention, action's by-product, and others.

The two approaches mentioned are not contradicting but mutually complementing. Really, regardless of the position adopted, it is necessary to introduce certain assumptions to answer the following question: where does the image of a "creative product" appear in the subject's mind (an artist's intention, a researcher's hypothesis, etc.)? The psychology of creativity investigates, in particular, the reflection mechanisms of the surrounding world in the subject's mind considering the latter's experience. Within the activity approach, followed below, the assumption is the existence of a graph that objectively and adequately describes the knowledge domain's structure; see Fig. 4 in Section 5.

Creativity has become a very popular research topic in management and management psychology since the 1980s–1990s. For example, we refer to the surveys in [10–13]. There has been a significant flow of publications on this range of problems; see a scientometric analysis in the paper [14]. However, the results of creativity studies presented therein are qualitative and, at best, at the level of structural models [15].

The general model of activity, proposed in the book [16], describes the activity of subjects considering their active choice and activity in the environment. This model assesses the results of activity and mastering technologies. Moreover, it describes and examines the dynamics of knowledge, experience, and technologies using a set of admissible *structural elements of activity* (SEAs) as a function of time and previous actions, experience, activity result, subject's state, and values of *uncertainty factors* (UFs). However, the general model relations used directly yield no constructive results due to their analytical complexity.

This paper considers the specifics of creative activity and formulates its general structure in terms of the methodology of complex activity [2]. Within the life cycle model of a knowledge domain (subject matter), we identify the key phases of creative activity, showing that creativity is concentrated in the stage of goalsetting only. The remainder of this paper is organized as follows. Section 1 introduces basic definitions. In Section 2, the life cycle of a knowledge domain is considered. In Section 3, we localize the creative aspects in the life cycle of CA. Section 4 contains a qualitative model for developing knowledge (experience); Section 5, a structural model of a knowledge domain.

#### **1. DEFINITIONS**

Based on the concepts of knowledge, experience, cognition, and skill from the dictionary [17], we will formally define knowledge and experience. Within this paper, the concepts of experience and knowledge (individual or collective) are considered equivalent. Experience and knowledge are defined as the result of cognizing the reality, reflected in the consciousness of an individual or a group of individuals and in the material forms available to them (documents, etc.) through beliefs, notions, judgments, inferences, theories, and skills to perform definite activities in definite conditions. Consciousness is understood as the process and result of creating a world's model for particular purposes [18]. In this sense, creative activity closely relates to consciousness since both involve the creation of new knowledge.

Let us define a *knowledge element* as an assertion about the properties of the external world that is confirmed to be true at some time instant (or period) by observations when executing an SEA or is verifiable by executing an SEA (a system of SEAs).

For example, a knowledge element can be a subset of the Cartesian product for the admissible ranges of environment's parameters (including the subject matter of technologies): it can determine a set of admissible values of parameters, particularly at different time instants. In other words, a knowledge element can describe the sequence of changes in the states of environment's elements or the relation between their parameters, particularly under the effect of the subjects' activities.

Activity, including the capability for independent goal-setting, the choice of states (actions), and reflection, is the basic characteristic of a human (active element, further also called an *agent*).

A knowledge element is said to be known to an *active system* (AS, a system containing agents) if the hypothesis on the corresponding assertion is confirmed to be true when executing one or several SEAs.

For each knowledge element, there are *preconditions*: a set of knowledge elements that must be known for the corresponding hypothesis to be tested. At each time instant, a set of knowledge elements with testable



hypotheses can be determined. For such knowledge elements, the preconditions (the current "knowledge front" in Fig. 4 below) satisfy the current state of experience (knowledge).

Then each knowledge element at any time instant has one of the following states with respect to the AS:

 not available for hypothesis testing (the preconditions are not satisfied);

- available for hypothesis testing, but the hypothesis has not been tested;

- known (the hypothesis has been tested).

Let us formalize the agent's experience (knowledge) accumulated by the current time instant using a set algebra for the set of his currently known knowledge elements.

According to [2], a *technology* is a system of conditions, criteria, forms, methods, and means of consistently achieving a given goal. Following this definition, we will consider a technology consisting of two components: technological knowledge and objects (*means* of activity). In this case, technological knowledge is a subset of experience (knowledge) as a whole, and the object part of the technology (means) can be treated as an environment's component (activity resources).

As shown by historical practice, the evolution of knowledge (experience) of humankind has a "spasmodic" character: short periods (*scientific revolutions* according to T. Kuhn [19]) of forming new paradigms (new areas of knowledge) are replaced by relatively long periods of the so-called *normal development* (mastering and productive use of knowledge). This process naturally "selects" knowledge domains, i.e., subsets of sets of knowledge elements, possibly conceptually close and interconnected with each other. Let us consider their life cycles.

#### 2. THE LIFE CYCLE OF A KNOWLEDGE DOMAIN

Based on the public-historical practice, we identify three phases<sup>2</sup> of the life cycle of a knowledge domain; see Figs. 1 and 2.

**Phase I** (*discovering a new knowledge domain and accumulating basic knowledge*). In this phase, a set of SEAs is implemented sequentially and (or) in parallel to gain knowledge: to test the hypotheses that make up knowledge (experience) elements. When implementing

the SEAs for testing hypotheses, the significant conditions of the assertions are set (selected) by the subject (individual or collective). The corresponding mathematical models were considered in subsection 5.2 of the book [16].

Each such test has a binary a priori unknown (!) result. Therefore, true uncertainty is realized in each SEA [20]; the hypothesis is either rejected or confirmed depending on the result. These SEAs can be implemented in the form of activity over different objects: material (e.g., physical experiments), informational (e.g., mathematical modeling), and imaginable (thought experiments).

Assume that the first phase continues while potentially useful applications of a given knowledge domain are unknown, potentially useful goals are not formulated, and technologies for achieving them are not developed.

Each SEA and the content of this phase are intended to gain knowledge of the environment.

Thus, basic knowledge (experience) is accumulated: knowledge of the UF properties (their values and dynamic laws) and the CA technologies executed under a certain set of UF values are acquired. In the first phase, CA technologies aim to gain new knowledge (constructing a model of the surrounding reality) rather than obtain a useful result.

Phase II (mastering the knowledge domain) includes technology development and single productive and experimental use. A sign of the transition between phases I and II is the emerging hypotheses about a potentially useful application of knowledge (the formulation of new useful goals). In this phase, the goals of SEAs are to form technologies for obtaining useful results based on the UF model yielded by phase I. The subject does not choose the UF values: they are realized by the environment's "natural choice." In other words, some significant conditions of the assertions are determined by the UF values and are not set by the subject (unlike phase I). Note that different values of the UFs can be realized: the already known ones (new knowledge elements are not formed) or the ones not encountered before (new knowledge elements are formed). With multiple repetitions, a known technology either confirms operability (the desired productive result of CA) under all or most of the UF values or identifies new UF values. The models presented in [21, 22] adequately describe this process.

In each knowledge domain and their combination, a finite number of "reasonable," "rational," and "optimal" technologies can be created. (For example, the best electric motor, steam engine, or airplane design from the currently available materials.)

<sup>&</sup>lt;sup>2</sup> The phase boundaries and the beginning of the life cycle as a whole are conditional. In most cases, we can hardly indicate a single event (the time when it occurs) corresponding to the beginning of a particular phase or the life cycle. However, this is not required to build formal models.

SYSTEMS ANALYSIS



Fig. 1. The life cycle model of a knowledge domain.

**Phase III** (*mass productive use*). A sign of the transition between phases II and III is the massive (not single) use of known technologies. A set of SEAs with an already known technology (phase II) is being implemented to obtain a productive result (the ultimate goal of this technology). Note that different values of the UFs can be realized: the already known ones (the desired productive result is obtained) or the ones not encountered before (a new experience is formed, and the productive result can be lost). In the case of a new UF value, a return to phase I or II within the same knowledge domain occurs, or a new knowledge domain appears.

New SEAs formed in phases I and II correspond to the hypotheses about the knowledge elements; new SEAs formed in phase III, to the needs for the productive CA to obtain a useful result.

#### 3. CREATIVE ASPECTS IN THE LIFE CYCLE OF COMPLEX ACTIVITY

Which elements (procedural, internal, or external) are "responsible" for creative activity? In other words, where is creativity concentrated? To answer these questions, we will analyze *the CA's life cycle (LC)*; see Table 10 of the monograph [2] and Table 1 below.

Stage II of the CA's life cycle (Table 1) is goal-

*setting* (forming the structure of goals) and checking whether technologies exist for all subgoals. If not, the goal is the development of an appropriate technology. Note that sometimes goals—an anticipated image of a future result—can be formulated by the subject unconsciously, leading to "unexpected" results. (This is often the case for creativity; from the subject's viewpoint, goal-setting is "absent.")

According to Table 1, creativity is concentrated in the goal-setting stage only! Really, technology formation and implementation (stages III, IV, and the subsequent ones in Table 1) are always (!) performed by known means and methods: we cannot assure the result without reliable means and methods (technology). In this stage, the subject sets a goal not achieved before (moreover, it is unknown whether the goal can be achieved). The corresponding CA has no technology, and the subject decomposes the goal into subgoals (this is a heuristic, creative activity) and checks the presence of a known technology for each subgoal. In the absence of an appropriate technology, he decomposes the subgoal further. This procedure continues until known technologies are found for all subgoals of the goal. Implementing a set of known technologies to achieve the obtained structure of (sub)goals is a regular CA. The subsequent phase (reflection, checking the compliance between the result and the original goal) is a regular CA as well.



#### Phases, stages and steps of the life cycle of a complex activity element (A-SEA) and their content

Phase	Stage	no.	Step	Content
	I. Fixing de- mand and un- derstanding needs	1	Fixing demand and un- derstanding needs	A superior U-SEA or environment forms the demand for the results of CA element. The subject (actor) fixes the demand, understands the needs and decides to perform activity.
	II. Setting goals, structur- ing goals and tasks	2	Creating logical model	The need is structured and checked whether it is known or not (in the former case, the activity is regular). If CA is regular, this step comes to extracting information about the logical model from an information store. Otherwise, the structure of goals is formed. The goals are formulated in terms of the expected characteris- tics of the results of CA elements; see Section 6.1 for a detailed discussion of the result of CA. Consistency is checked/the structure of goals is modified. With each goal of A-SEA the role of the subject and technolo- gy is associated (this has been done for the result earlier); in other words, the characteristics of the subjects and technolo- gies are specified. The result of this step is a logical model, i.e., the structure of A-SEA in the form of a set of subordinate SEAs (L-SEAs) and elementary operations (L-Op).
Design	III. Selecting and develop- ing technology	3	Checking the readiness of technology and the suffi- ciency of resources	The presence of already known components of the A-SEA's technology is checked: the causal model of A-SEA, the technologies of all L-SEAs and the technologies of all L-Ops. The logical consistency of A-SEA and resource pools is checked: the availability and sufficiency of resources for assigning the subjects of U-SEAs and supporting the technologies of L-Ops, taking into account the use of these resources in parallel when implementing other SEAs. The result of this step is confirmation of the readiness of the technology, confirmation of the availability of necessary resources and transition to step 7, or the implementation of steps 4, 5 or 6, respectively.
		4	Creating cause-effect model	The causal relationships between the goals/results of subordi- nate elements (L-SEAs and L-Op) are determined and de- scribed. Possible events of uncertainty and the response rules for them are described (SEAs to be performed, or escalation to a higher level). The result of this step is the cause-effect model of A-SEA.
			5	Creating technology of lower-level elements
		6	Forming/modernizing resources	In the absence of necessary resources, goals responsible for their generation are set; SEAs ensuring the creation or modern- ization of resource pools are implemented. The result of this step is resource pools required.



Table 1 (continued)

Phase	Stage	no.	Step	Content
		7	Calendar-network	A calendar-network schedule is being formed. The consistency
			scheduling and resource	of key deadlines of needs is checked. The temporal consistency
			planning	of the calendar-network schedule and resource pool is checked,
				CA.
				In case of inconsistency, a return to steps 2-4 is carried out or
				the impossibility to meet the deadlines is escalated to the sub-
				ject of a upper SEA.
				The result of this step is a calendar-network schedule for the use
				of resources.
		8	Performing optimization	The dynamics of resources use is optimized, taking into account
				the possibility of using these resources for other CAs imple-
				mented in parallel.
ng	III. Selecting			for the use of resources.
Jesi	and develop-			
Ι	ing technology		Assigning actors and	The responsibility matrix is fixed, which describes a corre-
			defining responsibilities	spondence between the subjects of SEAs and personnel. In fact,
				the results of lower SEAs and hence the recursive implementa-
		9		tion of the life cycle of L-SEAs: all steps of the Design phase
				are carried out.
				The result of this step is the responsibility matrix, which togeth-
				er with the structure of A-SEA determines its organizational
			Allocating resources	In accordance with the technologies of elementary operations,
				the resources required for the implementation of technologies
		10		are request and allocated.
				The result of this step is the resource allocation matrix of ele-
				nonary operations.
	IV. Performin	11	Performing actions and	In accordance with the causal model, the preconditions for the
mentation	g actions and		obtaining results	start of actions of elementary operations (L-Ops) and L-SEAs
	sults			The elementary operations (L-Ops) are performed.
				The execution of subordinate L-SEAs is started.
nple				The result of this stage is the execution of actions by A-SEA
In				and also the result of its activity.
	V. Assessing	12	Assessing results and	Comparison of the characteristics of the result with the required
ion	results and		reflecting	ones.
lecti	reflecting			Comparison of the volumes of resources with the given ones.
Ref				Design of the requirements to the corrections of goals, technol-
1				бу, с.с.

The general scheme of the life cycles of CA and a knowledge domain describes well research CA, practical CA (including engineering CA), and artistic CA.

**Research CA.** Nowadays, well-established paradigms and research approaches (methods of study and presentation of results) have already been developed in many branches of knowledge. Alternative approaches are perceived with a priori suspicion: researchers tend to follow the well-known (regular) technology involving SEAs for hypotheses testing, which are also regular! The exception is the periods of scientific revolutions. Some examples are provided in Table 2.



Table 2

LC phase	UF values	Electricity examples	Atomic energy examples
Phase I. Discovering a new	Are generated or	- the experiments of the an-	- ideas and hypotheses of the an-
knowledge domain and accumu-	chosen by the sub-	cients with amber and wool,	cients about the structure of mat-
lating basic experience	ject and are realized	observation of electric eels,	ter,
	in parallel	– Gilbert's work and appa-	- Dalton's theory, Mendeleev's
		ratus,	discovery,
		- Franklin's experiments with	– X-rays,
		a kite,	<ul> <li>the Bohr and Rutherford atom</li> </ul>
		- Galvani's and Volt's exper-	models,
		iments and devices,	- Becquerel's observation, the
		<ul> <li>Faraday's generator,</li> </ul>	research works of Curie, Flerov,
		- others	and Petrzhak,
			– others
Phase II. Mastering the	Are realized,	– the use of electricity for	- the Manhattan Project (the first
knowledge domain.	whereas the subject	lighting, industrial drives,	reactor and bomb, enrichment
The goal is to develop and mas-	fixes the new UF	transport, etc.,	technologies),
ter technologies	values	<ul> <li>industrial devices designed</li> </ul>	<ul> <li>nuclear weapons,</li> </ul>
		by Tesla, Edison, Dolivo–	<ul> <li>the first nuclear power plants</li> </ul>
		Dobrovolsky, etc.	
Phase III. Mass productive use.	Are realized		
The goal is to use the developed		Patah produci	tion and mass use
technologies for obtaining new		Batch product	uon and mass use
results			

The LC of knowledge domains: some examples of phases

Artistic CA [4] is "arranged" in a similar way. Some examples of projects implemented by many people are:

- filming,
- setting up a theatrical performance,
- creating a monument.

Forming an idea (plot, or meaning) as *a holistic image* of a book or picture (or their elements if the subject divides the object of creativity into elements) is an elementary but creative activity. After that, the subject expresses the idea until liking the result, i.e., hypothesis testing (or rethinking of the idea) takes place as well. At the same time, the technology of applying paint, processing marble, or typing text (formulas) is regular. Such technology can also be the object of the first phase; see item C below. However, once developed, it becomes regular.

Thus, for artistic activity, we have the following:

- The idea of a work of art is a hypothesis (hypotheses) formed.

- Attempts to express the idea and checking whether the result fits the desires are tests of the hypotheses.

- Artistic technique (gouache, oil, clay, bronze, or brushstroke), which reflects the idea up to the author's individuality, is regular and "imported" from the industry.

Stages I and II (and stage V) of CA (see Table 1) are always implemented for (and in terms of) an in-

formation model of the subject matter. The other stages, III and IV, may require a CA associated with a physical object.

The goals or hypotheses of CA (stage II in Table 1) can describe any subsets of elements (their interconnections) in the body of knowledge, regardless of the representation model. In particular, hypotheses can describe new technology components.

The generalized scheme of a single creative SEA and the life cycle of a knowledge domain coincide with *the scheme of research activity*. It includes the following:

A. understanding of existing knowledge domains;

B. forming goals (in the case of creative CA, this is the hypothesis about achievable goals since at the time of goal-setting, the technology is unknown, and the possibility of achieving the goal is also unknown; in the case of research CA, the goal is to acquire new knowledge, i.e., directly test hypotheses about the laws of the researcher's environment;

C. testing the hypotheses;

D. generalizing and forming new laws (technologies);

E. passing to item A.

Items A and B correspond to goal-setting (forming the structure of goals) and checking whether technologies exist for all subgoals. We emphasize again: creativity is concentrated right here.

Hypothesis testing (item C) is always (!) performed



using a known technology: we cannot assure the result without a reliable technology (means, methods, and techniques to obtain the result). If the testing technology is unknown, the hypothesis is decomposed, and the causal structure of the lower-level hypothesis testing is determined, followed by the aggregation of the intermediate results. Decomposition, formation of a causal structure, and aggregation are well-known operations: proven components of the system-wide technology for achieving complex goals (testing complex hypotheses). Decomposition is performed until a known technology is found for all goals (hypotheses). It yields a fractal set of SEAs. Next, the actions of the SEAs are executed according to the causal structures. After that, the original hypothesis is either confirmed or rejected.

Hypotheses (item B) can describe any subsets of elements (and their interconnections) in the body of knowledge, regardless of the representation model. As noted above, hypotheses can describe new technology components.

Goals are always formed to satisfy the needs of some interested parties and (or) solve their problems (equivalent). In a particular case, such an interested party is the subject itself. (In research and artistic activities, the researcher or artist himself.)

Well, the execution of complex activities is always regular: all stages after goal-setting and structuring of goals (see Table 1) are implemented using a technology known at the beginning of the action. When the goal is structured, and the structure of the SEAs is created with a verified and known technology leading to the required result, the implementation of CA becomes regular.

#### 4. A QUALITATIVE MODEL OF KNOWLEDGE (EXPERIENCE) EVOLUTION

In any AS, agents exist and operate in the following way:

a) An active system as a complex entity "permanently" implements a set of regular SEAs.

b) Events of true uncertainty occur.

c) The agents of the AS perform reflection, comprehending the factual occurrence of these events.

d) The structure of goals is (re)formed, yielding the structure of SEAs and the structure of complex subjects.

e) For a new structure of goals, a new technology is developed or reduced by decomposition to known ones.

f) The implementation of a regular, albeit different, CA continues, and a return to item a) takes place.

The sequence a)–f) is implemented for all life cycle stages of knowledge domains; see Fig. 2.

The events of true uncertainty (b) and the events of re-forming the structures of goals (d) occur asynchronously. They are "connected" through the process of *reflection*, which has uncertain duration and result. Reforming the goals is a manifestation of the subject's true uncertainty.

Thus, the subject's uncertainty has two forms:

- deciding to carry out the activity (or refuse),

- forming the structure of goals.

Generally speaking, the life cycle of knowledge domains includes all three phases. In some cases, however, the development of a knowledge domain cannot lead (yet) to its productive use, and the life cycle is interrupted at the first or second phase. (The corresponding graphic images are shown at the top of Fig. 2.)

Reflection is an assessment of the existing experience and the environment, including the events of true uncertainty. On the one hand, reflection precedes goalsetting and is its source: this is how hypotheses are generated. On the other hand, reflection fixes the experience: this is how the hypotheses are confirmed or rejected.

The "general model" (see the Introduction and the book [16]) describes the evolution of knowledge (experience) using the dynamics of the sets of admissible SEAs and their dependence on the history. However, as noted above, the "general model" relations used directly yield no constructive results.

Therefore, let us concretize the "general model of ASs" [16] to investigate the development of knowledge (experience) analytically as the process of discovering new knowledge domains and accumulating "basic knowledge." For this purpose, we will:

i. abstract from the multiplicity of agents;

ii. discard the set functions describing the SEAs in favor of another representation of the evolution process.

Consider the implementation features of the life cycle of experience (knowledge) with goals i and ii; see Table 3. For each agent, the set of admissible actions consists of SEAs attributed to one of the characteristic subsets for different phases of the life cycle of knowledge domain:

• Phase I, SEAs for testing hypotheses available at the current level of experience;

• Phase II, SEAs for acquiring and mastering technologies;

• Phase III, SEAs for productive use of mastered technologies.





Fig. 2. The general model of knowledge (experience) evolution in an active system.

When implementing the SEAs of subset I, the agent chooses parameter values for the technology and environment, testing the hypothesis under precisely this combination of the values. True uncertainty manifests itself through the CA result, which is a priori unknown to the agent. If the environment's true uncertainty manifests itself so that the UFs take values differing from the required ones, then the CA result characterizes the test of another hypothesis not coinciding with the original one. Upon completion of the hypothesis testing, the sets of UF values and available technologies can be transformed; see Table 3 and Figs. 3 and 4.

When implementing the SEAs of subsets II and III, the agents choose the number of the CA element being executed (the technology parameters). Note that the environment's parameters (the number of the UF state) are implemented independently of the agent and are a priori uncertain for him. In this case, the CA result depends on the parameters values of the technology and the environment.

#### 5. A STRUCTURAL MODEL OF A KNOWLEDGE DOMAIN

Consider a connected circuit-free digraph, i.e., a *network* G = (N, E) with proper numbering. (No edges connect a greater-number vertex to a smaller-number one.) The network vertices correspond to knowledge domains (the sets of hypotheses and assertions), and the edge set  $E \subseteq N \times N$  reflects the logical interconnections of vertices; see Fig. 4.

We denote by  $N_i = \{j \in N \mid (j; i) \in E\}$  the set of immediate *predecessors* of vertex *i* in the network *G*,  $i \in N$ . Let the network *G* have a set  $N_0 \subseteq N$  of *inputs* (vertices without predecessors, which reflect axioms and (or) facts of recognized common knowledge).



Fig. 3. The life cycle of a knowledge domain.

Table 3

Agent's choice	UF value	Outcome	Consequence
Hypothesis testing,	Coincides with the required one	The hypothesis is tested	The set of admissible technologies is transformed.
phase I	Differs from the required one	Another hypothesis is tested	Possibly, the set of UF values is transformed
Mastering the technology,	Known	The hypothesis is	The level of mastering remains the same
phase II	Unknown	tested	The level of mastering increases
Using the technology, phase III	Known	The productive SEA	The expected useful result is obtained
	Unknown	is executed	The level of mastering increases, but the expected useful result may be not obtained

The phases of creative CA

Assume that each vertex of the network *G* has a *precondition*, i.e., a Boolean predicate  $\pi_i(\cdot)$  defined on the set of  $|N_i| + 1$  *inputs* of two types: the initial facts  $z_{N_i} = \{z_j, j \in N_i\}$  and *external conditions*  $\omega_i(\cdot) \in \Omega_i$ . This predicate calculates a binary *output* (new fact), i.e., a logical variable  $z_i = \pi_i(z_{N_i}, \Omega_i)$ , which is determinate if the output has the same value under any admissible external conditions, and is uncertain otherwise.

Consequently, a hypothesis with known initial facts is tested by finding the output value for some value(s) of the external conditions (phase I of creative CA). In phase II of creative CA, the invariability of the output value is checked for different (all admissible) values of the external conditions (UFs).

Thus, vertex *i* of the graph G is given by the tuple  $(N_i, \Omega_i, \pi_i(\cdot))$ , which includes the initial facts, external conditions, and a logical predicate.

We denote by  $G_t$  a subgraph of the graph G that is reliably known to the researchers at a time instant t. (No matter how many subjects in parallel test hypotheses, exchanging their results.) For example, the graph  $G_t$  in Fig. 4 is shaded.

Within the structural model of a knowledge domain, a *hypothesis* assumes that some assertion or a combination of some assertions is true. A hypothesis is a subgraph or vertex in which the incoming arcs of all vertices are either contained in it or originate from the graph  $G_t$ .

*Confirming* or *rejecting a hypothesis* is testing the definiteness (truth) of a corresponding assertion under all external conditions figuring in it. In a special case, the predicate is known, and it is necessary to find the maximum set of external conditions under which its value is definite.

The hypothesis testing model was considered in subsection 5.2 of the book [16]. In particular, the following problems were posed and solved therein: the optimal distribution of the researcher's efforts between the testing of various hypotheses and the optimal scheduling of the hypotheses.

As noted, in phase I of the knowledge domain's life cycle, each knowledge element (hypothesis described by a vertex in Fig. 4) has one of the following states:

not available for hypothesis testing (the preconditions are not satisfied; see the dotted line);

- available for hypothesis testing, but the hypothesis has not been tested (see the thin line);

- known (the hypothesis has been tested; see the thick line).

The set of known knowledge elements (the vertices indicated in Fig. 4 by thick lines) is *the current amount of knowledge*. The set of hypotheses available for testing (the vertices indicated in Fig. 4 by thin lines) is *the* 

current horizon of cognition. The current amount of knowledge and horizon of cognition form the subgraph  $G_t$  of the graph G known to the researcher at the current time instant. Each vertex indicated by a thin line represents one hypothesis or a set of independently tested hypotheses (sequentially or in parallel).

There are two types of agent's actions at each time instant: algorithmic and creative. The former actions consist in fully automatic generation and conceptual analysis of all logically possible consequences from the existing body of knowledge  $G_t$ . The result is a graph  $\hat{G}_t$ , conditionally called a "logical closure" of the graph  $G_t$ . The latter actions are the advancement and confirmation or rejection of hypotheses, i.e., new subgraphs  $G_t^h$  of the graph  $G \setminus \hat{G}_t$ . We denote by  $G^{\scriptscriptstyle h+}_{\scriptscriptstyle \star}$ the set of confirmed hypotheses. Then  $G_{t+1} = \hat{G}_t \cup G_t^{h+}$ . The advancement (generation) of hypotheses is an essentially creative and nonformalized stage. Therefore, in modeling, it is advisable to describe the occurrence of hypotheses and the duration of their testing in stochastic terms.

Thus, new hypotheses can be made automatically (algorithmically) or creatively in phase I and algorithmically or creatively in response to the events of true uncertainty in phases II and III of creative CA (see arrows 6 and 7, respectively, in Fig. 4).

The hypotheses testing process (confirming or rejecting hypotheses) can be formalized using the models below.

• In phase I:

- The transition to a new horizon of cognition occurs during scientific revolutions [19]. How and why does this happen? We will not attempt to answer, simply supposing that the graph G is given. This as sumption is the essential one for the models of creative CA under consideration.



Fig. 4. The phases of creative activity (the LC of a knowledge domain).

- Hypothesis testing is described by the model given in subsection 5.2 of the book [16].

- The arrows of types 1 and 2 correspond to specifying or increasing (decreasing) the dimension (analysis and synthesis, decomposition, and generalization).

• In phase II, technology development is described by "experience models" [21, 22].

• In phase III, practical activity is described by the general schemes of CA given in the monograph [2].

The chain of arrows 3 and 4 describes the "fundamental research => technology development => production" life cycle.

Arrows 5–7 show that the problems arising in phase II or III (situations of true uncertainty) may require a return to the previous phase(s) with the advancement and confirmation of new hypotheses and (or) the development of appropriate technologies.

#### CONCLUSIONS

This paper has identified three phases of creative activity (the life cycle of a knowledge domain):

 phase I: discovering a new knowledge domain and accumulating basic knowledge (generating and testing hypotheses);

- phase II: developing mastering the knowledge domain;

– phase III: mass productive use.

As shown, creativity is concentrated in the stage of goal-setting only (in the case of research or artistic activity, in the generation of hypotheses). New models can be developed, and well-known models can be used to describe and study each phase of creative activity, including those introduced by the authors earlier:

 in the first phase, optimal distribution models for the researcher's efforts between the tested hypotheses and choice models for an optimal sequence of tested hypotheses [16];

in the second phase, mathematical models of experience [21, 22];

- in the third phase, structural and algorithmic models [2] and optimization models [23].

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*This paper was recommended for publication by A.A. Voronin, a member of the Editorial Board.* 

Received May 16, 2021, and revised July 31, 2021. Accepted August 2, 2021.

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#### Cite this article

Belov, M.V., Novikov, D.A. The Structure of Creative Activity. *Control Sciences* **5**, 17–28 (2021). http://doi.org/10.25728/cs.2021.5.2

Original Russian Text © Belov, M.V., Novikov, D.A., 2021, published in Problemy Upravleniya, 2021, no. 5, pp. 20–33.

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DOI: http://doi.org/10.25728/cs.2021.5.3

## ADAPTIVE NEURAL-NETWORK-BASED CONTROL OF NONLINEAR UNDERACTUATED PLANTS: AN EXAMPLE OF A TWO-WHEELED BALANCING ROBOT<sup>1</sup>

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**Abstract.** This paper proposes a new method to control nonlinear underactuated plants for eliminating unmatched parametric uncertainties. The method is based on a model reference adaptive control. The controller consists of a basic LQ one and an adaptive compensator reducing the uncertainty norm under certain assumptions. The compensator involves a multilayer neural network due to its universal approximation properties. The network is trained online. The equations to tune the compensator's neural network parameters are derived using Lyapunov's second method and the backpropagation algorithm. The asymptotic convergence of the tracking error (the difference between the plant's and reference model's outputs) to a given domain is proved. The theoretical results are validated by numerical experiments with the developed control system for the mathematical model of a balancing LEGO EV3 robot in MATLAB.

Keywords: model reference adaptive control, balancing robot, suppression of unmatched parametric uncertainties, neural networks, online training, stability.

#### INTRODUCTION

In modern automatic control practice, the problems of controlling technical systems (plants) with a significant effect of parametric uncertainties are becoming increasingly important. Classical examples include manipulators [1, 2], unmanned and manned aerial vehicles in special operating modes [3, 4], industrial electric drives [5], and technological processes in chemical [6] and metallurgical [7, 8] industries.

Most modern methods for constructing control systems for plants with parametric uncertainties can be divided into robust approaches and model reference adaptive control methods. Robust systems are designed so that the performance criteria of the closed-loop control system (usually the phase and gain margins) satis-

<sup>1</sup>This work was partially supported by the Russian Foundation for Basic Research, project no. 18-47-310003-r\_a.

fy a priori requirements in the worst operating conditions of the plant. On the other hand, adaptive systems estimate the uncertainty online and then form a control action to minimize the plant's deviation from a reference [3]. Compared to robust approaches, model reference adaptive systems need no a priori knowledge of the range of variations in the plant's parameters (the maximum value of parametric uncertainty); with a sufficient power margin for the control action, they yield a reference performance instead of a compromise one.

All methods of constructing model reference adaptive control systems can be divided into direct, indirect, and composite ones; see [3, 10, 11]. In the first case, the parameters of a preselected-structure controller are directly tuned; in the second case, the parameters of the plant and (or) parametric uncertainty are estimated, and this information is used to calculate the controller's parameters. Composite adaptive control systems combine direct and indirect adaptive control approaches to improve the adaptation process.



The main problems of all three groups of model reference adaptive control methods are as follows [3, 10]:

- The tuned parameters converge to the ideal values only when satisfying the regressor's persistent excitation requirement, which is rather restrictive.

- From the practice viewpoint, the standard tuning loop yields an unsatisfactory quality of transients for tunable parameters, control, and tracking error (especially when the number of tunable parameters increases).

- The gain matrix of the adaptation loop is selected experimentally (manually).

- It is necessary to know the sign of the plant's gain matrix.

In recent years, much effort has been applied by domestic and foreign researchers to solve these problems. Among the last significant results, we mention the publications [12–15].

However, there is a fifth (more fundamental and less developed) problem in the model reference adaptive control theory: the assumption about the plant's adaptability [3, 10]. According to this assumption, model reference adaptive control methods in the general statement can be applied only if the plant's parametric uncertainty is matched with the control signal. It means the theoretical possibility of fully compensating the uncertainty by direct subtraction of the generated control. If the adaptability condition does not hold, direct compensation becomes impossible: the uncertainty is unmatched, and special methods are required for designing the control law and its tuning.

Generally speaking, there are two main classes of uncertainties unmatched with the control signal. The first class includes disturbances in the plant's description by autonomous differential equations. The second class includes disturbances arising in the plant's nonautonomous equations with a deficit of control channels (the so-called *underactuated systems with unmatched uncertainties*).

For a long time, unmatched uncertainties of the first class have been compensated using adaptive backstepping methods [16] and indirect methods based on tuning functions [17]. The disadvantages of these approaches include a high dynamic order of the control law and its tuning and higher complexity of the design process when increasing the plant's order. New methods have recently been proposed [18–20] to solve these problems—consider and compensate the effect of unmatched parametric uncertainties—in a different way. These solutions directly combine the theory of adaptive [10] and robust [9] control. In particular, the following procedure was proposed in [18, 20]. First, indirect model reference adaptive control methods were used to estimate the unmatched uncertainties. Then the resulting information was adopted to recalculate the parameters of the controller and the reference model using LMIs synthesis. With such an approach, the robustness of the closed loop system to arbitrary unmatched uncertainties is adaptively maintained, and the effect of matched uncertainties is compensated.

The literature suggests few model reference adaptive control methods to compensate the effect of unmatched uncertainties from the second group. The main difficulty here is that the deficit of control channels leads to the presence of one control signal in several equations. In the general case, this leads to a nontrivial problem of compensating control design. Various methods of changing coordinates [21-25] are well known [21] in geometric and nonlinear control theory to solve this problem. These methods allow passing from a plant's model with a deficit of control actions to an equivalent normal-form model. As a result, an appropriate control law can be chosen by the feedback linearization method [21]. The disadvantages of such methods are the complexity (or even impossibility) of calculating the exact transformation for nonlinear highdimensional plants and the dependence of the transformation itself on the plant's parameters. (Hence, it should have adaptation.) Therefore, the problem under consideration is still not completely solved.

This paper proposes a new approach to compensate an unmatched parametric uncertainty for an underactuated plant. It rests on the assumed existence of an ideal compensating control for reducing the unmatched uncertainty by a norm. With this assumption, we obtain compensating control by solving an optimization problem. Most real underactuated plants are described by nonlinear differential equations. Hence, the unmatched uncertainty is nonlinear, and the optimization problem turns out to be nonlinearly parameterized and, in the general statement, rather difficult to solve. To solve it in the general statement, we use artificial neural networks with universal approximating properties to form a compensating control action [26]. In this case, the parameter tuning laws of the compensator's neural network are designed by combining Lyapunov's second method and the backpropagation method.

The main result is a new neural-network-based compensating control law and an online algorithm to tune its parameters, which ensure the asymptotic con-



vergence of the tracking error of a nonlinear underactuated plant to a given domain with a chosen reference model.

A two-wheeled balancing robot is a classical example of a nonlinear underactuated plant. Therefore, we construct an adaptive neural-network-based control for such a robot without loss of generality as an illustrative example.

This paper uses the following notations: (.)(i, j) or  $(.)_{i,j}$  is an element standing at the junction of the *i*th row and *j*th column of a given matrix; det(.) is the matrix determinant; tr(.) is the matrix trace (the sum of all elements on the principal diagonal of a given matrix); vec(.) is the matrix vectorization (stacking the columns of a given matrix); diag $\{a \ b \ \dots \ c\}$  is a diagonal matrix with elements  $a,b,\dots,c$  on the principal diagonal;  $\|\cdot\|_{\infty}$  is the  $L_{\infty}$  norm of a given matrix;  $\|\cdot\|$  is the Euclidean vector norm or Frobenius matrix norm, depending on the context;  $(.)_{n\times n}$  is a matrix of dimensions  $n \times n$ .

#### 1. THE MATHEMATICAL MODEL OF A TWO-WHEELED BALANCING ROBOT

The differential equations describing the dynamics of a two-wheeled balancing robot are derived using the Euler–Lagrange second method [27]. After reducing to the Cauchy form, they are given by

$$\begin{cases} \dot{x}_{1} = x_{3}, \\ \dot{x}_{2} = x_{4}, \\ \dot{x}_{3} = -2(\beta + f_{w}) \Big[ E^{-1}(1,1) - E^{-1}(1,2) \Big] x_{3} - \\ 2\beta \Big[ E^{-1}(1,2) - E^{-1}(1,1) \Big] x_{4} + \\ \Big[ E^{-1}(1,1) MLRx_{4}^{2} + E^{-1}(1,2) MgL \Big] \sin(x_{2}) + \\ \alpha \Big[ E^{-1}(1,1) - E^{-1}(1,2) \Big] (u_{1} + u_{2}), \end{cases}$$
(1.1)  
$$\dot{x}_{4} = -2(\beta + f_{w}) \Big[ E^{-1}(2,1) - E^{-1}(2,2) \Big] x_{3} - \\ 2\beta \Big[ E^{-1}(2,2) - E^{-1}(2,1) \Big] x_{4} + \\ \Big[ E^{-1}(2,1) MLRx_{4}^{2} + E^{-1}(2,2) MgL \Big] \sin(x_{2}) + \\ \alpha \Big[ E^{-1}(2,1) - E^{-1}(2,2) \Big] (u_{1} + u_{2}), \end{cases}$$

where

$$E^{-1} = \begin{bmatrix} \frac{J + ML^2 + 2n^2 J_m}{\det(E)} & -\frac{MLR \cdot \cos(x_2) - 2n^2 J_m}{\det(E)} \\ -\frac{MLR \cdot \cos(x_2) - 2n^2 J_m}{\det(E)} & \frac{2J_w + (2m_w + M)R^2 + 2n^2 J_m}{\det(E)} \end{bmatrix},$$
$$\det(E) = \left(2J_w + (2m_w + M)R^2 + 2n^2 J_m\right) \times$$

$$\left(J + ML^2 + 2n^2 J_m\right) - \left(MLR \cdot \cos\left(x_2\right) - 2n^2 J_m\right)^2,$$
  
$$\beta = \frac{nK_t K_b}{R_m} + f_m, \text{ and } \alpha = \frac{nK_t}{R_m}.$$

The model (1.1) is obtained by assuming the structural and parametric identity of the actuating motors. The parameters of the model (1.1) have the following physical interpretation:  $J_w$  is the wheel's moment of inertia;  $m_w$  is the wheel's mass; M is the robot's mass; R is the wheel's radius; n is the gear ratio of the gearbox;  $J_m$  is the motor's moment of inertia; L is the distance between the center of mass and the wheel axis;  $K_t$  is the motor torque constant;  $R_m$  is the resistance of the motor's armature circuit;  $K_b$  is the back emf constant;  $f_m$  is the coefficient of friction between the robot's body (further called the body) and the motor shaft;  $f_w$  is the coefficient of friction between the wheel and the motion surface; J is the body's moment of inertia; g is the acceleration of gravity. The robot's state variables are the average angle of rotation of the wheels  $(x_1)$ , the body's angle of deflection from the normal  $(x_2)$ , the wheel turning rate  $(x_3)$ , and the body's rate of deflection from the normal  $(x_4)$ . The voltages  $u_1$ and  $u_2$  applied to the left and right motors, respectively, are the control action.

For convenience, let us introduce the following additional notations for the model (1.1):

$$f_{3}(x_{2}, x_{3}, x_{4}) =$$

$$-2(\beta + f_{w}) \Big[ E^{-1}(1, 1) - E^{-1}(1, 2) \Big] x_{3} -$$

$$2\beta \Big[ E^{-1}(1, 2) - E^{-1}(1, 1) \Big] x_{4} +$$

$$\Big[ E^{-1}(1, 1) MLRx_{4}^{2} + E^{-1}(1, 2) MgL \Big] \sin(x_{2}),$$

$$g_{31}(x_{2}) = g_{32}(x_{2}) = \alpha \Big[ E^{-1}(1, 1) - E^{-1}(1, 2) \Big],$$

$$f_{4}(x_{2}, x_{3}, x_{4}) =$$

$$-2(\beta + f_{w}) \Big[ E^{-1}(2, 1) - E^{-1}(2, 2) \Big] x_{3} -$$

$$2\beta \Big[ E^{-1}(2, 2) - E^{-1}(2, 1) \Big] x_{4} +$$

$$\Big[ E^{-1}(2, 1) MLRx_{4}^{2} + E^{-1}(2, 2) MgL \Big] \sin(x_{2}),$$
and 
$$g_{41}(x_{2}) = g_{42}(x_{2}) = \alpha \Big[ E^{-1}(2, 1) - E^{-1}(2, 2) \Big].$$
(1.2)

#### 2. PROBLEM STATEMENT

Using the notations (1.2), we write the model (1.1) in the state space:

$$\dot{x} = A_0 x + B_3 f_3(x_2, x_3, x_4) + B_4 f_4(x_2, x_3, x_4) + g(x)u,$$

$$A_0 = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}; B_3 = \begin{bmatrix} 0 \\ 0 \\ 1 \\ 0 \end{bmatrix}; (2.1)$$

$$B_4 = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \\ 1 \end{bmatrix}; g(x_2) = \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ g_{31}(x_2) & g_{32}(x_2) \\ g_{41}(x_2) & g_{42}(x_2) \end{bmatrix}.$$

Here  $x \in \mathbb{R}^4$  is the measured state vector of the robot;  $u = [u_1; u_2] \in \mathbb{R}^2$  is the vector of voltages applied to the left and right motors;  $f_3(x_2, x_3, x_4)$ ,  $f_4(x_2, x_3, x_4)$ , and  $g(x_2)$  are the nonlinear functions given by (1.2), which satisfy the Lipschitz smoothness conditions.

**Assumption 1**. *The control action u is such that*  $u_1 = u_2$ .

This assumption is classical when the robot's rotation about the center of mass in the horizontal plane is not required to control.

Due to Assumption 1, the model (1.1) with two control actions can be reduced to an equivalent model with one control action  $v = u_1$ :

$$\dot{x} = A_0 x + B_3 f_3 (x_2, x_3, x_4) + B_4 f_4 (x_2, x_3, x_4) + B_v,$$
  

$$B = \begin{bmatrix} 0 & 0 & g_3 (x_2) & g_4 (x_2) \end{bmatrix}^{\mathrm{T}}; \qquad (2.2)$$
  

$$\begin{cases} g_3 (x_2) = g_{31} (x_2) + g_{32} (x_2), \\ g_4 (x_2) = g_{41} (x_2) + g_{42} (x_2). \end{cases}$$

In addition, we will consider an auxiliary linear model obtained from (2.2) by linearizing the functions  $f_3(x_2, x_3, x_4)$  and  $f_4(x_2, x_3, x_4)$  and the nonlinear elements of the vector *B* in the neighborhood of the unstable equilibrium  $x_2 = 0$ :

$$\dot{x} = A_0 x + B_3 \overline{f}_3 (x_2, x_3, x_4) + B_4 \overline{f}_4 (x_2, x_3, x_4) + \overline{B}v = \overline{A}x + \overline{B}v.$$
(2.3)

**Remark 1.** To derive the model (2.3) from the model (1.1), we set some values of the physical and geometric parameters of the robot model (1.1) and use the equalities  $\lim_{x_2\to 0} \sin(x_2) = x_2$ ,  $\lim_{x_2\to 0} \cos(x_2) = 1$ , and

 $x_4^2 = 0$  holding in the neighborhood of the linearization point  $x_2 = 0$ . When constructing the linearized model (2.3), the robot's parameters can be uncertain.

Based on the linear model (2.3), the LQ-optimal control law can be calculated by

$$v_{LQ} = K_{LQ}(r-x) = K_{LQ}e,$$
 (2.4)

where  $r \in \mathbb{R}^4$  is the vector of reference signals for the robot's state variables, and the matrix  $K_{LQ}$  is found by minimizing the criterion

$$J = \frac{1}{2} \int_{0}^{\infty} x^{\mathrm{T}} Q_{\mathrm{LQ}} x + R_{\mathrm{LQ}} v^{2} d\tau \qquad (2.5)$$

with positive definite diagonal matrices  $Q_{LQ} \in R^{4x4}$  and  $R_{LQ} \in R$ .

The desired control performance for the nonlinear plant (2.2) is given by the system (2.3) with the controller (2.4):

$$\dot{x}_{ref} = A_{ref} x_{ref} + B_{ref} r,$$

$$A_{ref} = \overline{A} - \overline{B}K_{LQ}; B_{ref} = \overline{B}K_{LQ}.$$
(2.6)

To obtain an error equation between the nonlinear plant equations (2.2) and its linear reference model (2.6), we introduce the relations:

$$f_{3}(x_{2}, x_{3}, x_{4}) = \overline{f}_{3}(x_{2}, x_{3}, x_{4}) + \Delta_{f3},$$
  

$$f_{4}(x_{2}, x_{3}, x_{4}) = \overline{f}_{4}(x_{2}, x_{3}, x_{4}) + \Delta_{f4},$$
  

$$B = \overline{B} + \Delta_{B},$$
  
(2.7)

where  $\Delta_{f3}$ ,  $\Delta_{f4}$ , and  $\Delta_B$  are unknown Lipschitz smooth functions due to the parametric uncertainties and the nonlinearities for  $x_2 \gg 0$ .

Substituting the relations (2.7) into the model (2.2), we have

$$\dot{x} = A_0 x + B_3 \Big[ \overline{f_3} (x_2, x_3, x_4) + \Delta_{f3} \Big] + B_4 \Big[ \overline{f_4} (x_2, x_3, x_4) + \Delta_{f4} \Big] + \Big( \overline{B} + \Delta_B \Big) v.$$
(2.8)

Considering the expression (2.8), we choose the control law v in the form

$$v = v_{LQ} - v_{ad}$$
. (2.9)

Due to the expressions (2.3), (2.6), (2.7), and (2.9), equation (2.8) reduces to

$$\dot{x} = A_{ref} x + B_{ref} r +$$

$$+ \underbrace{B_{3}\Delta_{f3} + B_{4}\Delta_{f4} + \Delta_{B}v}_{\Lambda(z)} - \overline{B}v_{ad}, \qquad (2.10)$$

where  $\Lambda(z)$  is an unknown Lipschitz smooth function that describes the effect of parametric uncertainties and nonlinearities on the control performance of the balancing robot, and  $z = \begin{bmatrix} x_2 & x_3 & x_4 & v \end{bmatrix} \in D \subset \mathbb{R}^4$  is the

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variable of the function  $\Lambda(z)$  defined in a compact domain *D* of the space  $R^4$ .

The error equation between (2.10) and (2.6) has the form

$$\dot{e}_{ref} = A_{ref} e_{ref} + \Lambda(z) - \overline{B} v_{ad}, \qquad (2.11)$$

where  $e_{ref} = x - x_{ref}$  is the tracking error of the plant (2.10) with the reference model (2.6).

As is easily checked, the vector  $\overline{B}$  has no Moore– Penrose pseudoinverse matrix such that  $\overline{B}^{\dagger}\overline{B} = I$ . Hence,  $\Lambda(z)$  is a disturbance unmatched with the control signal. To design the control law  $v_{ad}$  in the adaptive problem statement, we accept the following assumption regarding compensation.

**Assumption 2.** There exists a compensating signal  $v_{ad}^*$  of the variable *z* such that

$$\left\| \Lambda(z) - \overline{B} v_{ad}^* \right\| \le \varepsilon_{\Lambda} < \Lambda_{\max},$$
  
$$v_{ad}^* = \arg \left[ \min_{\substack{v_{ad} \\ v_{ad}}} \left\{ \sup \left\| \Lambda(z) - \overline{B} v_{ad}^* \right\| \right\} \right],$$
 (2.12)

where  $\Lambda_{\max} = \sup_{\forall z \in L_{\infty}} \|\Lambda(z)\|$ , and  $\varepsilon_{\Lambda}$  is the approxima-

tion error of  $\Lambda(z)$  with the compensating signal  $v_{ad}^*$ .

**Remark 2.** If Assumption 2 is not satisfied, then the adaptive compensation of the disturbance  $\Lambda(z)$  is a fundamentally unsolvable problem in the class of smooth functions.

For clarity, we provide an illustrative example of when Assumption 2 holds. Let the difference  $\Lambda(z) - \overline{B}v_{ad}^*$  have the form

$$\Lambda(z) - \overline{B}v_{ad}^* = \begin{bmatrix} 0 \\ 0 \\ a_{32}x_2 + a_{33}x_3 + a_{34}x_4 \\ a_{42}x_2 + a_{43}x_3 + a_{44}x_4 \end{bmatrix} - \begin{bmatrix} 0 \\ 0 \\ b_3 \\ b_4 \end{bmatrix} v_{ad}^*.$$

In this case, Assumption 2 is satisfied under at least one of the following relations:

$$\frac{a_{32}}{a_{42}} \propto \frac{b_3}{b_4}; \ \frac{a_{33}}{a_{43}} \propto \frac{b_3}{b_4}; \ \frac{a_{34}}{a_{44}} \propto \frac{b_3}{b_4}.$$

The part of the uncertainty  $\Lambda(z)$  unmatched with the control signal can be compensated.

Thus, without any information about the structure of the function  $\Lambda(z)$  and the impossibility to compensate it fully, we pose the following problem: minimize the error (2.12) and ensure the asymptotic convergence of the tracking error (2.11) to a bounded set:

$$\lim_{t \to \infty} \left\| e_{ref} \left( t \right) \right\| \le \overline{\varepsilon}_{e_{ref}}, \qquad (2.13)$$

where  $\overline{\varepsilon}_{e_{ref}}$  is the limit tracking error.

Before presenting the main result, we introduce and justify a constraint on  $\overline{e}_{e_{ref}}$ . For this purpose, we estimate the minorant and majorant of the tracking error  $e_{ref}$ . Letting  $v_{ad} = v_{ad}^*$  and  $v_{ad} = 0$  in equation (2.11) yields lower and upper bounds on the derivative  $\dot{e}_{ref}$ :

$$A_{ref}e_{ref} + \Lambda(z) - \overline{B}v_{ad}^* \le \dot{e}_{ref} < A_{ref}e_{ref} + \Lambda(z). \quad (2.14)$$

For calculating the minorant and majorant of the error  $e_{ref}$  , consider the quadratic form

$$L = e_{ref}^{1} P e_{ref},$$

$$\lambda_{\min}(P) \left\| e_{ref} \right\|^{2} \le L \le \lambda_{\max}(P) \left\| e_{ref} \right\|^{2},$$
(2.15)

where *P* is the solution of the Lyapunov equation  $A_{ref}^{T} P + PA_{ref} = -Q, Q > 0.$ 

Due to (2.14), the derivative of the quadratic form (2.15) satisfies the two-sided inequality

$$e_{ref}^{\mathrm{T}} \left( A_{ref}^{\mathrm{T}} P + P A_{ref} \right) e_{ref} + 2e_{ref}^{\mathrm{T}} P \left[ \Lambda(z) - \overline{B} v_{ad}^{*} \right] \leq \dot{L} < \qquad (2.16)$$
$$e_{ref}^{\mathrm{T}} \left( A_{ref}^{\mathrm{T}} P + P A_{ref} \right) e_{ref} + 2e_{ref}^{\mathrm{T}} P \Lambda(z).$$

According to Assumption 2, from (2.16) we obtain:

$$-\lambda_{\min}(Q) \|e_{ref}\|^{2} + 2\varepsilon_{\Lambda}\lambda_{\max}(P) \|e_{ref}\| \leq 
\dot{L} < -\lambda_{\min}(Q) \|e_{ref}\|^{2} + 2\lambda_{\max}(P)\Lambda_{\max} \|e_{ref}\|.$$
(2.17)

For any a > 0 and b > 0,

$$-a^{2} + ab = \frac{1}{2} \Big[ -a^{2} - (a - b)^{2} + b^{2} \Big] \le -\frac{1}{2}a^{2} + \frac{1}{2}b^{2}.$$
(2.18)

Hence, the relations (2.17) imply

$$-\frac{\lambda_{\min}(Q)}{2\lambda_{\max}(P)}L + \frac{2\varepsilon_{\Lambda}^{2}\lambda_{\max}^{2}(P)}{\lambda_{\min}(Q)} \leq \dot{L} < -\frac{\lambda_{\min}(Q)}{2\lambda_{\max}(P)}L + \frac{2\lambda_{\max}^{2}(P)\Lambda_{\max}^{2}}{\lambda_{\min}(Q)}.$$

Considering (2.15), we have the following minorant and majorant of the tracking error:

$$\frac{1}{\lambda_{\min}(P)} e^{-\frac{\lambda_{\min}(Q)}{2\lambda_{\max}(P)}t} \left\| e_{ref}(0) \right\|^{2} + \frac{4\varepsilon_{\Lambda}^{2}\lambda_{\max}^{3}(P)}{\lambda_{\min}(P)\lambda_{\min}^{2}(Q)} \leq \left\| e_{ref} \right\|^{2} < (2.19)$$

$$\frac{1}{\lambda_{\min}(P)} e^{-\frac{\lambda_{\min}(Q)}{2\lambda_{\max}(P)}t} \left\| e_{ref}(0) \right\|^{2} + \frac{4\lambda_{\max}^{3}(P)\Lambda_{\max}^{2}}{\lambda_{\min}(P)\lambda_{\min}^{2}(Q)}.$$

Letting  $t \rightarrow \infty$  in (2.19), we finally estimate the limit tracking error as

$$2\varepsilon_{\Lambda} \frac{\lambda_{\max}(P)}{\lambda_{\min}(Q)} \sqrt{\frac{\lambda_{\max}(P)}{\lambda_{\min}(P)}} \leq \overline{\varepsilon}_{e_{ref}} < 2\Lambda_{\max} \frac{\lambda_{\max}(P)}{\lambda_{\min}(Q)} \sqrt{\frac{\lambda_{\max}(P)}{\lambda_{\min}(P)}}.$$
(2.20)

Well, according to the expressions (2.13) and (2.20), it is required to construct a compensating law  $v_{ad}$  that ensures the asymptotic convergence of the tracking error  $e_{ref}$  to a given set with the boundary  $\overline{\epsilon}_{e_{ref}}$ .

#### 3. AUXILIARY RESULTS FROM THE THEORY OF NEURAL-NETWORK-BASED CONTROL

To achieve this goal under an unknown structure of the nonlinear function  $\Lambda(z)$  and a nonlinear parametrization of the optimization problem (2.12), we will construct the compensating control  $v_{ad}$  using neural networks with their universal approximation properties [26]. This section provides auxiliary results from the theory of neural-network-based control necessary for further considerations.

**Proposition 1** [26]. Any continuous function f(z):  $R^n \rightarrow R$  can be uniformly approximated in a compact domain  $D \subset R^n$  using a neural network with one hidden layer with a sigmoidal activation function: for all  $\overline{\varepsilon}_{NN} > 0$  and  $z \in D$ , there exist matrices V and W and values  $b^1$  and  $b^2$  such that

$$\left\|f\left(z\right) - f_{NN}\left(z\right)\right\|_{\infty} = \left\|f\left(z\right) - W^{\mathsf{T}}\sigma\left(V^{\mathsf{T}}\overline{z}\right)\right\|_{\infty} \le \overline{\varepsilon}_{NN},$$
  
where

$$\overline{z} = \begin{bmatrix} b^1 & z \end{bmatrix}^{\mathrm{T}}, \sigma \left( V \cdot \overline{z} \right) = \begin{bmatrix} b^2 & \sigma_1 & \sigma_2 & \cdots & \sigma_{N_2} \end{bmatrix}^{\mathrm{T}},$$
$$V = \begin{bmatrix} \theta_1^V & \cdots & \theta_{N_2}^V \\ v_{1,1} & \cdots & v_{1,N_2} \\ \vdots & \ddots & \vdots \\ v_{N_1,1} & \cdots & v_{N_1,N_2} \end{bmatrix}^{\mathrm{T}}, \text{ and } W = \begin{bmatrix} \theta_1^W & \cdots & \theta_{N_3}^W \\ w_{1,1} & \cdots & w_{1,N_3} \\ \vdots & \ddots & \vdots \\ w_{N_2,1} & \cdots & w_{N_2,N_3} \end{bmatrix}^{\mathrm{T}}$$

The matrices  $V \in \mathbb{R}^{N_1 \times N_2}$  and  $W \in \mathbb{R}^{N_2 \times N_3}$  are the weight matrices of the hidden and output layers, respectively;  $b^1$  and  $b^2$  are the biases of the hidden and output layers, respectively;  $\sigma$  is the sigmoidal activation function of the hidden layer.

In other words, a neural network with a sigmoidal activation function approximates any continuous function of the variable z in a compact domain  $D \in \mathbb{R}^n$  with the error  $\overline{\varepsilon}_{NN} = = \sup_{z \in D} \|\varepsilon(z)\|$ :

$$f(z) = W^{\mathrm{T}} \sigma (V^{\mathrm{T}} \overline{z}) + \varepsilon(z).$$
 (3.1)

In this case, the error  $\overline{\varepsilon}_{NN}$  can be made arbitrarily small by increasing the number of neurons  $N_2$  in the hidden layer.

Proposition 1 establishes the existence of ideal parameters of a neural network, not determining their values. Therefore, equation (3.1) is interpreted as describing the ideal output of a neural network. It is used to introduce the concept of a neural network with the parameters tuned by training:

$$\hat{f} = \hat{W}^{\mathrm{T}} \sigma \left( \hat{V}^{\mathrm{T}} \overline{z} \right). \tag{3.2}$$

The error between the current (3.2) and ideal (3.1) outputs of the neural network has the form

$$e = W^{\mathrm{T}} \sigma \left( V^{\mathrm{T}} \overline{z} \right) - \hat{W}^{\mathrm{T}} \sigma \left( \hat{V}^{\mathrm{T}} \overline{z} \right) + \varepsilon (z).$$
(3.3)

Hence, the ideal parameters V and W of the neural network can be found by optimizing the error function (3.3) with respect to the tuned parameters:

$$(V, W) = \arg\left[\min_{(\hat{V}, \hat{W})} \left\{ \sup_{z \in D} |e| \right\} \right].$$
(3.4)

**Assumption 3.** The ideal weights of the neural network are bounded in a compact domain D:

 $\|V\| \leq V_M, \|W\| \leq W_M.$ 

The optimization problem (3.4) is nonlinearly parameterized due to the nonlinear activation function of the hidden layer. The error (3.3) is therefore rewritten in an approximate linearly parameterized form by expanding the activation function of the hidden layer into the Taylor series.

**Proposition 2** [28]. The linearly parameterized error  $e_{lin}$  of the neural-network-based approximation is given by

$$e_{lin} = e = \tilde{W}^{\mathrm{T}} \left( \sigma \left( \hat{V}^{\mathrm{T}} \overline{z} \right) - \sigma' \left( \hat{V}^{\mathrm{T}} \overline{z} \right) \hat{V}^{\mathrm{T}} \overline{z} \right) + \\ \hat{W}^{\mathrm{T}} \sigma' \left( \hat{V}^{\mathrm{T}} \overline{z} \right) \tilde{V}^{\mathrm{T}} \overline{z} + \varepsilon(z) - d,$$
(3.5)

where  $\sigma'(\hat{V}^T \overline{z}) = \text{diag} \{ 0 \quad \sigma'_1 \quad \sigma'_2 \quad \cdots \quad \sigma'_{N_2} \}$  denotes the derivative of the activation function of the hidden layer;  $\tilde{V} = V - \hat{V}$  is the parametric error of the hidden layer of the neural network;  $\tilde{W} = W - \hat{W}$  is the parametric error of the output layer of the neural network;




*d* is the residual term. The difference  $(\varepsilon(z) - d)$  in equation (3.5) is bounded [28, 29] due to the condition

$$\|\varepsilon(z) - d\| \le \alpha_1 \|\tilde{Z}\| + \alpha_2,$$
  

$$\alpha_1 > 0, \ \alpha_2 > 0, \ \tilde{Z} = \begin{bmatrix} \tilde{W} & 0\\ 0 & \tilde{V} \end{bmatrix}.$$
(3.6)

Then problem (3.4) is equivalent to the linearly parameterized problem

$$(V,W) = \arg\left[\min_{(\tilde{V},\tilde{W})} \left\{ \sup_{z \in D} |e_{lin}| \right\} \right].$$
 (3.7)

The laws to tune the current parameters of the neural network [28] are obtained by solving (3.7):

$$\begin{cases} \dot{\hat{W}} = \Gamma_{W} \left( \sigma \left( \hat{V}^{\mathsf{T}} \overline{z} \right) - \sigma' \left( \hat{V}^{\mathsf{T}} \overline{z} \right) \hat{V}^{\mathsf{T}} \overline{z} \right) e, \\ \dot{\hat{W}} \left( 0 \right) = 0_{N_{2} \times N_{3}}, \\ \dot{\hat{V}} = \Gamma_{V} \overline{z} e \hat{W}^{\mathsf{T}} \sigma' \left( \hat{V}^{\mathsf{T}} \overline{z} \right), \quad \hat{V} \left( 0 \right) = 0_{N_{1} \times N_{2}}. \end{cases}$$
(3.8)

These auxiliary results from the theory of neuralnetwork-based control will be used below to obtain the compensating control  $v_{ad}$ .

### 4. THE MAIN RESULT

Under Assumption 2, adding and subtracting the value  $\overline{B}v_{ad}^*$  from equation (2.11) yield

$$\dot{e}_{ref} = A_{ref} e_{ref} + \Lambda(z) - \overline{B} v_{ad} \pm \overline{B} v_{ad}^* = A_{ref} e_{ref} + \overline{B} \Big[ v_{ad}^* - v_{ad} \Big] + \Lambda(z) - \overline{B} v_{ad}^*.$$
(4.1)

By the problem statement,  $z \in D \subset R^4$ , and  $v_{ad}^*$  is a function of the variable *z* (Assumption 2). According to Proposition 1, the function  $v_{ad}^*$  can be approximated using an artificial neural network:

$$v_{ad}^{*} = W^{\mathrm{T}} \sigma \left( V^{\mathrm{T}} \overline{z} \right) + \varepsilon(z),$$
  
$$\overline{z} = \begin{bmatrix} b^{1} & z \end{bmatrix}^{\mathrm{T}} = \begin{bmatrix} b^{1} & x_{2} & x_{3} & x_{4} & v \end{bmatrix}^{\mathrm{T}}.$$
 (4.2)

Therefore, we choose  $v_{ad}$  in the form

$$v_{ad} = \hat{W}^{\mathrm{T}} \sigma \left( \hat{V}^{\mathrm{T}} \overline{z} \right). \tag{4.3}$$

Due to the expressions (4.3), (4.2), and (3.5), equation (4.1) reduces to

$$\dot{e}_{ref} = A_{ref} e_{ref} + \overline{B} \bigg[ \tilde{W}^{\mathrm{T}} \Big( \sigma \Big( \hat{V}^{\mathrm{T}} \overline{z} \Big) - \sigma' \Big( \hat{V}^{\mathrm{T}} \overline{z} \Big) \hat{V}^{\mathrm{T}} \overline{z} \Big) + \\ \hat{W}^{\mathrm{T}} \sigma' \Big( \hat{V}^{\mathrm{T}} \overline{z} \Big) \tilde{V}^{\mathrm{T}} \overline{z} + \varepsilon (z) - d \bigg] + \Lambda (z) - \overline{B} v_{ad}^{*}.$$
(4.4)

Based on the laws (3.8), we introduce the following tuning laws for the weights of the hidden and output layers of the compensating neural network (4.3):

$$\begin{cases} \dot{\hat{W}} = \Gamma_{W} \left[ \left( \sigma \left( \hat{V}^{T} \overline{z} \right) - \sigma' \left( \hat{V}^{T} \overline{z} \right) \hat{V}^{T} \overline{z} \right) \times \\ e_{ref}^{T} P \overline{B} - \sigma_{W} \hat{W} \right], \hat{W} (0) = 0_{N_{2} \times N_{3}}, \\ \dot{\hat{V}} = \Gamma_{V} \left[ \overline{z} e_{ref}^{T} P \overline{B} \hat{W}^{T} \sigma' \left( \hat{V}^{T} \overline{z} \right) - \sigma_{V} \hat{V} \right], \\ \dot{\hat{V}} (0) = 0_{N_{1} \times N_{2}}, \end{cases}$$
(4.5)

where  $\sigma_W > 0$  and  $\sigma_V > 0$  are the coefficients of the sigma modifications [10].

**Remark 3.** Contrary to popular belief, the compensating neural network (4.3) needs no preliminary autonomous training: it can be tuned by formulas (4.5), starting from the zero parameters of the layers, directly during the plant's operation.

Based on equation (4.4), we introduce the generalized error vector  $\zeta = \begin{bmatrix} e_{ref}^{T} & \text{vec}^{T}(\tilde{W}) & \text{vec}^{T}(\tilde{V}) \end{bmatrix}^{T}$  and study its properties.

**Theorem 1.** Let the compensating law  $v_{ad}$  be given by (4.3), and let its parameters be tuned by formulas (4.5). Then the generalized error  $\zeta$  is uniformly and ultimately bounded. Moreover, the steady-state tracking error  $e_{ref}$  can be reduced to satisfy inequalities (2.13) and (2.20) by increasing the number of neurons  $N_2$  in the hidden layer and decreasing the values of the coefficients  $\sigma_V$  and  $\sigma_W$ .

The proof of Theorem 1 is postponed to the Appendix.

Thus, we ensure the asymptotic convergence of the tracking error  $e_{ref}$  to a given domain using the neural-network-based compensating law (4.3) and tuning its parameters by formulas (4.5).

**Remark 4.** These recommendations for increasing the number of neurons  $N_2$  in the hidden layer and decreasing the values of the coefficients  $\sigma_V$  and  $\sigma_W$  have rather simple interpretations:

- Increasing the number of neurons  $N_2$  in equation (4.4) allows satisfying the inequality  $\Lambda(z) + \overline{B} \Big[ -v_{ad}^* + \varepsilon(z) - d \Big] \le \|\varepsilon_{\Lambda} + \overline{B} \big[ \varepsilon(z) - d \big] \| < \Lambda_{\max}$ . In other words, the uncertainty  $\overline{B} \big[ \varepsilon(z) - d \big]$  introduced by the neural network into the closed loop does not increase the system uncertainty after compensating  $\varepsilon_{\Lambda}$  compared to the initial value  $\Lambda_{\max}$ .



- Choosing small values of the coefficients  $\sigma_V$  and  $\sigma_W$  allows the tunable neural network to better approximate the ideal compensating signal  $W^{T}\sigma(V^{T}\overline{z})$ , thereby reducing the more error  $\tilde{W}^{\mathrm{T}}\left(\sigma\left(\hat{V}^{\mathrm{T}}\overline{z}\right) - \sigma'\left(\hat{V}^{\mathrm{T}}\overline{z}\right)\hat{V}^{\mathrm{T}}\overline{z}\right) + \hat{W}^{\mathrm{T}}\sigma'\left(\hat{V}^{\mathrm{T}}\overline{z}\right)\tilde{V}^{\mathrm{T}}\overline{z}.$  However, decreasing the values of the coefficients  $\sigma_w$  and  $\sigma_V$  also reduces the robustness of the tuning laws (4.5) to the uncompensated uncertainty  $\left\|\varepsilon_{\Lambda} + \overline{B}\right[\varepsilon(z) - d\right\|$ (under  $\sigma_V$  and  $\sigma_W$  close to 0, the value  $\gamma_1$  in (A.7) can be negative). Hence, the values  $\sigma_V$  and  $\sigma_W$  should be assigned by the classical tradeoff between the quality of tracking the ideal trajectory  $x_{ref}$  and system robustness to the uncompensated uncertainty  $\|\varepsilon_{\Lambda} + \overline{B}[\varepsilon(z) - d]\|.$ 

## 5. EXPERIMENTAL VALIDATION OF THE RESULTS

The developed control system was applied during experiments to the mathematical model of a LEGO EV3 balancing robot in Matlab/Simulink. The nominal values of the robot's parameters in the model (1.1) are given below.

Parameter	Value	Parameter	Value	
$J_w, \mathrm{kg} \cdot \mathrm{m}^2$	8.75·10 <sup>-6</sup>	$K_b$ , V·s/rad	0.468	
$m_w$ , kg	0.024	$R_m, \ \Omega$	6.69	
<i>R</i> , m	0.027	$f_m$	0.0022	
п	1	$f_w$	0	
$J_m$ , kg·m <sup>2</sup>	10-5	g, m/s <sup>2</sup>	9.81	
<i>L</i> , m	0.105	<i>M</i> , kg	0.8	
$K_t$ , N·m/A	0.317			

Nominal values of robot's parameters

The body's moment of inertia *J* was calculated by the formula  $J = ML^2/3 = 0.0029 \text{ kg} \cdot \text{m}^2$ . The gain matrix K<sub>LQ</sub> of the LQ controller was obtained using the robot's linearized model (2.3) with the nominal parameter values (see the table) by optimizing the criterion (2.5) with the matrices Q = I and R = 1:

$$K_{\rm LQ} = \begin{bmatrix} -0.7071 & -77.0619 & -1.5816 & -9.3949 \end{bmatrix}.$$

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The experiments involved a neural network with a sigmoidal activation function with four neurons on the input layer  $(N_1 = 5)$ , forty neurons on the hidden layer  $(N_2 = 40)$ , and one neuron on the output layer  $(N_3 = 1)$ . In all experiments, the variable parameters of the neural network's tuning loop were as follows:

 $\Gamma_W = 10^5 I_{N_2 \times N_2}, \Gamma_V = 10^{-3} I_{N_2 \times N_2}, \sigma_W = 0.1, \sigma_V = 0.001.$ For switching from the control v back to control u, the relation  $v = u_1 = u_2$  was used in the experiments (see Assumption 1).

Two experiments were carried out in total. The first experiment was intended to check the compensator (4.3) when approximating the uncertainties caused by changes in the robot's parameters during operation in the neighborhood of the linearization point  $x_2 = 0$ . The second experiment was intended to check the compensator (4.3) when approximating the uncertainties caused by changes in the robot's parameters and nonlinearities during operation in a domain out of the neighborhood mentioned. The initial conditions of the plant (2.1) and the reference model (2.6) were the same in all experiments, and the zero vector was used as the reference signal *r* (the stabilization mode of the balancing robot).

In the first experiment, the models (2.1) and (2.6) began to move from the state-space point  $x(0) = \begin{bmatrix} 0 & 0.01 & 0 & 0 \end{bmatrix}^T$ , and the function  $\Lambda(z)$  was caused by doubling the robot's nominal mass *M*. Figure 1 shows the elements of the vector  $v_{ad}$  and function  $\Lambda(z)$  in this experiment.

The transients in Fig. 1 demonstrate the high accuracy and sufficiently fast approximation of the disturbance  $\Lambda(z)$  using the neural network.

Figures 2*a* and 2*b* show the norms of the tracking errors  $e_{ref}$  and the control actions *v*, respectively, obtained in the first experiment using the control system with the neural network (LQ-NN) and without it (LQ).

The upper bound on the target set (2.13), (2.20) is given by the upper bound on the trajectory of the closed loop system with the LQ controller (for  $v_{ad} = 0$ ). Hence, Fig. 2*a* confirms the uniform and ultimate boundedness of the tracking error  $e_{ref}$  by the target set (2.20). This result validates the conclusions of Theorem 1. According to Fig. 2*b*, the costs of the total control action *v* to compensate the uncertainty  $\Lambda(z)$  are not significant.



**Fig. 1.** The results of neural-network-based approximation: (*a*)  $\Lambda_{13}(z)$  and (*b*)  $\Lambda_{14}(z)$ .



Fig. 2. (a) Norms of the tracking errors eref in systems LQ-NN and LQ, (b) control actions v in systems LQ-NN and LQ.

Figure 3*a* shows the uncertainty  $\|\Lambda(z)\|$  in the control system with the LQ controller and the uncertainty  $\|\Lambda(z) - \overline{B}v_{ad}\|$  after compensation in the control system with the neural-network-based compensator LQ-NN.

According to Fig. 3*a*, during the entire experiment, the area under the system uncertainty curve after compensation,  $\|\Lambda(z) - \overline{B}v_{ad}\|$ , is smaller than the area under the system uncertainty curve without compensation,  $\|\Lambda(z)\|$ . Hence, the following inequality holds:

$$\forall t > 0: \dot{e}_{ref} \leq A_{ref} e_{ref} + \Lambda(z) - \overline{B} v_{ad} < A_{ref} e_{ref} + \Lambda(z).$$

Using this result and considerations similar to

(2.15)–(2.20), we validate the convergence of the tracking error  $e_{ref}$  to the domain specified by inequality (2.20); see Theorem 1. Figure 3b confirms the possibility of further reducing the tracking error  $e_{ref}$  by increasing the number of neurons  $N_2$  in the hidden layer.

In the second experiment, the robot's mass was also doubled, but the robot (2.1) started moving from the initial state  $x(0) = \begin{bmatrix} 0 & 0.8 & 0 & 0 \end{bmatrix}^T$ . Therefore, the uncertainty  $\Lambda(z)$  was caused by the robot's nonstationary parameters and the nonlinearities. Figure 4 shows the norm of the error  $(\Lambda(z) - \overline{B}v_{ad})$  and the norms of the error  $e_{ref}$  and control actions obtained in the second experiment using the control system with the neural network (LQ-NN) and without it (LQ).



**Fig. 3.** (a) Uncertainties  $\|\Lambda(z)\|$  and  $\|\Lambda(z) - \overline{B}v_{ad}\|$ , (b) norms of the tracking errors  $e_{ref}$  under different values  $N_2$ .



**Fig. 4.** (a) Norms of the tracking errors  $e_{ref}$  in systems LQ-NN and LQ, (b) uncertainties  $\|\Lambda(z)\|$  and  $\|\Lambda(z) - \overline{B}v_{ad}\|$ , (c) control actions v in systems LQ-NN and LQ.

According to Fig. 4, the developed system can compensate the disturbance  $\Lambda(z)$  caused by parametric uncertainty and robot's nonlinearities. Moreover, since the upper bound on the target set (2.13), (2.20) is given by the trajectories of the closed loop system with the LQ controller (for  $v_{ad} = 0$ ), Fig. 4*a* confirms the convergence of the tracking error  $e_{ref}$  to the given domain in the case of nonlinear uncertainty  $\Lambda(z)$ . Comparing the costs of the control action *v* for compensating the linear (Fig. 2*b*) and nonlinear uncertainties (Fig. 4*b*), we arrive at the following result: the costs of the control action *v* grow proportionally with increasing the complexity of the function  $\Lambda(z)$ .

#### CONCLUSIONS

This paper has proposed an adaptive neuralnetwork-based control system for a two-wheeled balancing robot with a rigorously proved stability of a closed control loop and a neural network compensator trained online.

The new procedure for designing an adaptive neural-network-based control system can be applied not only to a two-wheeled balancing robot but also to other nonlinear underactuated plants (e.g., industrial cranes [30, 31], manipulators [32], underwater vehicles [33], vertical and/or short take-off and landing (V/STOL) aircrafts [34, 35], and other mechanical systems [35, 36]).

Proof of Theorem 1.

Consider the bounded quadratic form

$$V = e_{ref}^{\mathrm{T}} P e_{ref} + \operatorname{tr}\left(\tilde{W}^{\mathrm{T}} \Gamma_{W}^{-1} \tilde{W}\right) + \operatorname{tr}\left(\tilde{V}^{\mathrm{T}} \Gamma_{V}^{-1} \tilde{V}\right),$$
  

$$\lambda_{m} \left\|\zeta\right\|^{2} \leq V\left(\left\|\zeta\right\|\right) \leq \lambda_{M} \left\|\zeta\right\|^{2},$$
  

$$\lambda_{m} = \min\left\{\lambda_{\min}\left(P\right), \lambda_{\min}\left(\Gamma_{W}^{-1}\right), \lambda_{\min}\left(\Gamma_{V}^{-1}\right)\right\}, \ \lambda_{M} = \max\left\{\lambda_{\max}\left(P\right), \lambda_{\max}\left(\Gamma_{W}^{-1}\right), \lambda_{\max}\left(\Gamma_{V}^{-1}\right)\right\}.$$
(A.1)

We calculate the derivative of (A.1) taking into account (4.4) and (4.5):

$$\dot{V} = -e_{ref}^{\mathrm{T}} Q e_{ref} + 2e_{ref}^{\mathrm{T}} P \overline{B} \bigg[ \tilde{W}^{\mathrm{T}} \Big( \sigma \big( \hat{V}^{\mathrm{T}} \overline{z} \big) - \sigma' \big( \hat{V}^{\mathrm{T}} \overline{z} \big) \hat{V}^{\mathrm{T}} \overline{z} \big) + \hat{W}^{\mathrm{T}} \sigma' \big( \hat{V}^{\mathrm{T}} \overline{z} \big) \hat{V}^{\mathrm{T}} \overline{z} + \varepsilon(z) - d \bigg] + 2e_{ref}^{\mathrm{T}} P \bigg[ \Lambda(z) - \overline{B} v_{ad}^{*} \bigg] - 2 \mathrm{tr} \Big( \tilde{W}^{\mathrm{T}} \Gamma_{W}^{-1} \Gamma_{W} \bigg[ \Big( \sigma \big( \hat{V}^{\mathrm{T}} \overline{z} \big) - \sigma' \big( \hat{V}^{\mathrm{T}} \overline{z} \big) \hat{V}^{\mathrm{T}} \overline{z} \big) e_{ref}^{\mathrm{T}} P \overline{B} - \sigma_{W} \hat{W} \bigg] \Big) - 2 \mathrm{tr} \Big( \tilde{V}^{\mathrm{T}} \Gamma_{V}^{-1} \Gamma_{V} \bigg[ \overline{z} e_{ref}^{\mathrm{T}} P \overline{B} \hat{W}^{\mathrm{T}} \sigma' \big( \hat{V}^{\mathrm{T}} \overline{z} \big) - \sigma_{V} \hat{V} \bigg] \Big) = -e_{ref}^{\mathrm{T}} Q e_{ref} + 2e_{ref}^{\mathrm{T}} P \overline{B} \bigg[ \varepsilon(z) - d \bigg] + 2e_{ref}^{\mathrm{T}} P \bigg[ \Lambda(z) - \overline{B} v_{ad}^{*} \bigg] + 2 \mathrm{tr} \big( \tilde{W}^{\mathrm{T}} \sigma_{W} \hat{W} \big) + 2 \mathrm{tr} \big( \tilde{V}^{\mathrm{T}} \sigma_{V} \hat{V} \big).$$
(A.2)

Due to the expression (2.12) II (3.6), an upper bound on the derivative (A.2) is given by

$$\dot{V} \leq -\lambda_{\min}(Q) \|e_{ref}\|^{2} + 2 \|e_{ref}\|\lambda_{\max}(P)\|\bar{B}\|(\alpha_{1}\|\tilde{Z}\|+\alpha_{2}) + 2 \|e_{ref}\|\lambda_{\max}(P)\varepsilon_{\Lambda} + 2\sigma_{W}\|\tilde{W}\|\|\hat{W}\| + 2\sigma_{V}\|\tilde{V}\|\|\hat{V}\| = (A.3)$$
  
$$-\lambda_{\min}(Q) \|e_{ref}\|^{2} + 2 \|e_{ref}\|\lambda_{\max}(P)\|\bar{B}\|(\alpha_{1}\|\tilde{Z}\|+\alpha_{2}+\alpha_{3}) + 2\sigma_{W}\|\tilde{W}\|\|\hat{W}\| + 2\sigma_{V}\|\tilde{V}\|\|\hat{V}\|,$$

where  $\alpha_3 = \frac{\varepsilon_{\Lambda}}{\|\overline{B}\|}$ .

According to inequality (2.18), the terms in (A.3) satisfy the following upper bounds:

$$-\lambda_{\min}(Q) \|e_{ref}\|^{2} + 2 \|e_{ref}\|\lambda_{\max}(P)\|\bar{B}\|(\alpha_{1}\|\tilde{Z}\|+\alpha_{2}+\alpha_{3}) \leq \frac{1}{2} \left[-\lambda_{\min}(Q)\|e_{ref}\|^{2} - \left(\sqrt{\lambda_{\min}(Q)}\|e_{ref}\| - \frac{2\lambda_{\max}(P)\|\bar{B}\|}{\sqrt{\lambda_{\min}(Q)}}(\alpha_{1}\|\tilde{Z}\|+\alpha_{2}+\alpha_{3})\right)^{2} + \frac{4\lambda_{\max}^{2}(P)\|\bar{B}\|^{2}(\alpha_{1}\|\tilde{Z}\|+\alpha_{2}+\alpha_{3})^{2}}{\lambda_{\min}(Q)}\right] \leq \frac{-\lambda_{\min}(Q)\|e_{ref}\|^{2}}{2} + \frac{2\lambda_{\max}^{2}(P)\|\bar{B}\|^{2}}{\lambda_{\min}(Q)}(\alpha_{1}\|\tilde{Z}\|+\alpha_{2}+\alpha_{3})^{2},$$

$$\|\tilde{W}\|\|\hat{W}\| = -\|\tilde{W}\|^{2} + \|\tilde{W}\|\|W\| \leq -\frac{1}{2}\|\tilde{W}\|^{2} + \frac{1}{2}W_{M}^{2}, \|\tilde{V}\|\|\tilde{V}\| = -\|\tilde{V}\|^{2} + \|\tilde{V}\|\|V\| \leq -\frac{1}{2}\|\tilde{V}\|^{2} + \frac{1}{2}V_{M}^{2}.$$
(A.4)

Considering (A.4), the upper bound (A.3) takes the form

$$\dot{V} \leq \frac{-\lambda_{\min}(Q) \left\| e_{ref} \right\|^2}{2} - \sigma_W \left\| \tilde{W} \right\|^2 - \sigma_V \left\| \tilde{V} \right\|^2 + \frac{2\lambda_{\max}^2(P) \left\| \bar{B} \right\|^2}{\lambda_{\min}(Q)} \left( \alpha_1 \left\| \tilde{Z} \right\| + \alpha_2 + \alpha_3 \right)^2 + \sigma_W W_M^2 + \sigma_V V_M^2.$$
(A.5)



Since

$$\sigma_{W} \left\| \tilde{W} \right\|^{2} + \sigma_{V} \left\| \tilde{V} \right\|^{2} \geq \underbrace{\min\left\{ \sigma_{W}, \sigma_{V} \right\}}_{\sigma_{Z}} \left\| \tilde{Z} \right\|^{2},$$

the expression (A.5) reduces to

$$\dot{V} \leq \frac{-\lambda_{\min}(Q) \|e_{ref}\|^{2}}{2} - \sigma_{Z} \|\tilde{Z}\|^{2} + \frac{2\lambda_{\max}^{2}(P) \|\bar{B}\|^{2}}{\lambda_{\min}(Q)} \Big[\alpha_{1}^{2} \|\tilde{Z}\|^{2} + 2\alpha_{1}(\alpha_{2} + \alpha_{3}) \|\tilde{Z}\| + (\alpha_{2} + \alpha_{3})^{2}\Big] + \sigma_{W}W_{M}^{2} + \sigma_{V}V_{M}^{2} \leq \frac{-\lambda_{\min}(Q) \|e_{ref}\|^{2}}{2} - \left(\sigma_{Z} - \frac{2\lambda_{\max}^{2}(P) \|\bar{B}\|^{2}}{\lambda_{\min}(Q)} \alpha_{1}^{2}\right) \|\tilde{Z}\|^{2} + \frac{2\lambda_{\max}^{2}(P) \|\bar{B}\|^{2}}{\lambda_{\min}(Q)} \Big(2\alpha_{1}(\alpha_{2} + \alpha_{3}) \|\tilde{Z}\| + (\alpha_{2} + \alpha_{3})^{2}\Big) + \sigma_{W}W_{M}^{2} + \sigma_{V}V_{M}^{2}.$$
(A.6)

By analogy with (A.4), completing the square for the terms containing  $\|\tilde{Z}\|$  in (A.6), we write

$$\dot{V} \leq \frac{-\lambda_{\min}(Q) \left\| e_{ref} \right\|^{2}}{2} - \frac{1}{2} \left[ \sigma_{Z} - \frac{2\lambda_{\max}^{2}(P) \left\| \overline{B} \right\|^{2}}{\lambda_{\min}(Q)} \alpha_{1}^{2} \right] \left\| \widetilde{Z} \right\|^{2} + \frac{2\lambda_{\max}^{4}(P) \left\| \overline{B} \right\|^{4}}{\lambda_{\min}^{2}(Q)} \gamma_{1}^{-1} \left( 2\alpha_{1}(\alpha_{2} + \alpha_{3}) \right)^{2} + \frac{2\lambda_{\max}^{2}(P) \left\| \overline{B} \right\|^{2}}{\lambda_{\min}(Q)} (\alpha_{2} + \alpha_{3})^{2} + \sigma_{W} W_{M}^{2} + \sigma_{V} V_{M}^{2} = \frac{1}{2\lambda_{\min}^{2}(Q)} (\alpha_{2} + \alpha_{3})^{2} + \alpha_{W}^{2} + \alpha_{V} V_{M}^{2} = \frac{1}{2\lambda_{\min}^{2}(Q)} (\alpha_{2} + \alpha_{3})^{2} + \alpha_{W}^{2} + \alpha_{V} V_{M}^{2} = \frac{1}{2\lambda_{\min}^{2}(Q)} (\alpha_{2} + \alpha_{3})^{2} + \alpha_{W}^{2} + \alpha_{V}^{2} + \alpha_{V}^{2}$$

Here  $\,\gamma_1$  and  $\,\gamma_2$  are special notations for the compact form of (A.7); see above.

Applying the comparison lemma [21], we obtain a solution of inequality (A.7):

$$\left\|\zeta(t)\right\| \leq \sqrt{\frac{\lambda_M}{\lambda_m} e^{-\kappa t} \left\|\zeta(0)\right\| + \frac{\gamma_2}{\kappa \lambda_m}}.$$
(A.8)

Hence, the error  $\zeta$  is uniformly and ultimately bounded [3, 10, 21].

Let us prove the asymptotic convergence of the tracking error  $e_{ref}$  to a given domain from inequality (A.8). Letting  $t \rightarrow \infty$  and using the value  $\gamma_2$ , we arrive at the following limit estimate of the tracking error:

$$\left\| e_{ref} \right\| \leq \sqrt{\frac{2\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(Q\right)\lambda_{\min}\left(P\right)}\gamma_{2}} \leq 2\varepsilon_{\Lambda} \frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(Q\right)} \sqrt{\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(P\right)}} + \sqrt{\frac{2\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(Q\right)}} \left[ \frac{\sqrt{2}\lambda_{\max}^{2}\left(P\right)\left\|\overline{B}\right\|^{2}}{\lambda_{\min}\left(Q\right)}\gamma_{1}^{-\frac{1}{2}} \left(2\alpha_{1}\left(\alpha_{2}+\alpha_{3}\right)\right) + \frac{\sqrt{2}\lambda_{\max}\left(P\right)\left\|\overline{B}\right\|}{\lambda_{\min}^{\frac{1}{2}}\left(Q\right)}\alpha_{2} + \sqrt{\sigma_{W}}W_{M} + \sqrt{\sigma_{V}}V_{M}} \right].$$

$$(A.9)$$

Substituting this expression into (2.20), we check the inequality

$$2\varepsilon_{\Lambda}\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(Q\right)}\sqrt{\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(P\right)}} \leq 2\varepsilon_{\Lambda}\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(Q\right)}\sqrt{\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(P\right)}} + \alpha_{4} < 2\left\|\Lambda\left(z\right)\right\|\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(Q\right)}\sqrt{\frac{\lambda_{\max}\left(P\right)}{\lambda_{\min}\left(P\right)}}.$$
(A.10)



Inequality (A.10) holds if and only if the value  $\alpha_4$  is sufficiently small:

$$0 \le \alpha_{4} < 2 \Big[ \|\Lambda(z)\| - \varepsilon_{\Lambda} \Big] \frac{\lambda_{\max}(P)}{\lambda_{\min}(Q)} \sqrt{\frac{\lambda_{\max}(P)}{\lambda_{\min}(P)}}$$

By the definition (A.9), the value of the coefficient  $\alpha_4$ depends on those of the coefficients  $\sigma_V, \sigma_W$ ,  $\alpha_1$ , and  $\alpha_2$ . In turn, the coefficients  $\alpha_1$  and  $\alpha_2$  (see formulas (3.1) and (3.6)) are inversely proportional to the number of neurons  $N_2$ in the hidden layer. Therefore, the value of the coefficient  $\alpha_4$  can be reduced (thereby, ensuring inequality (A.10) and the asymptotic convergence of the tracking error  $e_{ref}$  to the given domain (2.13)) by increasing the number  $N_2$  and decreasing the values of the coefficients  $\sigma_W$  and  $\sigma_V$ . The proof of Theorem 1 is complete.

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This paper was recommended for publication by S.A. Krasnova, a member of the Editorial Board.

Received March 15, 2021, and revised August 17, 2021. Accepted August 24, 2021.

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#### Cite this article

Glushchenko, A.I., Petrov, V.A., Lastochkin, K.A. Adaptive Neural-Network-Based Control of Nonlinear Underactuated Plants: An Example of a Two-Wheeled Balancing Robot. *Control Sciences* **5**, 29–42 (2021). http://doi.org/10.25728/cs.2021.5.3

Original Russian Text © Glushchenko, A.I., Petrov, V.A., Lastochkin, K.A., 2021, published in *Problemy Upravleniya*, 2021, no. 5, pp. 34–47.

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# A BLOCK APPROACH TO CSTR CONTROL UNDER UNCERTAINTY, STATE-SPACE AND CONTROL CONSTRAINTS<sup>1</sup>

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**Abstract.** This paper designs a control law to maintain the temperature in the jacket of a continuous stirred tank reactor (CSTR). The standard mathematical model describing the reactor operation is extended by introducing the actuator's dynamics. The state-space and control constraints are considered by a nonlinear change of the variables of the plant's initial model using linear sat functions. In the transformed system, these constraints are considered by feedback control law design. The block approach allows linearizing the feedback control law by sequentially solving the first-order design subproblems. Under incomplete information on the state vector and the effect of exogenous disturbances, an observer of the state vector and disturbances is constructed to estimate the unknown signals with a given accuracy. The effectiveness of the proposed approach is illustrated by simulating the CSTR–DC motor system in MATLAB.

Keywords: CSTR, tracking problem, block approach, observer of state vector and disturbances, statespace and control constraints.

### INTRODUCTION

The continuous stirred tank reactor (CSTR) is widespread in the chemical industry. The CSTR dynamics are usually described by two nonlinear firstorder differential equations [1, 2], a reference model for applying and testing new control algorithms.

Nowadays, improving CSTR control is a topical problem that attracts the attention of many control theorists and practitioners [3–8].

Much research in this field involves the sliding mode approach [9–16]: it ensures robust properties of closed loop systems and invariance to exogenous disturbances acting in control channels. Note that within this approach, control laws are often represented by the plant's variables that physically cannot be discontinuous functions, e.g., the flow rate of the coolant in the CSTR jacket. Hence, the practical importance of the sliding mode approach in the automation of various technological processes is significantly reduced. State observers based on sliding modes and systems with deep feedback [17–21] are widely used to obtain information about the state vector and disturbances. Note that under disturbances, all these vectors can be estimated only within such an approach.

The problem of considering physical constraints on the state vector and control is underinvestigated in control theory. For example, only control constraints were taken into account in [22–24].

This paper proposes a complex solution for CSTR control that develops the original control law design method [25] for mechanical systems with constraints. It is methodologically based on a block approach to control [26], which decomposes high-dimensional problems into independently solvable subproblems of lower dimensions when designing feedback control laws and state observers. Treating the state variables as fictitious control actions, first of all, we satisfy the matching conditions for the disturbances in each sub-problem. (In other words, the disturbance belongs

to the control space.) In addition, using sat functions in local feedback law design, we ensure the bounded components of the state vector and controls.



<sup>&</sup>lt;sup>1</sup>This work was supported by the Russian Foundation for Basic Research, project no. 20-01-00363 A.

This paper is organized as follows. Section 1 briefly describes the operation principle of a CSTR and its mathematical model and states the problem. In Section 2, we construct an observer of the state vector and disturbances with discontinuous and continuous corrections with large gains. In Section 3, feedback design algorithms are developed by combining local feedback laws with continuous sat functions and discontinuous control of the armature voltage of a DC motor. Section 4 illustrates the effectiveness of the proposed algorithms by simulation modeling in MATLAB.

# 1. PLANT'S MATHEMATICAL MODEL. PROBLEM STATEMENT

A CSTR is a key component of equipment needed to complete chemical reactions in many chemical and biochemical industries. A complex chemical reaction occurs in a CSTR, e.g., converting a hazardous chemical waste (reagent) into an acceptable chemical product. The schematic diagram of the reactor is shown in Fig. 1.





An irreversible first-order exothermic reaction  $A \rightarrow B + \Delta H$  occurs in the tank, where *A* is a reagent, *B* is a product, and  $\Delta H$  is the thermal effect of the chemical reaction (enthalpy).

The CSTR volume is equal to V. A reagent having a concentration  $C_{Af}$ , a temperature  $T_{f}$ , and a density  $\rho$ is supplied to the reactor input at a flow rate q.

The temperature in the reactor is maintained at a certain level through setting a coolant temperature in the jacket,  $T_c$ , by controlling the flow rate of the coolant. The reactor's output product is characterized by a temperature T, a concentration  $C_A$ , and a flow rate q under the invariable substance volume in the reactor.

Assuming the invariable substance volume in the reactor, ideal mixing, and the invariable substance density in the reactor, the laws of conservation of mass and energy yield the following dynamic model of CSTR [16]:

$$\dot{C}_{A}(t) = \frac{q}{V} (C_{Af}(t) - C_{A}(t)) - k_{0}C_{A}(t)e^{-E/RT(t)},$$
  
$$\dot{T}(t) = \frac{q}{V} (T_{f}(t) - T(t)) + \frac{(-\Delta H)}{\rho C_{p}} k_{0}C_{A}(t)e^{-E/RT(t)} + (1)$$
  
$$\frac{UA}{\rho V C_{p}} (T_{c} - T(t)).$$

The parameters of the model (1) are described in Table 1.

Table 1

#### The parameters of the CSTR model: values and units of measurement

		Unit of
Parameter	Value	measure-
		ment
$C_A$	The concentration of product	kmol/m <sup>3</sup>
	В	
Т	Temperature in the reactor	K
	and product temperature	
q	Reagent flow rate	m <sup>3</sup> /min
V	Reactor volume	m <sup>3</sup>
$C_{af}$	Reagent concentration	kmol/m <sup>3</sup>
$k_0$	First-order reaction rate con-	$\min^{-1}$
	stant	
Ε	Activation energy	J/mol
R	Universal gas constant	J/(mol·K)
$T_f$	Reagent temperature	Κ
$\Delta H$	Reaction enthalpy	cal/kmol
ρ	Reagent density	g/m <sup>3</sup>
$T_c$	Coolant temperature	K
$C_p$	Reagent specific heat	cal/(K·g)
U	Heat-transfer coefficient	$W/(m^2 \cdot K)$
A	Heat delivery surface	$m^2$

The goal of CSTR control is to tune the coolant temperature  $T_c$  by changing the coolant flow rate so that the temperature T in the reactor corresponds to the desired values.

The coolant is supplied to the reactor by a pump with a DC motor. According to [27], the operation of this motor is described by the system of equations

$$\dot{x}_3(t) = a_{21}(gx_4(t) - m_L(t)),$$
  
$$\dot{x}_4(t) = a_{32}(u_2(t) - gx_3(t) - a_{31}x_4(t)), \qquad (2)$$

where  $x_3(t)$  is the motor shaft rotation frequency;  $x_4(t)$  is the armature current;  $u_2(t)$  is the armature voltage; g = const is a magnetic flux;  $m_I(t)$  is the



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load moment;  $a_{ij} = \text{const} > 0$  are the motor's design parameters.

The problem is to track a given coolant temperature in the jacket,  $T_d(t)$ , by the output variable T(t):

$$e_2(t) = T(t) - T_d(t) \rightarrow 0.$$

Assume that the pump load is  $m_L = mx_3^2(t)$ , m = const, and the temperature in the jacket is proportional to the coolant flow rate:  $T_c(t) - T(t) = x_3(T_{c0} - T(t))$ , where  $T_{c0} = \text{const} > 0$  denotes the coolant temperature at the jacket input. Under these assumptions, we write the plant's model (1), (2) in new variables:

$$\dot{x}_{1}(t) = -ax_{1}(t) - f_{1}(x_{2})x_{1}(t) + \xi_{1}(t),$$
  

$$\dot{e}_{2}(t) = -ae_{2}(t) + bf_{1}(x_{2})x_{1}(t) +$$
  

$$\beta x_{3}(T_{c0} - T_{d}(t) - e_{2}) + \xi_{2}(t),$$
  

$$\dot{x}_{3}(t) = a_{21}(gx_{4}(t) - mx_{3}^{2}(t)),$$
  
(3)

 $\begin{aligned} \dot{x}_4(t) &= a_{32}(u_2(t) - gx_3(t) - a_{31}x_4(t)), \\ \text{где} \quad x_1(t) &= C_A(t), \quad x_2(t) = T(t), \quad e_2(t) = x_2(t) - T_d(t), \\ \xi_1(t) &= aC_{Af}(t), \quad \xi_2(t) = a(T_f - T_d) - \dot{T}_d, \quad f_1(x_2) = \\ k_0 e^{-\gamma/x_2(t)}, a &= \frac{q}{V}, \gamma = \frac{E}{R}, b = \frac{\Delta H}{C_p \rho}, \text{ and } \beta = \frac{UA}{\rho V C_p}. \end{aligned}$ 

For the control of (3), let the output variable  $y(t) = e_2(t)$  and the armature current  $x_4$  be measured.

The disturbance  $\xi_1(t)$  depends on the input reagent concentration  $C_{Af}(t)$  and is difficult to measure in real-time. On the contrary, the input reagent temperature  $T_f(t)$  is rather easy to measure. In the sequel, assume that the disturbance  $\xi_1(t)$  is unmeasured, whereas the signal  $\xi_2(t)$  is measured.

In view of the technological process features, the system variables should satisfy the constraints

$$\begin{aligned} x_i \in [0, X_i], & i = 2, 3; |x_4| \le X_4; \\ |u_2| \le U, \ X_i, U = \text{const} > 0. \end{aligned}$$
 (4)

The problem of considering state-space and control constraints is currently underinvestigated in control theory. This paper proposes to take them into account in the plant's mathematical model by introducing a change of variables—a linear sat function. For details, see Section 3.

Control should maintain given values for the output product's temperature and concentration. In this case, the voltage  $u_2$  at the DC motor armature is the control action in the temperature loop, and the reagent flow rate q is the control action in the concentration loop. In what follows, we will design the temperature control loop in the reactor, assuming the value  $a = \frac{q}{V}$ 

to be known. The next section deals with estimating the state vector and disturbances in the system (3).

# 2. DESIGNING AN OBSERVER OF THE STATE VECTOR AND DISTURBANCES

Consider the problem of estimating the state vector and disturbances in the system (3) from the measured output variables  $e_2$  and  $x_4$  under the assumption that the signals q and  $\xi_2$  are measured.

The observability of the system (3) in the output variables  $e_2$  and  $x_4$  is established based on the following considerations:

• The DC motor model (the last two equations in (3)) does not depend on the variables of the reactor model (the first two equations in (3)), and its observability in the variable  $x_4$  is obvious due to the block observability form (BOF) [28]. The variable  $x_3$  is estimated by an observer designed using the DC motor model.

• Since the variables  $e_2, x_3$ , and  $\xi_2$  are measured (see below), the reactor model structurally coincides with the block observability form considering the disturbances [21]. Hence, it is also observable [21].

Note that the desired observer will be constructed designed in the sequence indicated above. Now we formulate some theoretical results on systems with deep feedback, which will be used in the further presentation.

**Lemma.** Consider a first-order system of the form  $\dot{\varepsilon}(t) = u + \xi(t),$  (5)

where  $\varepsilon(t), u(t), \xi(t) \in R$  denote the state variable, control, and disturbance, respectively, such that  $|\xi(t)| \leq E = \text{const}$  and  $|\dot{\xi}(t)| \leq \overline{E} = \text{const}, \forall t \geq 0$ .

*Then there exists a control law*  $u(t) = -\alpha \varepsilon(t)$ ,  $\alpha = \text{const} > 0$ , such that the relations:

- 1)  $|\varepsilon(t)| \le \Delta_0$ , 2)  $|\dot{\varepsilon}(t)| < \overline{\Delta}_0$ ,
- 3)  $|\alpha \varepsilon \xi| < \overline{\Delta}_0$

hold for any given constants  $\Delta_0, \overline{\Delta}_0, \alpha, \overline{\alpha} = \text{const} > 0$  in a finite time  $t_0 > 0$ .

Proof

1) The convergence of the state variable  $\varepsilon$  of the system (5) to a given neighborhood of zero,  $|\varepsilon| \le \Delta_0$ , is ensured by choosing a candidate Lyapunov function  $V = 0.5\varepsilon^2$ . We define the gain as  $\alpha_0 = \frac{E}{\Delta_0}$ .

Then the condition  $\dot{V} = \varepsilon \dot{\varepsilon} = \varepsilon (-\alpha_0 \varepsilon + \xi) \le |\varepsilon| (-\alpha_0 |\varepsilon| + E) < 0 \Longrightarrow -\alpha_0 |\varepsilon| + E < 0$  holds out of the domain



 $\begin{aligned} |\varepsilon| \leq \frac{E}{\alpha_0} = \Delta_0 \text{ . In the case } |\varepsilon(0)| \leq \Delta_0 \text{ , the variable does not} \\ \text{leave the given neighborhood } |\varepsilon(t)| \leq \Delta_0, \forall t > 0 \text{ ; in the case } |\varepsilon(0)| > \Delta_0 \text{ , it can approach this neighborhood from outside without limit. Hence, choosing <math>\alpha > \alpha_0$  ensures the convergence of the state variable to the given neighborhood  $|\varepsilon(t)| \leq \Delta_0$  in a finite time. Really, since  $\Delta = \frac{E}{\alpha} < \Delta_0 = \frac{E}{\alpha_0}$ , the variable will stay in the new neighborhood  $|\varepsilon(t)| \leq \Delta, \forall t > 0$ , if  $|\varepsilon(0)| \leq \Delta$ , or approach the new neighborhood  $|\varepsilon(t)| \leq \Delta, \forall t > 0$ , if  $|\varepsilon(0)| \leq \Delta$ , reaching the neighborhood hood  $\Delta_0 \leq \frac{E}{\alpha_0}$  in a finite time. The solution of (5) satisfies the bound  $|\varepsilon(t)| \leq |\varepsilon(0)e^{-\alpha t}| + |e^{-\alpha t}\int_{0}^{t} e^{\alpha t}\xi(\tau) d\tau| \leq |\varepsilon(0)|e^{-\alpha t} + |\varepsilon(0)| \leq 1$ .

 $\frac{E}{\alpha}(1-e^{-\alpha t}), \text{ and the time } t_0 \text{ of reaching the domain}$ 

$$|\varepsilon| \le \Delta_0, \ \forall t \ge t_0, \ \text{can}$$
 be estimated as:  
 $(|\varepsilon(0)| - \Delta)e^{-\alpha t} + \Delta = \Delta_0 \Longrightarrow t_0 = \frac{1}{\alpha} \ln\left(\frac{|\varepsilon(0)| - \Delta}{\Delta_0 - \Delta}\right), \ |\varepsilon(0)| > \Delta_0.$ 

2) A similar result applies to the derivative of the state variable in the system  $\ddot{\varepsilon} = -\alpha \dot{\varepsilon} + \dot{\xi}$  obtained by differentiating both sides of equation (5). Let  $|\dot{\varepsilon}| \leq \overline{\Delta}_0 = \frac{\overline{E}}{\overline{\alpha}_0}$  be a given neighborhood of zero. We choose  $\alpha > \overline{\alpha}_0$  and denote  $\overline{\Delta} = \frac{\overline{E}}{\alpha} < \overline{\Delta}_0 = \frac{\overline{E}}{\overline{\alpha}_0}$ . Then the relation  $|\dot{\varepsilon}| \leq \overline{\Delta}_0$ ,  $t \geq \overline{t}_0$ , where  $\overline{t}_0 = \frac{1}{\overline{\Delta}_0} \ln \left( \frac{|\dot{\varepsilon}(0)| - \overline{\Delta}}{\overline{\Delta}_0 - \overline{\Delta}} \right)$  and  $|\dot{\varepsilon}(0)| > \overline{\Delta}_0$ , will hold in a finite time.

3) Due to equality (5) and item 2), we have  $|\alpha\varepsilon(t) - \xi(t)| \le \overline{\Delta}_0, \forall t \ge \overline{t_0}$ . Hence, the disturbance can be estimated with a given accuracy:  $\alpha\varepsilon(t) = \xi(t) + \delta(t)$ ,  $\delta(t) \le \overline{\Delta}_0, \forall t \ge \overline{t_0}$ . Note that for constant disturbances, the system  $\ddot{\varepsilon} = -\alpha\dot{\varepsilon} + \dot{\xi}, \dot{\xi} = 0$ , has the solution  $\ddot{\varepsilon} = \dot{\varepsilon}(0)e^{-\alpha t}$ . Therefore, the disturbance estimate is asymptotically convergent:  $\alpha\varepsilon \rightarrow \xi, t \rightarrow \infty$ .

Choosing the parameter  $\alpha$  based on the condition  $\alpha \ge \max\{E/\Delta_0, \overline{E}/\overline{\Delta}_0\}$  ensures the desired convergence of the variable  $\varepsilon(t)$  and its derivative  $\dot{\varepsilon}(t)$  to the given domains: 1)  $|\varepsilon(t)| \le \Delta_0$ , 2)  $|\dot{\varepsilon}(t)| \le \overline{\Delta}_0$ , 3)  $|\alpha\varepsilon(t) - \xi(t)| \le \overline{\Delta}_0$ ,  $\forall t \ge \max\{t_0, \overline{t_0}\}$ . The proof of this lemma is complete.  $\blacklozenge$ 

For estimating the state vector of the system (3), we design an observer of the form

$$\dot{z}_{1} = -[a + f_{1}(x_{2})]z_{1} + v_{1},$$

$$\dot{z}_{2} = -az_{2} + \beta z_{3}(T_{c0} - T_{d} - e_{2}) + \xi_{2} + v_{2},$$

$$\dot{z}_{3} = a_{21}gx_{4} + v_{3},$$

$$\dot{z}_{4} = a_{32}(u_{2} - a_{31}z_{4}) + v_{4},$$
(6)

where  $\nu_i$  are the observer corrections determined below.

Using formulas (3) and (6), we rewrite the system in the residues  $\varepsilon_i = x_i - z_i$ , i = 1, 3, 4, and  $\varepsilon_2 = e_2 - z_2$ :  $\dot{\varepsilon}_1 = -(a + f_1(x_2))\varepsilon_1 + \xi_1(t) - v_1$ ,  $\dot{\varepsilon}_2 = -a\varepsilon_2 + \beta(T_{c0} - T_d - e_2)\varepsilon_3 + bf_1(x_2)x_1(t) - v_2$ ,  $\dot{\varepsilon}_3 = -a_{21}mx_3^2 - v_3$ ,  $\dot{\varepsilon}_4 = -a_{32}a_{31}\varepsilon_4 - a_{32}gx_3 - v_4$ . (7)

Within the cascade approach [28], a design procedure for the observer (7) of the state vector and disturbances includes the following steps:

1. We choose an appropriate correction for the last subsystem of the system (7), i.e., the discontinuous function  $v_4 = l_4 \operatorname{sign}(\varepsilon_4)$ , where  $l_4 = \operatorname{const} > |a_{32}gx_3|$ , to ensure the occurrence of a sliding mode on the line  $\varepsilon_4 = 0$ . The average (equivalent) value of the discontinuous signal is  $v_{4eq} = -a_{32}gx_3$ . In practice, the equivalent value of the discontinuous control can be obtained using the first-order filter  $\mu \dot{\tau} = -\tau + v_4$ , where  $\mu > 0$  and  $v_{4eq} \approx \tau$  [28].

2. Using the equivalent value and  $x_3 = -v_{4eq} / (a_{32}g)$ , we construct the correction  $v_3 = -a_{21}m \times (v_{4eq} / (a_{32}g))^2 + l_3[-v_{4eq} / (a_{32}g) - z_3]$  for the third subsystem of the system (7). As a result, the third subsystem of (7) takes the form  $\dot{\varepsilon}_3 = -l_3\varepsilon_3$ , and the variable  $\varepsilon_3$  asymptotically vanishes with an appropriately assigned coefficient  $l_3 > 0$ :  $\varepsilon_3 \rightarrow 0 \Longrightarrow z_3 \rightarrow x_3$ .

3. Under the assumption  $\varepsilon_3 \rightarrow 0 \Rightarrow z_3 \rightarrow x_3$ , the second equation of the system (7) takes the form

$$\dot{\varepsilon}_2 = -a\varepsilon_2 + bf_1(x_2)x_1 - v_2.$$

We choose the correction  $v_2 = l_2 \operatorname{sign}(\varepsilon_2)$ ,  $l_2 = \operatorname{const} > bf_1(x_2)x_1$ , to ensure the occurrence of a sliding mode on the plane  $\varepsilon_2 = 0$ . The average value of the discontinuous signal is  $v_{2eq} = bf_1(x_2)x_1$ .

4. In the last step, we choose  $v_1 = (-f_1(x_2) + l_1) \left( \frac{v_{2eq}}{bf_1(x_2)} - z_1 \right) = (-f_1(x_2) + l_1)\varepsilon_1.$ 

Then the first equation of (7) takes the form  $\dot{\varepsilon}_1 = -(a+l_1)\varepsilon_1 + \xi_1$ . Under the assumptions  $\left|\dot{\xi}_{1}(t)\right| \leq \overline{E},$  $|\xi_1(t)| \leq E_1$  and the relations  $|\varepsilon_1| \le \frac{E_1}{a+l_1} = \Delta_1$  and  $|\dot{\varepsilon}_1| \le \frac{\overline{E}_1}{a+l_1} = \overline{\Delta}_1$  hold by the lemma. Thus, assigning an appropriate coefficient  $l_1 > 0$ , we ensure the desired stabilization accuracy  $|\varepsilon_1| \le \Delta_1$  in a finite time, thereby estimating the varia $x_1(t) = z_1(t) + \delta_1(t)$ ble and the disturbance  $\xi_1(t) = (a+l_1)\varepsilon_1(t) + \overline{\delta}_1(t)$  with the desired accuracy:  $|\delta_1(t)| \le \Delta_1$  and  $|\overline{\delta}_1(t)| \le \overline{\Delta}_1$ . Note that  $\delta_1, \overline{\delta}_1 \to 0$  as  $l_1 \rightarrow \infty$ . (The estimation accuracy grows infinitely in the case of large gains.)

Thus, the state vector of the system (3) and the disturbance  $\xi_1(t)$  have been estimated.

This design procedure for an observer of the state vector and disturbances involves discontinuous and continuous corrections. Clearly, the design based on sliding modes is easier from a computational viewpoint. However, the average (equivalent) values of corrections are produced by first-order filters [28], which dynamically extends the state space of the original model. The technique for linearizing the residual equations (7) with continuous corrections (Step 4 of the observer design procedure) can be used in all other steps by analogy. In this case, the cascade approach [28] allows decomposing the observer design procedure into one-dimensional subproblems successively solved with a predetermined accuracy of the resulting estimates without extending the state space of the closed loop system.

The next section considers a control law design procedure for the system (7) under complete information (the available estimates of the state vector and disturbances).

#### **3. FEEDBACK LAW DESIGN**

Using the block approach, we present the general solution under complete information about the state vector and disturbances provided by the observer (6). In the case of complete information and the state-space and control constraints (4), the feedback law design procedure based on the block approach [21] includes the following steps.

Step 1. We rewrite the second equation of the system (3) as

$$\dot{e}_2 = -ae_2(t) + \beta d(t)x_3 + \overline{\xi}_2(t), \qquad (8)$$

where  $\overline{\xi}_2 = bf_1(x_2)x_1 + \xi_2$  and  $d(t) = T_{c0} - T_d(t) - e_2(t) < 0$ ,  $|d| \le D = \text{const}$ .

In the system (8), the control action is the motor shaft rotation frequency  $x_3$ . Under pump loading, this frequency is positive. We introduce the change of variable

$$e_3 = \beta d(t)x_3 + M_3 \text{sat}^+(s_3),$$
 (9)

where  $s_3 = k_2 e_2 + \overline{\xi}_2$ . Stabilizing the variable  $e_3 = \beta d(t)x_3 + M_3 \text{sat}^+(s_3) = 0$ , we consider the constraint (4) on the variable  $x_3 \in [0, X_3]$  by choosing  $0 < M_3 < \beta DX_3$ .

**Definition.** Let M = const > 0 and b = const. Then  $M\text{sat}(s) = \min(M, |s|)$  sign(s) and  $M\text{sat}^+(s) = M\text{sat}(s)0.5[1 + \text{sign}(s)]$ .



Fig. 2. The graph of Msat(t).



Fig. 3. The graph of  $Msat^+(t)$ .

Substituting the sum (9) into formula (8), we transform the second equation to

$$\dot{e}_2 = -ae_2 + e_3 - M_3 \text{sat}^+(s_3) + \overline{\xi}_2.$$
 (10)

If the variable  $s_3$  falls into the linear zone  $0 < s_3 < M_3$ , equation (10) reduces to

$$\dot{e}_2 = -(a+k_2)e_2 + e_3$$
,

where the parameter  $k_2 > 0$  determines the convergence of  $e_2$  to a given neighborhood of zero. According to the lemma, this neighborhood is described by  $|e_2| \le \Delta_3 / (a + k_2)$  under the condition  $|e_3| \le \Delta_3 = \text{const}$ .

Falling into the linear zone depends on the amplitude  $M_3$  due to the relation  $s_3\dot{s}_3 < 0$  holding in the nonlinear zone  $s_3 \notin (0, M_3)$ ; see Fig. 3. We write the equation for the variable  $s_3$ :  $\dot{s}_3 = -(as_3 - \overline{\xi}_2) + \dot{\overline{\xi}}_2 + k_2[-M_3 \text{sat}^+(s_3) + e_3 + \overline{\xi}_2]$ .

The parameters  $M_3$  and  $k_2$  of the function  $M_3 \text{sat}^+(s_3)$  are assigned from the following considerations.

• For  $s_3 \in [M_3, \infty]$  (which implies  $-M_3 \operatorname{sat}^+(s_3) = -M_3$ ), the inequality  $\dot{s}_3 = -a(s_3 - \overline{\xi}_2) + \dot{\overline{\xi}}_2 + k_2[-M_3 + e_3 + \overline{\xi}_2] < 0$  holds. Hence, the amplitude  $M_3$  should satisfy the constraint  $M_3 > \frac{1}{k_2}[a\overline{\xi}_2 + \dot{\overline{\xi}}_2] + e_3 + \overline{\xi}_2$ .

• For  $s_3 \in (-\infty, 0)$  (which implies  $M_3 \operatorname{sat}^+(s_3) = 0$ and  $M_3 \operatorname{sat}^+(s_3) = 0$ ), the inequality  $\dot{s}_3 = -a(s_3 - \overline{\xi}_2) + \dot{\overline{\xi}}_2 + k_2[e_3 + \overline{\xi}_2] > 0$  holds. Hence, under the conditions  $e_3 + \overline{\xi}_2 > 0$  and  $|e_3| \leq \Delta_3 = \operatorname{const}$ , the coefficient

 $k_2$  should satisfy the constraint  $k_2 > \frac{\left|a\overline{\xi}_2 + \dot{\overline{\xi}}_2\right|}{e_3 + \overline{\xi}_2}$ . Note that due to  $bf_1(x_2)x_1 + a(T_f(t) - T_d) > 0$ , the requirement  $\overline{\xi}_2 + e_3 = bf_1(x_2)x_1 + a(T_f(t) - T_d) - \dot{T}_d + e_3 > 0$ restricts the desired stabilization accuracy  $|e_3| \le \Delta_3$ (see the next step) and the rate of change of the given neighborhood:  $\Delta_3 < bf_1(x_2)x_1 + a(T_f(t) - T_d(t))$  and  $\left|\dot{T}_d(t)\right| < bf_1(x_2)x_1 + a(T_f(t) - T_d(t)) - \Delta_3$ . In the physical sense, the latter inequality limits the rate of change of the given neighborhood to an acceptable level when the tracking problem becomes solvable. Step 2. We ensure that the variable  $e_3$  from equation (10) falls into the neighborhood of zero:  $|e_3| \le \Delta_3$ . According to (9), the dynamics of the variable  $e_3$  are described by

$$\dot{e}_3 = \beta d(t) a_{21} g x_4 + \xi_3$$
. (11)

Treating the variable  $x_4$  in the system (11) as a fictitious control action, we make it equal to

$$e_4 = \beta d(t)a_{21}gx_4 + M_4 \operatorname{sat}(s_4),$$
 (12)

where  $M_4 \operatorname{sat}(s_4) = \min(M_4, |s_4|) \operatorname{sign}(s_4)$  (Fig. 2) and  $s_4 = k_3 e_3 + \xi_3$ . Stabilizing the variable  $e_4 \to 0$ , we consider the constraint  $|x_4| \le X_4$  by choosing an amplitude  $M_4 < \beta Da_{21}gX_4$ .

Equation (11) with the local feedback law (12) takes the form

$$\dot{e}_3 = e_4 - M_4 \text{sat}(s_4) + \xi_3$$
.

In the linear zone  $(|s_4| < M_4 \Longrightarrow M_4 \text{sat}(s_4) = s_4)$ , it reduces to

$$\dot{e}_3 = -k_3e_3 + e_4$$

The amplitude  $M_4$  under which the variable  $s_4$  falls into the linear zone is found using Lyapunov's second method.

We write the derivative of the function  $s_4$  as

$$\dot{s}_4 = k_3 [-M_4 \text{sat}(s_4) + \varphi_4(.)] + \xi_3,$$
  
 $\dot{\xi}_3 = 2\beta da_{21}mx_3\dot{x}_3 + \frac{d^2}{dt^2} [M_3 \text{sat}^+(s_3)]$  and

 $\varphi_4 = e_4 + \beta da_{21}mx_3^2 + \frac{d}{dt}[M_3 \text{sat}^+(s_3)].$ 

where

We choose a candidate Lyapunov function of the form  $V = 0.5s_4^2$ . Then the requirement  $\dot{V} = s_4\dot{s}_4 < 0$  outside the linear zone  $(|s_4| \ge M_4)$  yields the derivative  $\dot{V} = s_4 \{k_3[-M_4 \text{sign}(s_4) + \varphi_4(.)] + \dot{\xi}_3\} < 0$ . Hence, the amplitude should be assigned from the condition  $M_4 > \Phi_4 + \bar{E}_3 / k_3$ , where  $|\dot{\xi}_3| \le \bar{E}_3 = \text{const}$  and  $|\varphi_4| \le \Phi_4 = \text{const}$ .

Step 3. The last step is to stabilize the variable (12) described by

$$\dot{e}_4 = \beta da_{21}ga_{32}u_2 + \xi_4, \tag{13}$$

where  $\xi_4 = \beta \dot{d}(t) a_{21} g x_4 - \beta d(t) a_{21} g a_{32} (g x_3 + a_{31} x_4) + \frac{d}{dt} [M_4 \text{sat}(s_4)].$ 

We choose the discontinuous control

$$u_2 = M_2 \operatorname{sign}(e_4) \,. \tag{14}$$







In the system (13), a sliding mode occurs on the plane  $e_4 = 0$  in a finite time under the existence con-

dition 
$$M_2 > \left| \frac{\xi_4}{\beta da_{21}ga_{32}} \right|.$$

For clarity, we write the dynamic equation of the system (5) in this sliding mode:

$$\dot{x}_1 = -ax_1 - f_1(T_d)x_1 + \xi_1(t),$$
  
$$\dot{e}_2 = -(a+k_2)e_2 + e_3, \dot{e}_3 = -k_3e_3, e_4 = 0.$$
 (15)

In the system (15), the variables  $e_3$  and  $e_2$  asymptotically vanish:  $e_4 = 0 \Longrightarrow e_3 \rightarrow 0 \Longrightarrow e_2 \rightarrow 0$ . At the same time, the first subsystem is an equation of zero dynamics. Due to  $f_1(T_d) > 0$ , this equation is stable.

Thus, the discontinuous control (14) ensures a sliding motion on the plane  $e_4 = 0$  in a finite time, described by the stable system (15) of linear differential equations.

Note that if the subsystem (13) is stabilized using the continuous feedback law

$$\beta da_{21}ga_{32}u_2 + \xi_4 = -k_4e_4 \Longrightarrow u_2 = -\frac{\xi_4 + k_4e_4}{\beta da_{21}ga_{32}}$$

then the last equation of the system (15) will take the form  $\dot{e}_4 = -k_4 e_4$ . Hence, the closed loop system will be stable. Among additional requirements for implementing this continuous control, we mention information about the variable  $\xi_4$  and a pulse-width modulation device to control the DC motor voltage inverter. For implementing the discontinuous control (14), we need only an upper bound on this variable:  $|\xi_4| \le E_4 = \text{const.}$ 

The first subsystem of (15), treated as an equation of zero dynamics, has a bounded solution:  $|x_1| \leq \frac{E_1}{a + f_1(T_d)}$  for  $|\xi_1(t)| = aC_{Af} \leq E_1 = \text{const}$ ; see the lemma. Particularly for  $\xi_1 = E_1 = \text{const}$  and  $x_2 = T_d = \text{const}$ , the lemma implies  $x_1 \rightarrow \frac{qC_{Af}}{q + Vf_1(T_d)}$ . (Recall that  $a = \frac{q}{V}$ .) Choosing the parameter q as the control action in the product concentration loop, we have the following limit relations:  $q \rightarrow \infty \Rightarrow$  $C_A \rightarrow C_{Af}$  and  $q \rightarrow 0 \Rightarrow C_A \rightarrow 0$ . Therefore, we can maintain the product concentration within a reasonable range  $C_A \in [C_{A1}, C_{A2}]$  by tuning the reagent flow rate into the reactor.

Clearly, increasing (decreasing) the reagent flow rate into the reactor, we decrease (increase, respectively) the product concentration. Consider three sets of parameters in which the values  $C_{Af} = 0.9$  and q = 0.9 are fixed, whereas the desired temperature varies: 1)  $T_d = 350, 2$ )  $T_d = 380, 3$ )  $T_d = 400$ . According to the relation  $x_1 \rightarrow \frac{qC_{Af}}{q + Vf_1(T_d)}$ , we obtain: 1)  $x_1 \rightarrow 0.3251, 2$ )  $x_1 \rightarrow 0.3214, 3$ )  $x_1 \rightarrow 0.3192$ .

Thus, the product concentration decreases as the temperature in the reactor increases, and this conclusion agrees with the simulation results; see Fig. 10 in Section 4.

#### **4. SIMULATION RESULTS**

The effectiveness of the proposed approach was verified by numerical simulations of the CSTR–DC motor system in MATLAB. The parameters for the system (3), observer (6), and control (14) were selected from Table 2.

Table 2

Group of	Parameter values		
parameters			
CSTR parameters	$q = 0.9, V = 1, \beta = 0.003, \gamma = 80, b = 5,$		
	$k_0 = 2$ , $T_{c0} = 300$ ,		
	$C_{Af} = 0.9 + 0.005 \sin(0.03\pi t),$		
	$T_f = 395 + 0.01 \sin(0.05\pi t)$ .		
DC motor parameters	$a_{21} = 0.8, g = 0.7, m = 0.0001,$		
1	$a_{31} = 12.5, a_{32} = 2$		
Initial conditions	$x_1(0) = 0.3, x_2(0) = 400, x_3(0) = 100,$		
and reference	$x_4(0) = 20$		
Simulation	$T_d(0) = 350 \text{ for } t \in [0, t_1],$		
scenario	$T_d(t_1) = 380$ for $t \in (t_1, \infty], t_1 = 75$		
Observer	$z_i(0) = 0, \ i = \overline{1, 4} ,$		
parameters	$l_1 = 100, \ l_2 = 3, \ l_3 = 100, \ l_4 = 500$		
Controller	$M_2 = 400, M_3 = 50, M_4 = 170,$		
parameters	$k_2 = 0.5, k_3 = 0.1$		
Physical	$C_A \in [0,1], T \in [0,400], x_3 \in [0,100],$		
constraints on state variables	$ x_4  < 150, u_2 = \pm 400$		

Model parameters

The temperature in the reactor jacket,  $x_2$ , the armature current  $x_4$ , the reagent flow rate q, and its temperature  $T_f$  were assumed measurable in the plant. The state observer (6) was constructed to obtain complete information about the state variables of the CSTR-DC motor system and the exogenous disturbances (to estimate the unknown signals with a given accuracy).



The observation errors  $\varepsilon_i = x_i - z_i$ ,  $i = \overline{1, 4}$ , are presented in Figs. 4–7.

Figure 8 shows the graph of the exogenous disturbance  $\xi_1 = 0.9(0.9 + 0.005\sin(0.03\pi t))$  reconstructed using the observer (6). The dashed line corre-



Fig. 4. The observation error  $\varepsilon_1(t)$ .



Fig. 5. The observation error  $\varepsilon_2(t)$ .



Fig. 6. The observation error  $\varepsilon_3(t)$ .

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sponds to the real disturbance values and the solid line to the restored ones. Figure 9 shows the graph of temperature variations in the reactor jacket.



Fig. 7. The observation error  $\varepsilon_4(t)$ .



Fig. 8. The exogenous disturbance  $\xi_1(t)$ .



Fig. 9. Temperature in the jacket,  $x_2(t)$ .

Figures 10–13 show the graphs of the product concentration  $x_1(t)$ , the motor shaft rotation frequency  $x_3(t)$ , the armature current  $x_4(t)$ , and the armature voltage  $u_2(t)$ , respectively.



Fig. 10. The product concentration  $x_1(t)$ .



Fig. 11. The motor shaft rotation frequency  $x_3(t)$ .



Fig. 12. The armature current  $x_4(t)$ .



Fig. 13. The armature voltage  $u_2(t)$ .

#### CONCLUSIONS

The mathematical model of a continuous stirred tank reactor has been extended by introducing the dynamics of an actuator (DC motor) to apply the theory of sliding modes by (de)activating the switches of the voltage inverter.

An observer with mixed corrections (discontinuous and continuous) has been designed to obtain information about the unmeasured state variables and disturbances.

In the proposed feedback law design procedure, an observer estimates the real signals with a given accuracy both in a real sliding mode and when using deep feedback. A key feature of this work is the feedback law designed within the block approach to consider state-space and control constraints for the CSTR.

The effectiveness of this control algorithm has been validated analytically and illustrated numerically by simulations in MATLAB.

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This paper was recommended for publication by V.N. Afanas'ev, a member of the Editorial Board.

Received May 24, 2021, and revised July 28, 2021. Accepted August 24, 2021.

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#### Cite this article

Gulyukina, S.I., Utkin, V.A. A Block Approach to CSTR Control under Uncertainty, State-Space and Control Constraints. *Control Sciences* **5**, 43–52 (2021). http://doi.org/10.25728/cs.2021.5.4

Original Russian Text © Gulyukina, S.I., Utkin, V.A., 2021, published in *Problemy Upravleniya*, 2021, no. 5, pp. 48–59.

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Control in Medical and Biological Systems

DOI: http://doi.org/10.25728/cs.2021.5.5

# ESTIMATING THE INFLUENCE OF ENVIRONMENTAL FACTORS ON MORTALITY IN ELDER AGE GROUPS: AN EXAMPLE OF KRASNOYARSK<sup>1</sup>

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**Abstract.** This paper analyzes and forecasts the influence of climatic parameters and air quality on the mortality of the Krasnoyarsk industrial agglomeration. The long-term climatic data for the city of Krasnoyarsk are studied. The diseases sensitive to climate change are considered for the period from 2011 to 2014. The relationship between climatic parameters, air pollution, and the number of deaths is established using multivariate statistical analysis. The factors with the greatest contribution to mortality for climate-dependent diseases are identified. The mortality causes associated with negative environmental factors are examined. The age groups most affected by climatic parameters and air pollution are determined. As shown below, the nonlinear Poisson regression model predicts population mortality quite close to the factual data.

Keywords: air quality, climatic parameters, generalized linear model with Poisson regression, population mortality, social-natural-technogenic (S-N-T) system.

## INTRODUCTION

The joint influence of anthropogenic activity and natural factors on the population's health increases the sustainable development risks of territorial entities. In addition, the urbanization trends of the environment create new types of technological and natural threats. Due to the economic and social conditions emerging in Russia, the main object of urbanization in the nearest future will be the territories of Siberia and the Arctic. Considering the uniqueness of natural systems and the global significance of these regions for the country's sustainable development, a particularly important scientific problem is studying natural and technological threats and development risks using new-generation

<sup>1</sup>The reported study was funded by Russian Foundation for Basic Research, Government of Krasnoyarsk Territory, Krasnoyarsk Regional Fund of Science, project number 19-413-240013 "Risk assessment methodology caused by environmental factors on population health and mortality in industrial agglomerations".

information technologies. Sustainable development of territories is associated with a quantitative assessment of complex security and regional management using a risk-based approach [1–4].

The territory of an industrial region is a single complex social-natural-technogenic (S-N-T) system. The elements of this system mutually influence each other and have interconnected types of risks [5]. The operation of the sociosphere as an element of the S-N-T system is associated with individual strategic risks of life and health loss. The most significant factors forming individual strategic risks include environmental pollution (primarily air pollution) and the globally changing climate. Specific meteorological phenomena affecting the health of the population are short-term; nevertheless, their frequency is determined by the climatic characteristics of the territory. Therefore, this formulation involves the terms "meteorological conditions" and "climatic parameters."

A relationship between mortality rates and climatic factors (temperature waves) was established in several studies carried out for the cities of Arkhangelsk, Yakutsk, Astrakhan, Krasnoyarsk, Moscow, and others





[6–11]. For the city of Krasnoyarsk, the relative risk of mortality from the negative influence of cold and heat waves was assessed using the Poisson regression model; for details, see [7].

According to the 1999–2008 data, periods of intense heat in Arkhangelsk (a temperature threshold of 21°C) were associated with an increase in mortality from cardiovascular diseases and all-natural causes among the over-65s (the people aged 65 and elder) and from all external causes among the over-30s. During that period, 110 additional deaths were recorded due to heat waves and 179 additional deaths due to cold waves [6].

Temperature waves were first identified and studied in 1881–1884 and were described as an area of high or low temperatures of atmospheric air not staying in one place for a long time [12]. Temperature waves are associated with serious environmental and health problems and may even incur significant economic damage. High temperatures can lead to thermal stress and deteriorate air quality, causing adverse health effects, especially for more vulnerable groups of population (including those with cardiovascular or respiratory diseases and the elderly) [13].

Extremely low and high temperatures in winter and summer, respectively, are characteristic of the climate of Central Siberia, Yakutia, and other regions. They are associated with sharply continental weather. With these unique climatic characteristics of industrial agglomerations, there is a need to analyze the synergistic effect of environmental factors on mortality. This problem can be partially solved by developing a methodology for assessing the risks of a multifactorial influence on the population's mortality using statistical regression models and elaborating measures to protect the population from the negative influence of environmental factors. The methodology should consider the structural characteristics of the population and the region.

Methods for assessing public health risks have been developed for three decades. The negative influence of high-level air pollution on the population's morbidity and mortality has been demonstrated by many studies. Children and adolescents are most susceptible to the influence of chemical pollution of the atmospheric air [14-16]. The child's respiratory system is a prime target for air pollutants. They cause a wide range of acute and chronic effects, either as a single risk factor or, most often, in combination with other external agents and (or) characteristics of the child's susceptibility. Age plays an important role during exposure to inhaled pollutants [17-19]. Infants are more susceptible to lung damage from the same-type toxins than adults, even at doses below the latter's optimal value [20, 21].

According to the population data for the city of Krasnoyarsk for the period 2000–2018, the total mortality rate decreased until 2009 (by 17%) and increased until 2018 (by 9%) with a population growth of more than 20%; see Fig. 1. (The data are provided by the Krasnoyarsk Regional Body of the Federal State Statistics Service.)

From 2006 to 2015, mortality from cerebrovascular diseases decreased by 9.9% (Fig. 2). By contrast, the proportion of deaths from ischemic heart disease increased from 18% in 2000 to 30% in 2018.

The Poisson regression model is recommended to identify the relationship between mortality and environmental factors [22]. In foreign epidemiological studies, the application of the Poisson model goes back to the second half of the 20th century. This model was



Fig. 1. Dynamics of mortality and average annual number of population: city of Krasnoyarsk and Krasnoyarsk Region.



Fig. 2. Dynamics of mortality from ischemic heart and cerebrovascular diseases: city of Krasnoyarsk.

adopted to study the relationship between mortality and different environmental factors: air pollution [23], average temperature [24], and temperature waves [25– 27].

Some studies identified the relationship between cancer incidence and polluted air and bad habits [28, 29]. The research where this model was applied to Moscow included three influence factors: the average daily air temperature, ozone concentration, and the concentration of suspended particles  $PM_{10}$ . Higher mortality from all causes, except for external ones (accidents, conditions, and circumstances due to an injury, poisoning, and other adverse effects) was discovered under increasing the number of days with abnormally low and high temperatures [30].

This paper assesses the relative risks of the sociosphere of the S-N-T system for the city of Krasnoyarsk. We consider the characteristics of climatic parameters and levels of air pollution as influence factors. We predict the risk of population's premature mortality from the combined influence of climatic and environmental factors. The main problem includes estimating the relationship between climatic parameters, air pollution, and population mortality using multivariate statistical analysis methods. In addition, we analyze the significance of the contribution of each factor mentioned to mortality rates for the main groups of climate-dependent diseases.

### **1. DATA AND METHODS**

We studied mortality rates in the city of Krasnoyarsk from **the initial causes** with the greatest sensitivity to climatic factors, according to the International Classification of Diseases (ICD-10) [22] in two age groups (60–74 years old and the over-75s), divided by sex for the period from 2011 to 2014:

- ischemic heart disease (I20-I25),

- cerebrovascular diseases (I60–I69),

- respiratory diseases (J00–J22, J30, J40–J44, and J45).

The climatic characteristics were determined and estimated using data from state observation networks of meteorological stations [31, 32]. To normalize the time series, we divided them by the standard deviation [33]. Missed values of these factors were restored using the moving average method.

For constructing a nonlinear Poisson regression model describing the dependence of mortality on environmental factors, we adopted the daily mortality data for the city of Krasnoyarsk for the period from 2011 to 2014. (The data are provided by the Krasnoyarsk Regional Body of the Federal State Statistics Service [22].) The model has the form

$$\log (\mu_t) = \beta_0 + \beta_1 X_{1,t-L} + \dots + \beta_k X_{k,t-L} + \beta_{k+1} Z,$$
  
$$\mu_t = e^{\beta_0} e^{\beta_1 X_{1,t-L}} \dots e^{\beta_k X_{k,t-L}} e^{\beta_k Z},$$

where  $\mu_t$  is the predicted parameter, i.e., the Poisson independent (output) variable;  $X_{1,t}, \dots, X_{k,t}$  are the explanatory (input) variables; *Z* is the short-term time trend by day of the week; *L* denotes the negative effect delay (lag);  $\beta_1, \dots, \beta_{k+1}$  are the model parameters;  $\beta_0$  is the free term.

Epidemiological studies determine the relationship between the concentration of air pollutants in one day and the health effects on some lag days. Lag means time delay: when exposed to a negative phenomenon, its effect may appear only after some period [22]. In this study, we chose a lag of 1 to 15 days.

For interpreting the regression coefficient, we used *the relative risk (RR)*. This indicator measures the relationship between an independent variable (e.g., the concentration of air pollutants) and the risk of a particular result (e.g., the number of people with respiratory injury). It shows how many times the population's mortality increases due to a negative environmental factor relative to background mortality (mortality of people not exposed to the negative influence). For the Poisson regression, the relative risk is defined as follows [13, 34]:

### $RR = \exp(\beta_i),$

where  $\beta_i$  denotes the regression coefficient.

Table 1 presents the dynamics of the studied environmental factors from 2000 to 2018. For the city of Krasnoyarsk, the daily concentration data on air pollutants are provided by the Central Siberian Department for Hydrometeorology and Environmental Monitoring.

The pollutants were chosen based on the evidence of their influence on the morbidity and mortality rates of the population; see Table 2. Ozone  $(O_3)$  is one of the most dangerous air pollutants. This substance belongs to the first hazard class and represents the main component of photochemical smog. An increased level of ground-level ozone is observed only in hot sunny weather. The inhalation effect of the substance on the body is accompanied by irritation of the respiratory system, a reduction in lung functions, the development of asthma and allergies, and a significant decrease in immunity to infections [22, 35-37]. Ozone concentrations are not measured within the state observation network of the Krasnovarsk Region. Therefore, we used the concentrations of nitrogen dioxide and formaldehyde as precursors of ozone.

When processing the data and performing a numerical experiment, we analyzed the collinearity or multicollinearity of the factors and correct short-term time trends (by days of the week) by adding another coefficient into the model.

Table 1

		Hot v	t waves Cold waves				%	Extreme temperature drops during the day		Average concentration of pollutants						
Year	Number of waves	buration of waves, days	Amplitude of	waves, °C	Number of waves	Duration of waves, days	Amplitude of	wāves, °C	Temperature, °C	Relative humidity, <sup>6</sup>	Number of events		Nalige, C	Suspended ubstances, mg/m <sup>3</sup>	Nitrogen dioxide (NO <sub>2</sub> ), mg/m <sup>3</sup>	ormaldehyde (F), mg/m <sup>3</sup>
	, ,	П	min	max			min	max			~	min	max	S		Ц
2000	6	37	7.2	20.2	5	52	-38	-14	0.8	73	49	10	16.6	0.254	0.024	_
2003	4	29	6.4	20.6	3	30	-14	-10	2.5	71	73	10	31	0.265	0.042	0.008
2006	2	24	13.6	20.4	4	49	-24	-10	0.8	71	15	10.1	20	0.198	0.051	0.0043
2009	3	19	8.3	21	9	82	-24	-7.7	0.2	73	15	10.1	22.2	0.173	0.062	0.0094
2012	4	31	8.3	21.1	4	78	-21	-9	0.6	71	14	10.1	30.3	0.228	0.051	0.0185
2015	7	42	7.5	20.8	1	14	-11	-11	3.8	66	16	10	13.7	0.1380	0.0360	0.0120
2018	4	40	7.7	20.7	8	60	-20	-11	1.3	70	26	10.1	19.4	0.1063	0.0379	0.0168

## **Dynamics of environmental factors**

Table 2

## The effect of exposure to pollutants on the body [38]

Substance	Hazard class	Critical organ	Critical effect
Suspended substances	3	Respiratory system	<ul> <li>An increase in total mortality,</li> <li>Mortality from diseases of the cardiovascular system and respiratory system,</li> <li>Frequent symptoms from the upper and lower respiratory tract,</li> <li>Consulting a doctor for respiratory diseases,</li> <li>Frequent exacerbation of bronchial asthma</li> </ul>
Ozone (O <sub>3</sub> ) [22, 35]	1	Respiratory system	<ul> <li>An increase in total mortality,</li> <li>Irritation of the respiratory system,</li> <li>Decreased lung function,</li> <li>Development of asthma and allergies,</li> <li>A significant decrease in immunity to infections</li> </ul>
Nitrogen diox- ide (NO <sub>2</sub> )	2	Respiratory system, blood and hemopoi- etic organs	<ul> <li>Increased incidence and duration of diseases of the upper and lower respiratory tract,</li> <li>An increased number of lower respiratory tract diseases among children,</li> </ul>
Formaldehyde (F)	2	Respiratory system, organs of vision, and immune sys- tem	<ul> <li>An increased incidence of diseases of the upper and lower respiratory tract,</li> <li>Inflammatory processes in the lungs,</li> <li>Diseases of the immune system, including allergic reactions,</li> <li>Diseases of the organs of vision</li> </ul>



## 2. RESULTS OF THE STUDY

For a particular factor, the level of influence on the mortality rate was determined using the regression coefficient. It shows by how many units the result will change when a given factor changes by one unit [39].

Among females aged 60–74 years, mortality from respiratory diseases was most influenced by temperature waves on the ninth day of exposure (lag equaled 9). The influence of pollutants was represented by the effect of formaldehyde:

$$\ln(\mu_t) = -7.4 + 0.9F_t - 4.9Temp_{t-2} - 3.3Temp_{t-12} + 0.9Diff_{t-1} + 0.9Diff_{t-12} + 1.3Hum_{t-11} + 2.8W_{t-9},$$
(1)

where *F* is the concentration of formaldehyde in the atmospheric air,  $mg/m^3$ ; *Temp* denotes temperature, °C; *Diff* is the temperature drop, °C; *Hum* indicates the relative humidity, %; *W* gives the number of temperature waves.

Since the free term  $\beta_0$  has a small value compared to the total value of the factor weights, the factors not included in this model are insignificant. According to the expression (1), the same influence factors can negatively affect different lags. The influence of the average daily temperature manifested on the 2nd and 12th days after the inhalation exposure; the temperature drop during the day, on the 1st and 12th days. Figure 3 shows the distribution of factual mortality and mortality calculated using the Poisson model. Clearly, for an example of mortality from respiratory diseases among females aged 60–74, the calculated mortality rate describes well the factual one with a correlation coefficient of 0.8.

We processed the available statistical dataset for 2011 to 2014 for two age groups (from 60 to 74 years old and over 75 years old) to obtain 12 mortality prediction models for three main immediate causes belonging to the climate-dependent category. The relative risk of mortality from exposure to environmental factors, the lags, and the correlation indices for each model are presented in Table 3.

#### 3. DISCUSSION OF THE RESULTS

The analysis of the population mortality from the negative influence of environmental factors using the Poisson regression model has yielded the following outcomes.

• According to the lag distribution, for males, most of the negative effects from ischemic heart disease and cerebrovascular diseases manifest within six days after exposure to the combination of factors; from respiratory diseases, after the 7th day. In contrast, for females, negative health effects accumulate more smoothly within two weeks; see Fig. 4.



Fig. 3. Factual and calculated mortality from respiratory diseases in 2014: females aged 60-74 years.

#### Table 3

	Sex	Ischemic Cerebrovascular heart disease diseases		Respiratory system dis- eases	Ischemic heart disease	Cerebro- vascular diseases	Respiratory system dis- eases
Influence factors				Age gro	oup		
			60–74 years		75	b years and eld	ler
			Value of correlation	coefficient for f	factual and calcu	ilated mortality	у
	F	0.69	0.66	0.8	0.63	0.53	0.66
	М	0.6	0.63	0.6	0.55	0.64	0.66
			Relativ	ve risk of mortal	ity		
Nitana and dianida	F	1.2 / L = 5	1.3 / L = 15	-	_	_	1.3
Nitrogen dioxide	М	1.1 / L = 5	1.2 / L = 13	1.4	_	_	1.4 / L = 14
Earmaldaharda	F	1.2 / L = 9	_	2.5	_	1.2	1.4 / L = 11
Formaldenyde	М	—	1.5 / L = 4	_	0.7 / L = 14	0.7 / L = 2	1.6 / L = 3
Suspended sub-	F	1.2 / L = 5	1.3 / L = 8	_	—	0.8	_
stances	М	1.2 / L = 15	1.3 / L = 5	-	1.2 / L = 3	—	1.4 / L = 8
	F	0.4 / L = 11	—	0.04 / L = 12	0.7 / L = 4	_	3.7 / L = 5
Temperature	М	0.6 / L = 15	_	_	_	0.4 / L = 12	3.3 / L = 10
Dalating hours dites	F	1.3 / L = 6	_	3.6 / L = 11	0.8 / L = 11		1.6 / L = 2
Relative numicity	М	1.1	_	_	1.2 / L = 5	1.5 / L = 4	_
	F	_	_	2.5 / L = 12	_	_	_
Extreme tempera- ture drops	М	_	_	1.4 / L = 9	_	0.8 / <i>L</i> = 14	1.4 / L = 1
T	F	1.8 / L = 2	_	16.4 / L = 9	1.5 / L = 13	—	0.5 / L = 2
Temperature waves	М	—	0.4 / L = 10	—	—	_	_

The relative risk of population mortality





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• The climatic factors have the greatest contribution to population mortality in all age and sex groups: the ratio of influence is 58.1% (climatic factors) to 41.9% (polluted atmospheric air). In addition, females are most exposed to the negative effects of environmental factors.

• Among the pollutants considered, formaldehyde has the greatest relative contribution to mortality. Among the climatic factors considered, the greatest harm is caused by temperature waves.

• The highest correlation coefficients between the predicted and factual mortality rates correspond to mortality from respiratory diseases.

• Among the age groups considered, the people aged 60–74 years are most sensitive to the effect of all factors.

#### CONCLUSIONS

As demonstrated in the paper, the nonlinear Poisson regression model predicts population mortality from environmental factors quite close to the factual data. Hence, this approach can be used to estimate the combined influence of climatic parameters and air quality on mortality. The approach to risk assessment will be further developed by excluding insignificant factors and considering in detail the influence of significant factors by groups, depending on the direction of the identified relationships.

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This paper was recommended for publication by A.I. Mikhalsky, a member of the Editorial Board.

> Received June 17, 2021, and revised July 19, 2021. Accepted August 24, 2021.

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#### Cite this article

Taseiko, O.V., Chernykh, D.A. Assessing the Impact of Environmental Factors on Mortality in Elder Age Groups: An Example of Krasnoyarsk, *Control Sciences* **5**, 53–60 (2021). http://doi.org/10.25728/cs.2021.5.5

Original Russian Text © Taseiko, O.V., Chernykh, D.A., 2021, published in *Problemy Upravleniya*, 2021, no. 5, pp. 60–69.

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DOI: http://doi.org/10.25728/cs.2021.5.6

# NON-BLOCKING FAULT-TOLERANT DUAL PHOTON SWITCHES WITH HIGH SCALABILITY

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Abstract. This paper continues the construction of a fundamentally new class of system area networks (dual photon networks) with the following features: non-blocking property and static self-routing, high scalability with the maximum achievable speed and a small complexity compared to a full switch, and balancing the scalability-speed and complexity-speed ratios. These networks are implemented in an extended circuit basis consisting of dual photon switches and separate photon multiplexers and demultiplexers. We propose a method for constructing a fault-tolerant dual network with the indicated properties based on networks with the quasi-complete graph and quasi-complete digraph topologies and the invariant extension method with internal parallelization. Also, we propose a method for extending the two-stage dual network designed previously into four-stage and eight-stage dual networks with high scalability while maintaining the original network period and reducing its exponential complexity.

**Keywords:** photon switch, dual switch, photon multiplexers and demultiplexers, multistage switch, conflict-free self-routing, non-blocking switch, static self-routing, quasi-complete digraph, quasi-complete graph, invariant extension of networks, switching properties, direct channels, scalability and speed.

## INTRODUCTION

This paper develops a method for constructing a fundamentally new class of system area networks [1-6], the so-called dual photon networks. They are non-blocking networks with static self-routing [1-3, 5] and can have a given degree of channel fault tolerance [6].

In what follows, we propose a method for constructing non-blocking self-routing photon networks with high scalability. These are dual networks based on a non-blocking dual  $p \times p$  switch with a signal period of p cycles [1–3]. For resolving signal conflicts, the dual switch combines the bus method (separation of conflicting signals to different cycles in one channel) and the switch method (separation of conflicting signals to different channels). The dual switch is a nonblocking switch on any input traffic if data bits are transmitted with a signal period of p cycles. The dual switch was developed by the author's colleagues [1, 2] and then applied and named in the joint publications [3–5]. This switch turned out to be a prerequisite for constructing non-blocking networks with high scalability and acceptable complexity.

The dual photon switch transmits signal and control information in parallel at different frequencies for each data bit. This method eliminates the problem of synchronizing signals from different channels.

The photon specifics of such networks consist in using in-bit channel virtualization with feedback links through delay lines with a duration of one cycle and control signals at different frequencies to route individual bits. The separation of information signals to different cycles is accompanied by the separation of the corresponding control information to the same cycles. It is used to route the bits by moving them between different channels without changing the established cycle numbers.

Throughout the paper, the terms "dual switch" and "dual network based on dual switches" imply using bits with a period of p cycles in them. These bits en-



sure the non-blocking property in the first stage of the network; in the other stages, they remain by "inertia" without using the bus method of conflict resolution.

The scalability of dual networks is provided using networks with the quasi-complete graph or digraph topology [4], which are implemented in an extended circuit basis consisting of dual photon switches and separate photon multiplexers and demultiplexers. In [1-3, 5], high scalability was achieved using the invariant extension method of system area networks with many additional demultiplexers and multiplexers.

In the papers [5, 6], a new method for extending dual networks by their internal parallelization without additional devices was developed and applied for the first time. Particularly in [5], a two-stage non-blocking network was constructed. It consists of networks with the quasi-complete digraph topology with  $N = p^2$  channels on each stage, whereas the two-stage non-blocking network has  $N^2$  channels in total. On the other hand, a two-stage non-blocking network with the quasi-complete digraph topology with  $(\sigma - 1)$ -channel fault tolerance was constructed in [6]. It consists of networks with the quasi-complete graph topology with  $N = p(p-1)/(\sigma + 1)$  channels, whereas the two-stage fault-tolerant non-blocking network has  $N^2$  channels.

Below, we construct four-stage and eight-stage fault-tolerant networks by developing and applying the generalized method of internal parallelization. In this case, the same degree of network scalability is achieved as in the invariant method, but without using external demultiplexers and multiplexers, and the resulting networks have a significantly smaller complexity.

This paper is organized as follows. Section 1 briefly considers the non-blocking property and channel fault tolerance in modern system area networks. In Section 2, following [6], we introduce the notions of p-permutations (crucial for proving the non-blocking property of four- and eight-stage networks) and repeat the proofs of the non-blocking property for two-stage networks. Section 3 presents four-stage non-blocking self-routing switches with one-channel and twochannel fault tolerance and their performance characteristics. The method of internal parallelization from [5, 6] is generalized for four-stage switches.

Section 4 compares the performance characteristics of four-stage non-blocking self-routing switches based on switches with the dual quasi-complete graph and digraph topologies. Finally, in Section 5, we construct eight-stage non-blocking self-routing switches based on switches with the dual quasi-complete graph and digraph topologies. Moreover, the method from Section 3 is generalized for eight-stage switches. Section 6 discusses the properties of these networks compared to other non-blocking networks (particularly their disadvantages and possible ways to overcome them).

In the Conclusions, we analyze the generalized method of internal parallelization, which is the core for constructing dual non-blocking networks with high scalability and low specific complexity. There are three main components of the proposed methodology: a non-blocking dual switch, a switch with the quasicomplete graph or digraph topology based on a dual switch, and the method of internal parallelization.

# 1. NON-BLOCKING PROPERTY AND FAULT TOLERANCE IN SYSTEM AREA NETWORKS

The problem of constructing non-blocking faulttolerant system area networks of supercomputers has not been completed solved so far.

A system area network is non-blocking if for any packet permutation, conflict-free paths from sources to sinks can be built in it. A system area network is selfrouting if conflict-free paths can be built locally over network nodes without their interaction based on routing information in packets only. Finally, self-routing is static if any source can independently choose conflictfree paths to its sink without interacting with other sources.

The existence of non-blocking networks was proved by Clos [7, 8]. Self-routing procedures for nonblocking Clos networks have not yet been developed. However, these networks can be a qualitative measure for other non-blocking networks.

A network in the form of a two-dimensional generalized hypercube with the quasi-complete digraph topology is non-blocking, e.g., in the YARK and RO-SETTA switches used in several networks of different structure: a reconfigurable Clos network [9], a threedimensional torus [10], and a hierarchy of complete and quasi-complete digraphs [11–13]. Unfortunately, a quasi-complete digraph has a small number of channels  $N = p^2$ , where *p* is the degree of internal switches, and a high switching complexity  $S \ge N^2$ , exceeding the complexities of a complete digraph and a nonblocking Clos network (in the latter case, considerably). Modern literature widely describes system area networks with the fat tree structure (particularly reconfigurable Clos networks), the generalized hypercube structure, the multidimensional torus structure, and system area networks with a hierarchy of complete and quasi-complete digraphs.

Fat-tree networks are reconfigurable networks [9, 14, 15] with conflict-free transmission only according to predetermined schedules for specific packet permutations. In the case of arbitrary permutations, these networks turn out to be blocking; permutations in them are implemented in several jumps between network nodes. The maximum number of such jumps determines the network diameter. In reconfigurable Close networks, the diameter equals the number of network stages.

Networks with the generalized hypercube structure [16–19] are not even reconfigurable [20, 21]. They can be made such by increasing the number of channels in some dimensions. Generalized cubes have a diameter equal to the number of dimensions or less by 1 in the extended hypercube [17, 18]. Generalized hypercubes with a doubled number of channels in each dimension are reconfigurable networks for two permutations simultaneously. Note that an attempt to use a generalized hypercube as a non-blocking network for a photon computer [22] seems to be a very dubious venture.

For arbitrary permutations, networks with the multidimensional torus structure cannot transmit packets over direct channels [11, 23–25]. They implement permutations in several jumps between network nodes. Multidimensional tori are the simplest, albeit slowest, networks due to their large diameters. For example, the networks considered in [11, 23–25] have diameters measured in tens of jumps.

On the contrary, networks with a hierarchy of complete or quasi-complete digraphs [10, 12, 26] have the smallest diameter of three jumps. Many networks with small diameters have appeared recently [27–32]. All of them have serious problems with balancing network load under channel faults.

Channel fault tolerance is the network's ability to preserve full availability under channel failures while maintaining its original performance characteristics (the non-blocking property, transmission delays, or network diameter).

Apparently, only networks with the quasi-complete graph topology possess channel fault-tolerance in pure form. These networks are isomorphic to such a mathematical object as an incomplete balanced symmetric block design [33]. These networks have an element base of  $p \times p$  switches,  $1 \times p$  demultiplexers, and  $p \times 1$ multiplexers and are non-blocking networks with static self-routing. They have direct channels between  $N = p(p-1)/\sigma + 1$  network users and  $\sigma$  different channels between any two users [4].

In other networks, the restoration of full network availability under channel faults is accompanied in one way or another by an increase in network transmission delays. For example, under channel faults in a reconfigurable Clos network, the load on the remaining channels grows, increasing the number of conflicts and delays in the transmission of some packets.

The TOFY network with the three-dimensional torus structure [25] uses three more dimensions to create redundant channels. If some network rings fail, their integrity is restored by increasing the network diameter by 1.

Generalized hypercubes with doubled channels in each dimension are one-fault-tolerant networks with a constant diameter [19].

In networks with a hierarchy of complete or quasicomplete digraphs [10, 12, 26], if some of the channels fail, the network's full availability is restored using bypass paths with a duration of five jumps. In other words, the transmission delay increases by a factor of 5/3.

# 2. DUAL QUASI-COMPLETE GRAPH, &PERMUTATIONS, AND DUAL TWO-STAGE SWITCH

The dual switch SFN<sub>1</sub> with the quasi-complete graph topology, SQG( $N_1$ , p,  $\sigma$ ), consists of  $N_1 = p(p - 1)/\sigma + 1$  dual  $p \times p$  switches SSp,  $N_1$  input  $1 \times p$  demultiplexers, and  $N_1$  output  $p \times 1$  multiplexers [6]. They are interconnected using the combinatorial method [4]: there are  $\sigma$  different paths through different dual switches SSp. Figure 1 shows the circuit of an SF4 switch as an SQG (4, 3, 2) graph with one-channel fault tolerance. Two paths are highlighted, connecting two randomly selected inputs and outputs, (2, 4) and (3, 3).

Any dual switch  $SFN_1$  has the same signal period  $T_1$  as the dual switch SSp included in it. For the switch  $SFN_1$ , the following performance characteristics are calculated: the switching complexity  $S_1$ , expressed in the number of switching points, and the channel complexity, expressed in the number of channels. They are







written exponentially through the number of channels and are called exponential complexities<sup>1</sup>; see Table 1.



Fig. 1. Dual quasi-complete switch SF4 with signal period of three cycles represented by graph SQG(4, 3, 2): (a) original form with duplex channels, (b) application with simplex channels. Dashed lines and dots indicate different paths between selected inputs and outputs.

Table 1

Performance characteristics of dual switches SF*N*<sub>1</sub> with one-channel fault tolerance

р	$N_1$	$T_1$	$S_1$	$L_1$
2	2	2	$24 = N_1^{4.58}$	$8 = N_1^3$
4	7	4	$280 = N_1^{2.9}$	$56 = N_1^{2.07}$
6	15	6	$1\ 260 = N_1^{2.64}$	$180 = N_1^{1.92}$
8	27	8	$3\ 888 = N_1^{2.51}$	$432 = N_1^{1.84}$

For the dual switch  $SFN_1$  with the quasi-complete *digraph* topology [5], the performance characteristics are given in Table 2. They are better than their counterparts from Table 1 but without channel fault tolerance.

Table 2

Performance characteristics of dual switches SF N<sub>1</sub> based on the quasi-complete digraph topology

р	$N_1$	$T_1$	$S_1$	$L_1$
2	4	2	$48 = N_1^{2.79}$	$16 = N_1^2$
4	16	4	$640 = N_1^{2.33}$	$128 = N_1^{1.75}$
6	36	6	$3\ 024 = N_1^{2.24}$	$432 = N_1^{1.69}$
8	64	8	$9\ 216 = N_1^{2.19}$	$1\ 024 = N_1^{1.67}$

We introduce the notion of p-partitions of packets transmitted through some cross-section of a network at the multiplexer inputs. All packets are divided into groups of variable composition, each containing at most p packets. For a common permutation of packets, a 1-partition occurs at the input and output of the switch. A transmission in which a 1-partition occurs at the network input and a p-partition on a given crosssection will be called a p-permutation.

For the dual switch  $SQG(N_1, p, \sigma)$ , this crosssection is through the inputs of the output multiplexers and is called the output cross-section. In Fig. 1, the output cross-section is indicated by the vertical dashed line. According to the property of the dual switch DS*p*, a *p*-partition occurs on the output cross-section of the dual switch  $SQG(N_1, p, \sigma)$  for any traffic.

**Lemma 1** [6]. *The dual switch*  $SFN_1$  *with a signal period of p cycles is a non-blocking switch with static self-routing under any common permutation and has* ( $\sigma$  – 1)*-channel fault-tolerance.* 

The papers [5, 6] developed a method for constructing two-stage non-blocking switches with  $N_1^2$ channels. Consider this method on an example of the dual switch SF2 with the dual quasi-complete graph topology SQG (2, 2, 2) (Fig. 2).



Fig. 2. Dual non-blocking switch SF2 with one-channel fault tolerance.

On its basis, a four-channel two-stage network  $N_24$  is constructed. This network contains two SF2 switches on each stage, connected by exchange links (Fig. 3). The network is blocking on multiplexers of the first stage, highlighted in grey, and loses channel fault tolerance on them.



Fig. 3. Dual two-stage blocking network N<sub>2</sub>4 with exchange links.



<sup>&</sup>lt;sup>1</sup> This term was introduced by the author.



The network  $N_24$  can be transformed into a nonblocking switch  $S_24$  with one-channel fault tolerance by internal parallelization [5, 6]. The second stage of this switch uses two copies of the second stage of the  $N_24$  network. On the first stage, multiplexers on the cross-sections are eliminated, and their inputs are connected to the inputs of the second stage copies: odd to the first copy and even to the second copy. These links preserve the order of connecting the channels located on the second stage in the network  $N_24$ . The cut-out multiplexers are moved to connect the outputs of the second stage copies, forming the output multiplexers of the switch  $S_24$  (Fig. 4), which becomes a nonblocking switch with static self-routing and onechannel fault tolerance.

In the general case  $(p \ge 2)$ , the dual switch SFN<sub>1</sub> has the dual quasi-complete graph topology SQG( $N_1$ , p,  $\sigma$ ) with a signal period of p cycles. On its basis, a two-stage blocking network N<sub>2</sub> $N_2$  ( $N_2 = N_1^2$ ) is constructed, where each stage contains  $N_1$  switches SFN<sub>1</sub> with exchange links between the stages. For internal parallelization, p copies of the second stage of the network N<sub>2</sub> $N_2$  are formed, and the first stage multiplexers are used to combine the same-name outputs of the second stage copies.



Fig. 4. Dual non-blocking self-routing switch S<sub>2</sub>4 with one-channel fault tolerance.

On the cross-sections of the first stage, there are  $pN_2$ inputs to the multiplexers. They are renumbered topto-bottom by I ( $1 \le I \le pN_2$ ), and the inputs i = I(modp)+1 are connected to the same-name inputs of the *i*th copy of the second stage, maintaining the same arrangement of the switches S<sub>1</sub>N<sub>1</sub> as in the network N<sub>2</sub>N<sub>2</sub>.

**Lemma 2.** The dual switch  $S_2N_2$  has a ppermutation on the indicated cross-section. It is a nonblocking switch with static routing on any common permutation and has  $(\sigma - 1)$ -channel fault tolerance.

The switching and channel complexities of the switch  $S_2N_2$  are given by the recursive formulas  $S_2 = N_1S_1 + pN_1S_1$  and  $L_2 = N_1L_1 + pN_1L_1$ , respectively. The performance characteristics of the switches  $S_2N_2$  for  $\sigma = 2$  are presented in Table 3. Note that the exponential complexities decrease compared to Table 1.

Table 3

## Performance characteristics of dual switches $S_2N_2$ with one-channel fault tolerance

р	$N_1$	$N_2 = N_1^2$	$T_2 = p$	$S_2$	$L_2$
2	2	4	2	$N_2^{3.58}$	$N_2^{2.9}$
4	7	49	4	$N_2^{2.37}$	$N_2^{1.97}$
6	15	225	6	$N_2^{2.18}$	$N_2^{1.84}$
8	27	729	8	$N_2^{2.09}$	$N_2^{1.77}$

The performance characteristics of a dual switch  $S_2N_2$  based on the quasi-complete digraph topology [5] are combined in Table 4. They are significantly better than in Table 2, but without channel fault tolerance.

Table 4

## Performance characteristics of switches $S_2N_2$ based on the quasi-complete digraph topology

р	$N_1$	$N_2 = N_1^2$	$T_2 = p$	$S_2$	$L_2$
2	4	16	2	$N_2^{2.29}$	$N_2^{1.95}$
4	16	256	4	$N_2^{1.96}$	$N_2^{1.68}$
6	36	1 296	6	$N_2^{1.89}$	$N_2^{1.63}$
8	64	4 096	8	$N_2^{1.86}$	$N_2^{1.6}$

Also, note that the dual switch  $S_2N_2$  has two stages of output multiplexers containing  $pN_2$  and  $N_2$  multiplexers, respectively. For the purposes of Section 3, we cut the switch  $S_2N_2$  through the inputs of the first stage of multiplexers; see the dash-and-dot line in Fig. 4.

# 3. FOUR-STAGE FAULT-TOLERANT NON-BLOCKING SELF-ROUTING SWITCH WITH TWO-DIMENSIONAL INTERNAL PARALLELIZATION

In the papers [1–3, 5], the number of channels of a non-blocking switch was further increased using the invariant extension method with external parallelization. This method does not change the signal period. However, it is of little use for fault-tolerant non-blocking switches [6].

In this section, we increase the number of channels without changing the signal period using the generalized method of internal parallelization of the network by constructing four-stage switches  $S_4N_4$  from twostage switches  $S_2N_2$  with the number of channels  $N_4 = N_2^2$  and the signal period  $T_4 = T_2 = p$ .

The network is constructed using switches  $S_24$  as an example (Fig. 4). First, the two-stage network  $N_416$ is created. Each stage in this network consists of four copies of the  $S_24$  switch, and the stages are interconnected by exchange links (Fig. 5).



Fig. 5. Blocking dual network N<sub>4</sub>16 without channel fault tolerance.

In reality, the network  $N_416$  consists of four stages SQG(2, 2, 2), which explains the subscript in the notation. The  $N_416$  network is blocking due to possible signal conflicts on the two stages of the output multi-

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plexers M2 (highlighted in grey). There are two layers of such multiplexers,  $W_4 = 48$  in total. In addition, channel fault tolerance is violated on them. The dashand-dot cut is made through the inputs to the first stage (Fig. 5). For this cut, the notion of a *p*permutation has been formulated in Section 2.

Then the network  $N_416$  is created. It contains the first stage of the network  $N_416$  and two copies of the second stage of the network  $N_416$ . One parallel circuit of the first dimension is created in the network  $N_416$  (Fig. 6). For this, the outer layer highlighted in pale grey is first cut out with  $W_2^* = 16$  multiplexers M2 in total. They remain unconnected for now. Then the multiplexers M2 of the inner layer highlighted in dark grey are cut out, and their odd inputs are routed to the inputs of two second-stage copies of the network  $N_416$ . In this case, the  $W_{4,1} = 16$  cut multiplexers combine the outputs of these two copies.

The remaining  $W_{4,2} = 16$  multiplexers M2 highlighted in dark grey are used to create the second parallel circuit of the first dimension in the same way (Fig. 7). Their even inputs are routed to the inputs of two additional copies of the second stage of the network N<sub>4</sub>16.

Looking ahead, note that Figs. 6 and 7 show the new connections of the multiplexers of the first and second layers. They define the combination of the first dimension circuits into the second dimension circuit.

As a result, two circuits of the first dimension are constructed, each consisting of two switches  $S_24$  connected in parallel. (The connections of the second circuit of the first dimension are not shown in Fig. 8.) The two circuits of the first dimension form a twodimensional circuit. The outputs of the twodimensional circuit combine  $W_2^* = 16$  multiplexers highlighted in pale grey, forming the outputs of the switch  $S_416$ . In Fig. 8, the latter connections are indicated by dotted lines and are shown completely in one copy only due to the lack of space. (They can be found in Figs. 6 and 7.)

The resulting sixteen-channel network consists of 16 copies of the switch  $S_24$  connected in parallel. Their inputs receive sparse alternative direct *p*-permutations implemented without conflict according to a single static schedule; see Lemma 2. (Alternatives *p*-permutations intersect neither in inputs nor outputs.) The paths between sources and sinks in switches  $S_24$  follow two subpaths through different circuits of the first dimension. Therefore, the switch  $S_24$  has one-channel fault tolerance since  $p = \sigma = 2$ .



Fig. 6. Constructing the first circuit of the first dimension. Multiplexers of network N<sub>4</sub>16 not used in this circuit are shown on the left.

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Fig. 7. Constructing the second circuit of the first dimension using the multiplexers not included in the first circuit of the first dimension.



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Fig. 8. Non-blocking dual switch  $S_4 16$  with one-channel fault tolerance.

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In the general case (p > 2 and  $\sigma \ge 2$ ), the network  $N_4N_4$  is first constructed. It consists of two stages with  $N_2$  switches  $S_2N_2$  in each, connected by exchange links. This network has  $N_4 = N_2^2$  channels, and signals can conflict at output multiplexers Mp. Therefore, it is a blocking network without channel fault tolerance.

The first stage of the network  $N_4N_4$  has two layers of output multiplexers Mp,  $W_4 = N_2V_2 = N_4(p+1)$  in total. The first inner layer of multiplexers Mp contains  $W_{4,1} = pN_4$  multiplexers Mp, which together have  $p^2N_4$ inputs.

Then the network  $\underline{N}_4N_4$  is created. It contains the first stage of the network  $N_4N_4$  and  $p^2$  copies of the second stage of the network  $N_4N_4$ . The network  $\underline{N}_4N_4$  is constructed with the following structure: copies of the second stage of the network  $N_4N_4$  form *p* circuits of the first dimension, and all together, they form the circuit of the second dimension.

The network  $N_4N_4$  contains  $p^2N_2$  switches  $S_2N_2$ , which have  $p^2N_4$  inputs in total. Further, both layers of multiplexers Mp in the first stage of the network  $N_4N_4$ are cut out, and the inputs of the first layer of multiplexers Mp are connected to the inputs of switches  $S_2N_2$ . This is possible since the latter and former have the same number of inputs.

We divide the  $p^2$  copies of the second stage of the network N<sub>4</sub>N<sub>4</sub> in the network <u>N</u><sub>4</sub>N<sub>4</sub> into *p* groups, denoting by *G* ( $1 \le G \le p$ ) the group number and by *I* ( $1 \le I \le p$ ) the copy number. In fact, *G* is the circuit number of the first dimension, and *I* is the copy number in the first dimension circuit.

In addition, we introduce the following notations: J  $(1 \le J \le N_2)$  is the number of the switch  $S_2N_2$  in the first dimension circuit, and K  $(1 \le K \le N_2)$  is the input number of each such switch  $S_2N_2$ . Thus, the input of any switch  $S_2N_2$  is given by the composite number G, I, J, K.

Also, we denote by i  $(0 \le i \le p - 1)$  the modp input number of each multiplexer Mp in their first layer. Each switch in the first stage of the network  $N_4N_4$  contains  $pN_2$  such multiplexers Mp. We divide them into  $N_2$  groups, denoting by g  $(1 \le g \le N_2)$  the group number and by j  $(1 \le j \le p)$  the number of the multiplexer Mp in the group. Let k  $(1 \le k \le N_2)$  denote the number of the switch  $S_2N_2$  in the first stage of the network  $N_4N_4$ . Thus, the input of any multiplexer Mp in the first layer is given by the composite number i, g, j, k.

An arbitrary input of the multiplexer Mp with the number *i*, *g*, *j*, *k* is connected to the input of the switch  $S_2N_2$  with the number *G*, *I*, *J*, *K*, where G = i + 1,

I = j, J = g, and K = k. As a result,  $p^2N_2$  switches  $S_2N_2$  are connected in parallel, and their inputs receive sparse disjoint direct *p*-permutations.

The cut-out multiplexers Mp of the first layer combine the outputs of the first dimension circuits. The cut-out multiplexers Mp of the second layer combine the outputs of the *p* circuits of the first dimension, forming the outputs of the second-level circuit (the outputs of the switch  $S_4N_4$ ).

Switches  $S_4N_4$  have the following property.

**Lemma 3.** The dual switch  $S_4N_4$  is a nonblocking switch with static routing on any common permutation for any p. It has  $(\sigma - 1)$ -channel fault tolerance.

P r o o f. The first statement is based on using the switch  $S_2N_2$  and Lemma 2. The second statement is based on the non-blocking property of the switch  $S_2N_2$  and the fact that the *p*-permutation on the cross-section consists of sparse 1-permutations separated to different channels and cycles.

According to the assignment G = i+1, different inputs of one multiplexer in the first layer are connected to different one-dimensional circuits (different copies of the second stage of the network N<sub>4</sub>N<sub>4</sub>), and the inputs of different multiplexers are connected to the inputs of different switches S<sub>2</sub>N<sub>2</sub> in the second stage of the network N<sub>4</sub>N<sub>4</sub>. In other words, there are  $\sigma$  different paths of switches S<sub>2</sub>N<sub>2</sub> along p different paths in the switch S<sub>4</sub>N<sub>4</sub>. Due to  $p \ge \sigma$ , the latter switch is therefore ( $\sigma - 1$ )-channel fault-tolerant.  $\blacklozenge$ 

As a result, the non-blocking self-routing switch  $S_4N_4$  with  $N_4 = N_2^2$  channels and  $(\sigma - 1)$ -channel fault tolerance has the performance characteristics shown in Tables 5 and 6. They are calculated using the recursive formulas  $S_4 = N_2S_2 + p^2N_2S_2$  and  $L_4 = N_2L_2 + p^2N_2L_2$ . The switch  $S_4N_4$  has four layers of output multiplexers Mp,  $V_4 = N_4(p^4 - 1)/(p - 1)$  in total.

Note the further decrease in the exponential switching and channel complexities of the switch  $S_4N_4$  (Table 4) compared to the switch  $S_2N_2$  (Table 2).

Abandoning the requirement of channel fault tolerance, we can construct a non-blocking self-routing switch  $S_4N_4$  based on the switch with the quasicomplete digraph topology. Its performance characteristics are presented in Table 7. Compared to the faulttolerant modifications, it has more channels and smaller complexity. In addition, the entire set of switches has switching and channel complexities less than those of a two-stage switch based on a switch with the complete graph topology (Table 3) and less than those of a switch with the complete graph topology (*switchboard*).


## 4. EIGHT-STAGE NON-BLOCKING SWITCHES WITH FOUR-DIMENSIONAL INTERNAL PARALLELIZATION BASED ON THE GRAPH AND DIGRAPH TOPOLOGIES

The method of extending two-stage switches  $S_2N_2$ into four-stage switches  $S_4N_4$  can be generalized to construct eight-stage switches  $S_8N_8$  from four-stage switches  $S_4N_4$ .

First, the network  $N_8N_8$  is constructed. It consists of two stages with  $N_4$  switches  $S_4N_4$  in each, connected by exchange links. This network has  $N_8 = N_4^2$  channels, and signals can conflict at output multiplexers Mp of the first stage. Therefore, it is a blocking network without channel fault tolerance.

The first stage of the network  $N_8N_8$  has four layers of output multiplexers Mp,  $W_8 = N_4V_4 = N_8(p^4 - 1)/(p - 1)$  in total. The first inner layer of multiplexers Mpcontains  $W_{8,1} = p^3N_8$  multiplexers Mp, which together have  $p^4N_8$  inputs.

Then the network  $N_8N_8$  is created. It contains  $p^4$  co pies of the second stage of the network  $N_8N_8$ . The network  $N_8N_8$  is created with the following structure.

The second stage copies of the network  $N_8N_8$ , p in total, form the circuit of the first dimension, and p circuits of the first dimension form the second

dimension circuit. Similarly, the third dimension circuit consists of

p second dimension circuits, and

the fourth dimension circuit con-

sists of *p* third dimension circuits. The network  $N_8N_8$  contains  $p^4N_4$  switches  $S_4N_4$ , which have  $p^4N_8$  inputs in total. Further, all four layers of multiplexers Mp in the first stage of the network  $N_8N_8$  are cut out, and the inputs of the first layer of multiplexers Mp are connected to the inputs of switches  $S_4N_4$ . This is possible since the latter and former have the same

Table 5

## Performance characteristics of dual switches $S_4N_4$ with one-channel fault tolerance

р	$N_1$	$N_4 = N_1^4$	$T_4 = p$	$S_4$	$L_4$
2	2	16	2	$2720 = N_4^{2.85}$	$1\ 120 = N_4^{2.53}$
3	4	256	3	$238\ 080 = N_4^{2.23}$	$69\ 120 = N_4^{2.01}$
4	7	2 401	4	$8\ 000\ 132 = N_4^{2.04}$	$1\ 795\ 948 = N_4^{1.85}$
5	11	14 641	5	$1.35E+08 = N_4^{1.95}$	$24\ 743\ 290 = N_4^{1.77}$
6	15	65 536	6	$1.41E+09 = N_4^{1.9}$	$2.18E + 08 = N_4^{1.73}$
7	21	194 481	7	$8.64E + 09 = N_4^{1.88}$	$1.16E+09 = N_4^{1.71}$
8	27	531 441	8	$4.45E+10 = N_4^{1.86}$	$5.25E+09 = N_4^{1.7}$

Table 6

Performance characteristics of dual switches  $S_4N_4$  with two-channel fault tolerance

р	$N_1$	$N_4 = N_1^4$	$T_4 = p$	$S_4$	$L_4$
3	3	81	3	$75\ 330 = N_4^{2.56}$	$21\ 870 = N_4^{2.27}$
4	5	625	4	$2\ 082\ 500 = N_4^{2.26}$	$467\ 500 = N_4^{2.03}$
5	7	2 401	5	$22\ 161\ 230 = N_4^{2.17}$	$4\ 057\ 690 = N_4^{1.95}$
6	11	14 641	6	$3.15E+08 = N_4^{2.04}$	$48\ 754\ 530 = N_4^{1.85}$
7	15	50 625	7	$2.25E + 09 = N_4^{1.99}$	$3.01E+08 = N_4^{1.8}$
8	19	130 321	8	$1.09E+10 = N_4^{1.96}$	$1.29E+09 = N_4^{1.78}$

1 4010 0

We divide the  $p^4$  copies of the second stage of the network N<sub>8</sub>N<sub>8</sub> in the network  $\underline{N}_8N_8$  into  $p^3$ groups, denoting by G $(1 \le G \le p^3)$  the group number and by I  $(1 \le I \le p)$  the number in the group. In fact, G is the circuit number of the first dimension, and I is the copy number of S<sub>4</sub>N<sub>4</sub> in the first dimension circuit.

number of inputs.

Table 7

#### Performance characteristics of dual switches $S_4 N_4$ based on the digraph topology

p	$N_1$	$N_2 = N_1^2 = p^4$	$N_4 = N_2^4 = p^8$	$T_4 = p$	$S_4$	$L_4$
2	4	16	256	2	43 520 = $N_4^{1.93}$	$17\ 920 = N_4^{1.77}$
3	9	81	6 561	3	$6 \ 101 \ 730 = N_4^{1.78}$	$1\ 771\ 470 = N_4^{1.64}$
4	16	256	65 536	4	$2.18E + 08 = N_4^{1.73}$	$49\ 020\ 928 = N_4^{1.60}$
5	25	625	390 625	5	$3.61E + 09 = N_4^{1.71}$	$6.6E + 08 = N_4^{1.58}$
6	36	1 296	1 679 616	6	$3.62E+10 = N_4^{1.70}$	$5.59E + 09 = N_4^{1.57}$
7	49	2 401	5 764 801	7	$2.56E + 11 = N_4^{1.69}$	$3.43E+10 = N_4^{1.56}$
8	64	4 096	16 777 216	8	$1.4\text{E}+12 = N_4^{1.68}$	$1.66E+11 = N_4^{1.55}$



In addition, we introduce the following notations: J  $(1 \le J \le N_4)$  is the number of the switch  $S_4N_4$  in the first dimension circuit, and K  $(1 \le K \le N_4)$  is the input number of each such switch  $S_4N_4$ . Thus, the input of any switch  $S_4N_4$  is given by the composite number G, I, J, K.

Also, we denote by i  $(0 \le i \le p - 1)$  the modp input number of each multiplexer Mp in their first layer. Each switch in the first stage of the network N<sub>8</sub>N<sub>8</sub> contains  $p^3N_4$  such multiplexers Mp. We divide them into N<sub>4</sub> groups, denoting by g  $(1 \le g \le N_4)$  the group number and by j  $(1 \le j \le p)$  the number of the multiplexer Mp in the group.

Let k ( $1 \le k \le N_4$ ) denote the number of the switch  $S_4N_4$  in the first stage of the network  $N_8N_8$ . Thus, the input of any multiplexer Mp in the first layer is given by the composite number *i*, *g*, *j*, *k*. For this purpose, all layers of multiplexers Mp in the first stage of the network  $N_8N_8$  are cut out. The first inner layer contains  $W_{8,1} = N_4p^3$  multiplexers Mp, and their inputs are connected to the inputs of switches  $S_4N_4$  in the network  $N_8N_8$ .

An arbitrary input of the first layer multiplexer Mp with the number *i*, *g*, *j*, *k* is connected to the input of the switch  $S_4N_4$  with the number *G*, *I*, *J*, *K*, where G = i + 1, I = j, J = g, and K = k. As a result,  $p^4N_4$ 

switches  $S_4N_4$  are connected in parallel, and their inputs receive sparse disjoint direct *p*-permutations.

The cut-out multiplexers Mp of the first layer combine the outputs of the  $p^3$  first dimension circuits. The cut-out multiplexers Mp of the second layer combine the outputs of the first dimension circuits with the same number G, forming the outputs of the secondlevel circuits with this number. The cut-out multiplexers Mp of the third layer combine the outputs of the second dimension circuits with the same numbers G, forming the outputs of the third-level circuits with this number. The cut-out multiplexers Mp of the fourth layer combine the outputs of the third dimension circuits, forming the outputs of the switch  $S_8N_8$ .

Switches  $S_8N_8$  have the following property.

**Lemma 4.** The dual switch  $S_8N_8$  is a nonblocking switch with static routing on any common permutation for any p. It has  $(\sigma - 1)$ -channel fault tolerance.

P r o o f. The first statement is based on using the switch  $S_4N_4$  and Lemma 3. The second statement is based on the non-blocking property of the switch  $S_4N_4$  and the fact that the *p*-permutation on the cross-section consists of sparse 1-permutations separated to different channels and cycles.

Channel fault-tolerance holds since the paths between sources and sinks in the switch  $S_4N_4$  are through different

circuits of each dimension and  $p \ge \sigma$ .

Table 8 <sup>I</sup>

The resulting non-blocking self-routing switch  $S_8N_8$  includes  $N_8 = N_4^2$  channels and is  $(\sigma - 1)$ -channel fault-tolerant.

The switch  $S_8N_8$  has eight layers of output multiplexers Mp,  $V_8 = N_8(p^8 - 1)/(p-1)$  in total.

Tables 8–10 present the performance characteristics of the fastest modifications of the switch  $S_8N_8$  for  $\sigma = 2$ and  $\sigma = 3$ . The switching and channel complexities are calculated by the recursive formulas  $S_8 = N_4 S_4 +$  $p^4N_4S_4$ and  $L_8 = N_4 L_4 +$  $p^4N_4L_4$ , respectively. Note that the switching and channel complexities of the switch  $S_8N_8$  based on a digraph can be made significantly less than those of a lumped switch with the complete graph topology.

Performance characteristics of dual switches  $S_8N_8$  with one-channel fault tolerance

р	$N_1$	$N_8 = N_1^{8}$	$T_8 = p$	$S_8$	$L_8$
2	2	256	2	739 840 = $N_8^{2.44}$	$304\ 640 = N_8^{2.28}$
3	4	65 536	3	$4.998E + 09 = N_8^{2.01}$	$1.451E+09 = N_8^{1.9}$
4	7	5 764 801	4	$4.937E+12 = N_8^{1.88}$	$1.108E+12 = N_8^{1.78}$

Table 9

#### Performance characteristics of dual switches $S_8 M_8$ with two-channel fault tolerance

р	$N_1$	$N_8 = N_1^{8}$	$T_8 = p$	$S_8$	$L_8$
3	3	6 561	3	$500 \ 341 \ 860 = N_8^{2.28}$	$145\ 260\ 540 = N_8^{2.14}$
4	5	390 625	4	$3.345E+11 = N_8^{2.06}$	$7.509E + 10 = N_8^{1.94}$
5	7	5 764 801	5	$3.331E+13 = N_8^2$	$6.099E + 12 = N_8^{1.89}$

Table 10

#### Performance characteristics of dual switches $S_8 N_8$ based on the digraph topology

р	$N_1$	$N_8 = N_1^{8}$	$T_8 = p$	$S_8$	$L_8$
2	4	65 536	2	$189\ 399\ 040 = N_8^{1.72}$	77 987 840 = $N_8^{1.64}$
3	9	5 764 801	3	$3.283\text{E}{+}12 = N_8^{1.64}$	$9.531E + 11 = N_8^{1.57}$
4	16	4.29E+09	4	$3.678E+15 = N_8^{1.62}$	$8.256E + 14 = N_8^{1.55}$





## 5. ANALYSIS OF THE RESULTS. PRACTICAL DEVELOPMENT OF THE CONSTRUCTED NETWORKS

This paper has proposed a method for constructing a new class of non-blocking self-routing photon networks with high scalability. These are the so-called dual networks based on a non-blocking dual  $p \times p$ switch with a signal period of p cycles.

The dual switch is used as an integral part of the non-blocking self-routing  $N_1 \times N_1$  switch  $S_1N_1$  with the quasi-complete graph or digraph topology. In the former case, the number of channels is  $N_1 = p(p - 1)/\sigma + 1$ , and  $(\sigma - 1)$ -channel fault tolerance can be provided. In the latter case, the number of channels is  $N_1 = p^2$ , which can be even increased. The switch with the quasi-complete (di)graph topology consists of  $N_1$  dual  $p \times p$  switches together with  $N_1 1 \times p$  demultiplexers Dp and  $p \times 1$  multiplexers Mp without delay lines. The switching complexity of  $S_1N_1$  is given by  $S_1 = N_1(S_0 + 2p)$ . The signal period  $T_1$  of the switch  $S_1N_1$  equals that of the dual switch:  $T_1 = p$ .

Switches  $S_1N_1$  form two stages to construct a blocking  $N_2 \times N_2$  network  $N_2N_2$  with  $N_2 = N_1^2$  channels. Each stage consists of  $N_1 N_1 \times N_1$  switches, and the channels between the stages are built using exchange links. The network  $N_2N_2$  is transformed into a non-blocking self-routing two-stage switch  $S_2N_2$  by one-dimensional internal parallelization.

If the  $N_1 \times N_1$  switch is based on a quasi-complete graph, the switch  $S_2N_2$  has  $(\sigma - 1)$ -channel fault tolerance since  $p \ge \sigma$ . The switching and channel complexities of the switch  $S_2N_2$  are given by the recursive formulas  $S_2 = N_1S_1 + pN_1S_1$  and  $L_2 = N_1L_1 + pN_1L_1$ , respectively. By construction, the signal period  $T_2$  of the switch  $S_2N_2$  equals that of the switch  $S_1N_1$ :  $T_2 = T_1 = p$ .

If the  $N_1 \times N_1$  switch is based on a quasi-complete graph, the four-stage switch  $S_4N_4$  has  $(\sigma - 1)$ -channel fault tolerance since  $p \ge \sigma$ . The switching and channel complexities of the switch  $S_4N_4$  are given by the recursive formulas  $S_4 = N_2S_2 + p^2N_2S_2$  and  $L_4 = N_2L_2 + p^2N_2L_2$ , respectively. By construction, the signal period  $T_4$  of the switch  $S_4N_4$  equals that of the switch  $S_2N_2$ :  $T_4 = T_2 = p$ .

Similarly, switches  $S_4N_4$  form two stages to construct a blocking  $N_8 \times N_8$  network  $N_8N_8$  with  $N_8 = N_4^{2} = N_1^{8}$  channels. Each stage consists of  $N_4$ switches  $S_4N_4$ , and the channels between the stages are built using exchange links.

The network  $N_8N_8$  is transformed into a nonblocking self-routing two-stage switch  $S_8N_8$  by fourdimensional internal parallelization. If the  $N_1 \times N_1$  switch is based on a quasi-complete graph, then the eight-stage switch  $S_8N_8$  has  $(\sigma - 1)$ channel fault tolerance as well. The switching and channel complexities of the switch  $S_8N_8$  are given by the recursive formulas  $S_8 = N_4S_4 + p^4N_4S_4$  and  $L_8 = N_4L_4 + p^4N_4L_4$ , respectively. By construction, the signal period  $T_8$  of the switch  $S_8N_8$  equals that of the switch  $S_4N_4$ :  $T_8 = T_4 = p$ .

The performance characteristics of the switches  $S_2N_2$ ,  $S_4N_4$ , and  $S_8N_8$  have several degrees of freedom. First of all, the number of channels grows with increasing the base p, and the speed decreases. In addition, the exponential complexity decreases with increasing the base p, and the speed can be traded for complexity. Also, more channels due to increasing the number of stages reduce the exponential complexity.

The proposed method allows constructing nonblocking self-routing networks with a self-similar structure. The switch  $S_2N_2$  consists of dual switches  $S_1N_1$  with the dual graph or digraph topology and uses one-dimensional internal parallelization. In turn, the switch  $S_4N_4$  consists of switches  $S_2N_2$  and uses twodimensional internal parallelization. Finally, the switch  $S_8N_8$  is composed of switches  $S_4N_4$  and uses four-dimensional internal parallelization. All these switches inherit the basic properties of the switch  $S_1N_1$ , such as the non-blocking property under static self-routing and channel fault tolerance (if necessary), but with significantly less complexity.

The high scalability of non-blocking switches can also be achieved by repeated application of the invariant extension method to the switch  $S_1N_1$  with the digraph topology based on a conventional  $p \times p$  switch. Such extended switches have a signal period of one cycle but increased complexity. Table 11 compares the switching complexity of dual switches  $S_4N_4$  and  $S_8N_8$  and extended switches  $S_1N_1$ . Clearly, the switching complexity of dual switches is by several orders of magnitude lower.

Note that the dual switch resolves conflicts by the bus method only in the first stage of the switch  $S_1N_1$ . All other conflicts in all stages are prevented using internal parallelization, and the dual switches in them are used as common  $p \times p$  switches. Therefore, it seems reasonable to use the dual switch in its original form [1–3] (the multiplexer–demultiplexer pair): its switching complexity is p times less. This approach will reduce the switching complexity of the dual switches  $S_4N_4$  and  $S_8N_8$  by several times (1.5–4.5).

Note that for small p, the complexity of the faulttolerant switches  $S_2N_2$  and  $S_4N_4$  is greater than that of the complete graph; for large p, it is smaller. In this



Table 11

Switching complexities of non-blocking four-stage switches $(S_{4, D})$										
	and extended switches based on the quasi-complete digraph topology $(S_{PO})$									
р	N <sub>4</sub>	Dual switch $S_4N_4$ $S_{4, D}$	Extended switch $S_1 N_1$ $S_{PO}$	Ratio S <sub>PO</sub> /S <sub>4, D</sub>						
2	256	$46\ 080 = N_4^{1.94}$	$261\ 120 = N_4^{\ 2.25}$	5.67						
3	6 561	$6\ 298\ 560 = N_4^{1.78}$	$129\ 120\ 480 = N_4^{2.12}$	20.5						
4	65 536	$22\ 282\ 400 = N_4^{1.74}$	$11\ 453\ 071\ 360 = N_4^{2.09}$	514						
	Switching complexities of non-blocking eight-stage switches $(S_{8,D})$									
	and extended switches based on the quasi-complete digraph topology $(S_{PO})$									
р	$N_8$	Dual switch $S_8N_8$	Extended switch $S_1N_1$	Ratio $S_{\rm PO}/S_{8, D}$						
		$S_{8, D}$	$S_{ m PO}$							
2	65 536	$18\ 939\ 9040 = N_8^{1.72}$	$17\ 179\ 607\ 040 = N_8^{2.12}$	85.7						
3	43 046 721	$3.283E+12 = N_8^{1.64}$	$5.55822E + 15 = N_8^{2.06}$	1 640						
4	4.29E+09	$3.678E + 15 = N_8^{1.62}$	$4.91906E + 19 = N_8^{2.04}$	13 107						

#### Complexity analysis: dual switches vs. extended digraphs

case, the complexity of switches  $S_8N_8$  is significantly less than that of the complete graph for any *p*. We emphasize the performance characteristics of the switch  $S_8N_8$  for p = 2 and  $\sigma = 1$ . With  $N_8 = 65536$  channels and half the speed, its switching complexity is comparable to that of a five-stage non-blocking Clos network based on a 64-channel YARC router [9] with N =32 768 channels constructed as a non-blocking network [7, 8]. The complexity of this non-blocking Clos network is estimated as  $S = N^{1.73}$ . However, this network has no parallel static or dynamic self-routing procedures. The other switches  $S_8N_8$  with p > 2 and  $\sigma = 1$  have even lower switching complexity and higher scalability but with lower speed.

The *p*-times reduced speed of the switches  $S_2N_2$ ,  $S_4N_4$ , and  $S_8N_8$  can be compensated by different protocols. It is possible to choose processors with *p* independent ports, divide packets into *p* parts and transmit them in parallel. The high scalability of these switches supports such an operation mode, albeit by reducing the number of users by *p* times and increasing the network complexity. An alternative is to apply the parallel-serial method for transmitting packets over *p* lines, as in the PCI Express protocol, without reducing the number of users.

A shortcoming of the switches  $S_1N_1$ ,  $S_2N_2$ ,  $S_4N_4$ , and  $S_8N_8$  is their optimization for the conflict-free implementation of arbitrary permutations. What will be their behavior on arbitrary traffic? To determine it, we can assign the one-channel property to the multiplexers in the output stages. When receiving several input packets, such a multiplexer passes only one packet and blocks the others. The blocked packets not acknowledged by sinks are re-transmitted by the sources.

A considerable disadvantage of the proposed method is the need for parallel transmission of signal and control information, which significantly increases the required bandwidth. This disadvantage is not fatal for photon switches since an optical cable can simultaneously carry hundreds of different frequencies. However, this disadvantage can be generally eliminated with bit synchronization of signals from different channels. This can be done using the method [34, 35], locating the mutual arrangement of sources and sinks and the corresponding transmission delays from the sources. In this case, control information for dual switches and demultiplexers can be transmitted, as usual, in the form of sets of bits in the packet header.

A pleasant bonus of bit synchronization is the ability to construct an arithmetic logic unit (ALU) in the channel at each sink's input using network means. Such ALUs were developed for computing in the common channel [36]. For implementing them, it is necessary to transmit through the channel the digit values in the two-signal form: along two lines with active signals for values 0 and 1 in each. In the network ALU, an operation is performed over the number arriving through the channel and the number at the sink; the result is formed in the channel after the ALU. In the channel, it is possible to perform addition, multiplication, and any bitwise logical operations, including finding the maximum (minimum).

#### CONCLUSIONS

This paper has proposed a method for constructing non-blocking fault-tolerant photon networks with high scalability, considered in [5], but with much less complexity. This method is based on three main components:

• A p-channel dual switch with a signal period of p cycles, which turns out to be non-blocking on any input traffic (a prerequisite for constructing more complex non-blocking networks).

• A switch with the quasi-complete graph or digraph topology and a dual switch inside. As a result, the non-blocking property is maintained, and channel fault tolerance and higher scalability during cascading are provided compared to a pure dual switch.

• Internal parallelization to maintain the nonblocking property by preventing conflicts and maintaining fault tolerance, which provides high scalability when cascading non-blocking networks.

In the paper [5], scaling was implemented by cascade application of the invariant extension method with additional external multiplexers and demultiplexers. In this paper, scaling has been implemented by cascading smaller non-blocking networks and applying the generalized internal parallelization method at each cascading step.

The cascading of a non-blocking network with Nchannels is performed by constructing a blocking network with  $N^2$  channels. This network consists of two stages with exchange links with N original nonblocking networks in each. Interlocks in this two-stage network occur at the output multiplexers of the first stage. These interlocks are prevented by separating the conflicting channels to multiple copies of the second stage and moving the multiplexers to the outputs by the second stage part responsible for packet routing. No conflicts occur in this part of the network since it consists of copies of non-blocking subnetworks that route sparse permutations. Sparse permutations are united into a complete permutation on a network with  $N^2$  channels by the moved stages of multiplexers without conflict.

During the first cascading [5], internal parallelization is performed using p second stage copies and a one-layer stage of output multiplexers. When constructing a non-blocking network with  $N^4$  channels, the second cascading is performed using  $p^2$  second stage copies and a two-layer stage of the output multiplexers. When constructing a non-blocking network with  $N^8$  channels, the third cascading is performed using  $p^4$  second stage copies and a four-layer stage of the output multiplexers. Thus, we have designed nonblocking two-, four-, and eight-stage networks with stages consisting of non-blocking dual networks with the quasi-complete graph or digraph topology.

During each cascading, internal parallelization maintains the signal period and reduces the specific complexity of the non-blocking network. In particular, we have constructed non-blocking networks with a specific complexity not exceeding that of the theoretical non-blocking Clos network.

This method can be a fundamental base for constructing practical non-blocking switches with high scalability, static self-routing, and channel fault tolerance.

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*This paper was recommended for publication by V.M. Vishnevsky, a member of the Editorial Board.* 

> Received March 25, 2021, and revised August 12, 2021. Accepted August 24, 2021.

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#### Cite this article

Podlazov, V.S. Non-blocking Fault-Tolerant Dual Photon Switches with High Scalability. *Control Sciences* **5**, 61–76 (2021). http://doi.org/10.25728/cs.2021.5.6

Original Russian Text © Podlazov, V.S., 2021, published in *Problemy Upravleniya*, 2021, no. 5, pp. 70–87.

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DOI: http://doi.org/10.25728/cs.2021.5.7

# SCENARIO METHODS TO IMPROVE THE EFFICIENCY OF IMPLEMENTING THE LIFE CYCLE OF PROGRAM-TARGET MANAGEMENT: A CONCEPTUAL ANALYSIS

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Abstract. This paper is devoted to methodological and applied problems of improving the efficiency of program-target planning and management in socio-economic systems based on a scenario approach. The main features of the methodology and mechanisms of program-target management are considered. The idea proposed below is to manage and control the implementation of target programs using a scenario approach and life cycle models. A scenario W-shaped life cycle model is developed to improve the efficiency of program-target planning and management. This model involves the end-to-end methodology of proactive management to implement long-term socio-economic programs under uncertainty and risk. The model's methodological core consists of the following mechanisms: scenario analysis, simulation, forecasting, planning, and group management. The new approach is most effective under increasing uncertainty due to its focus on anticipating future conditions and alternatives for the development of socioeconomic systems.

Keywords: planning, management, efficiency, target program, scenario approach, life cycle, uncertainty.

### INTRODUCTION

Strengthening of negative processes in world development in the 2010s, the deepening of the global economic crisis caused by the COVID-19 pandemic, the destruction of international economic ties, the strengthening of the anti-Russian sanctions policy of the Western countries, the deterioration of budget indicators related to internal problems, and the decline in production have resulted in significant growth of various risks and uncertainties. All these factors tighten the requirements for the quality and efficiency of management of socio-economic systems (SESs), considerably narrowing the scope of conventional approaches and methods to plan and manage the sustainable development of such systems.

In the current unstable conditions, scenario analysis and simulation methods based on the program-target approach increase their role and importance. These methods reduce uncertainty and consider a wide range of external and internal threats to achieving the goals by assessing the most probable and reasonable trends in the development of dynamic processes in the sectors under consideration, their stability, and other desired and undesired properties based on information about their structural features. Note that the sectors are characterized by a complex spatial, administrative, and managerial organization of SESs.

## 1. IMPROVING THE MECHANISMS OF PROGRAM-TARGET PLANNING AND MANAGEMENT

Program-target planning and management is a special public administration method intended primarily for the critical development areas of national SESs that combines the interests of society, the state, and economic entities of various forms of ownership. This methodology allows solving large-scale and complex problems of SESs development by structuring and interconnecting their components and organizing mutually beneficial partnerships between the state, business, science, education, and civil society [1]. To a large extent, program-target planning relies on the hierarchy analysis methodology [2, 3] to achieve the main goal by decomposing it into a system of intermediate goals (subgoals). The subgoals are achieved by implementing a set of interconnected measures (programs) by a group of executors.

The complexity of increasing the efficiency of program-target planning and management is due to different methods and approaches to solving management problems at particular stages of the process. In addition, the implementation of any large-scale, long-term project is associated with numerous variable and newly emerging risks and external and internal threats of various natures to be considered during program-target planning and management. These risks and uncertainties are different and significantly affect the problem specifics at different stages of program-target management.

Large-scale projects and programs involve many economic entities of various forms of ownership from different industries. They directly participate in the measures of state or federal target programs or (directly or indirectly) work to achieve the main goal on their initiative based on the state's interest in particular problems of socio-economic development. (In the latter case, they expect the economic demand for the results of such work.) The economic entities possibly use market mechanisms of economic management [4].

In this situation, when using conventional approaches to management, there is a threat of a reduced socio-economic effect from implementing particular projects and programs due to possible errors in the process of analysis, forecasting, and goal-setting, determining the real needs for resources, improper coordination in the activities of the managing substructures and executors of programs.

A way to improve the efficiency of program-target planning and management is to develop and implement a methodology based on the process and scenario approaches [5]. This methodology should reduce uncertainty and consider a wide range of external and internal threats to achieving the goals at different stages of designing and implementing target programs and projects. In addition, this methodology should effectively coordinate and control the activities of numerous economic entities for executing program measures while solving strategic development problems characterized by a complex spatial, administrative, and managerial organization of socio-economic systems.

## 2. ANALYSIS OF MANAGEMENT PROCESSES FOR IMPLEMENTING TARGET PROGRAMS BASED ON LIFE CYCLE MODELS

Any target program goes through several life cycle stages: identifying problems that need to be solved, forming a system of goals and efficiency criteria, planning and developing mechanisms to achieve the goals under uncertainty (hard-to-predict changes in the internal and external environments), managing the implementation of program measures (particularly by exception), assessing the results, and completing the program (the last stage).

Nowadays, the term "life cycle" (LC) is widely used in natural sciences, engineering, humanities, and other sciences. We note the structural and, roughly speaking, conceptual similarity of the LC definitions in application domains [6]. A systems engineering definition of the complete LC model was given in the book [7]: "The complete life cycle model of an individual object is a description of the sequence of all phases and stages of its existence from conception and emergence ("birth") to disappearance ("extinction")." Consequently, the complete life cycle model is characterized by structural invariance: the set of the LC phases does not significantly depend on the object described. Due to this invariance, the model becomes quite universal and widely applicable since, in the general case, almost any control object (a program or project) goes through all the main LC phases: design, implementation and development, and completion. The most common LC parameters are the time instants of beginning and end, the total duration, the duration of each phase (stage, step, task, etc.), the sequence of stages and steps, the type and form of intraphase changes, and the set of indicators characterizing the object's state (by phases, stages, or steps). Each LC stage has particular managerial tasks and features.

At its core, the LC model is an applied organizational management tool based on a systems approach that allows:

• ensuring compliance with the goals and an adequate understanding of management processes by representing the final results as a cumulative outcome of intermediate tasks that are easier to understand and assess;

• decomposing the management process into relatively independent and, at the same time, logically interconnected temporal blocks;



• structuring the management process;

• determining step by step the content of managerial tasks, the technology for their solution, partial efficiency criteria, and resource-time constraints, considering the goals and the object's dynamics;

• as a consequence, improving the methodological base and organizational principles of the management system.

Various LC models are currently used for solving a wide class of problems. Among them, we mention the cascade, incremental, spiral, V-shaped, and other models. Each model is focused on particular application domains and, accordingly, has some advantages and disadvantages. Perhaps, only the V-shaped model agrees with the tasks and features of program-target management [6].

Note that program-target management has the following specifics: for a long life cycle of programs (3-5 years or more), a complex set of interconnected but different problems should be solved under uncertainty, risk, and continuous changes in the external and internal environments. (For example, this set includes goalsetting, risk analysis, forecasting, design of program measures, financial and resource support, operational management of the implementation of measures, monitoring and assessment of the program's result and the efficiency of using allocated resources, to name a few.) Moreover, ineffective decisions at any stage will entail a whole "bunch" of problems at the subsequent ones (in the worst case, return to the initial stages of goalsetting and planning to make significant corrections affecting the entire management cycle).

At present, most organizational management systems of various levels and designated purposes focus on the object's internal development processes. As a result, they solve mainly operational and partially tactical tasks with maximum success. For the same reason, as the input data they use difficult-to-predict and, in many cases, significant changes in the internal environment and almost all events outside the system (in the external environment).

For such management systems, the "negative" input information requiring an immediate response is mainly the results of direct (external or internal) influence on the object when, in most practical cases, the damage has been done (in the broadest sense of this term). If the management system is limited to counteracting the consequences of unfavorable situations, it cannot guarantee an efficient solution for strategic and medium-term tasks of managing the socio-economic development of the state and society. All these circumstances are the main source of reducing the efficiency of management processes for developing socioeconomic systems.

In these conditions, the role and importance of the scenario approach in program-target management increase. This approach yields a predictive assessment of the quality of decisions made at various stages of the life cycle of large projects and programs. Moreover, it reduces uncertainty by considering a wide range of external and internal threats to achieving goals. Uncertainty is understood as a situation when information about the structure and possible states of the SES and (or) its external environment is partially or completely absent. This concept is one of the key methodological concepts of the scenario approach. The construction of scenarios pursues two goals: reducing the original uncertainty as much as possible within this approach and describing the residual uncertainty using some scenarios. Thus, the uncertainty in situation development is subsequently decreased to find the best solutions of the arising problems.

Scenario analysis, forecasting, planning, and group management are technologies to assess comprehensively the efficiency and consistency of a set of managerial decisions when selecting and implementing the development programs of SESs and to consider and coordinate the goals and activities of the economic entities involved in the implementation of program measures. They independently plan and execute their tactical actions in the middle- and short-term horizons within common long-term goals, rules, and obligations regulated by the relevant target programs and projects.

For increasing the efficiency of program-target planning and management, we modify the V-shaped life cycle model by adding a branch that reflects:

- the methodological basis for ensuring the sustainable progressive socio-economic development of the state and society under uncertainty requiring a response to unpredictable changes in the internal environment,

- the influence of negative phenomena and processes occurring in the external environment; see Fig. 1.

In the classical V-shaped life cycle model, special importance is attached to actions and procedures to verify and attest the results of planning and project management. Along with traditional analytical technologies, the W-shaped model yields a predictive assessment of the final goals and most significant intermediate results for them. Hence, we pass from the passive consideration of already accomplished events to proactive management of the implementation of target programs.



Fig. 1. The life cycle of program-target planning and management: W-shaped model.

The end-to-end methodology of scenario planning and group management of complex systems under uncertainty increases the efficiency of managing groups of objects within complex SESs (economic entities) by forming and analyzing a wide range of scenarios for the development of these systems and selected sets of their segments (including objects) that consider the influence of the external environment and associated risks of various nature.

The new approach is most efficient in current conditions of growing uncertainty, which forms the need for tools and mechanisms of scenario analysis, simulation, forecasting, planning, and group management. To a large extent, they allow foreseeing future conditions and corresponding alternatives for the development of SESs on different (mostly long) time horizons, identifying external and internal threats and risks, and exploring possible paths of situation development in the internal and external environments.

When elaborating managerial decisions, the methodology under consideration reduces the object of analysis to a finite number of alternatives (including the most probable ones) that reflect the most significant risks and threats to the goals at any stage of the target program life cycle. Moreover, this methodology allows developing optimistic and pessimistic scenarios and making forecasts for the development of given segments of SESs at different management levels, adjusting the strategic vision of the situation and highlighting the desired directions of its development, and assessing the level of coordination between the decisions in different stages of the life cycle of target programs and large projects.

#### CONCLUSIONS

Currently, considerable experience has been accumulated in developing a formal methodology for the analysis and design of models and methods that implement the scenario approach and in solving applied problems in the field of analysis, forecasting, planning, and management of the development of different segments of socio-economic systems. However, most of the solutions focus on the specifics of particular tasks and are limited by their scope. At the same time, there are almost no publications on the integrated application of the scenario approach:

- to increase the efficiency of program-target management for developing and implementing large-scale spatially distributed projects throughout their life cycle,

 to improve the quality and coordination of managerial decisions under uncertainty and risks of different nature.

Further fundamental and applied research in this area will solve a wide range of tasks to improve the efficiency of program-target planning and management processes and the implementation of priority national projects and socio-economic programs.

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This paper was recommended for publication by V.N. Burkov, a member of the Editorial Board.

Received August 16, 2021. Accepted August 31, 2021.

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#### Cite this article

Chernov, I.V. Scenario Methods to Improve the Efficiency of Implementing the Life Cycle of Program-Target Management: A Conceptual Analysis, *Control Sciences* **5**, 77–81 (2021). http://doi.org/10.25728/cs.2021.5.7

Original Russian Text © Chernov, I.V., 2021, published in *Problemy Upravleniya*, 2021, no. 5, pp. 88–93.

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